# INSTRUCTION SUPPLEMENT <br> FOR <br> WJ-8718A/MFP <br> MICROPROCESSOR FRONT PANEL OPTION 

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## WARNING

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## WJ-8718A/MFP INSTRUCTION MANUAL REVISION A CHANGE 1

TITLE:
MANUAL DATE:
CHANGE 1
DATE:
APPLICABILITY:
CHANGES/ERRATA INFORMATION:

CHANGE 1
SUMMARY:

CHANGE:
ERRATA:

INSTRUCTION MANUAL FOR THE WJ-8718A/MFP OPTION
March 1984

February 1991
All units.

Changes refer to updates to the instruction manual to cover design modifications. Errata refer to corrections to and clarifications of information in the manual.

This change adds Mieroprocessor Front Panel installation information.

Add the following paragraph to page 2-1.
None at this time.

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# FIELD MODIFICATIONS INSTRUCTIONS 

FOR INSTALLING THE MFP OPTION
IN A WJ-8718A HF RECEIVER

The following tools are necessary to perform the modification steps:
Phillips screwdriver
Soldering gun and solder
Perform the following steps to install the Microprocessor Front Panel (MFP) option in the WJ-8718A HF Receiver.

1. Remove power from the unit.
2. Remove top and bottom covers from the unit.
3. Remove six screws mounting the manual front panel to the unit. Disconnect all cable assemblies from the old manual front panel except A10P1. Disconnect A10P1 from the Front Panel Interconnect board A6A2. Remove manual front panel from the unit.
4. Place the unit with the bottom side facing up.

## NOTE

E1 thru E59 are adjacent to the double row of capacitors C1 thru C58 depicted in Figure 6-14's side view of the board. Terminals E1 thru E29 are located on the right side of the double row of capacitors (looking from the front of the unit with the unit flipped over). Terminals E30 thru E58 are located on the left side. E1, E29, E30, and E58 are labeled on the board.
5. Cheek the Synthesizer Motherboard (A5) Type number 791570 revision level stamped on the bottom of the board. For A5 boards having revision level $E$ or earlier, perform the following modification to install MFP BITE. Refer to the WJ-8718A HF Receiver Instruction manual.
a. Find A5 location by referring to Figure 5-8.
b. Referring to Figure 6-14, install JW1 by wire wrapping one end of a wire to XA1B, pin B9 (labeled as TP5). Solder the opposite end to terminal E36. To locate E36, find E30 and count down six more terminals.

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c. Referring to Figure 6-14, install JW2 by wire wrapping one end of a wire to XA2A, pin A55 (labeled as TP8). Solder the opposite end to terminal E57. To locate E57, locate E58 and count in one terminal.
d. Referring to Figure 6-14, install JW3 by wire wrapping one end of a wire to XA3, pin A7 (labeled as TP15). Solder the opposite end to terminal E34. To find E34, locate E30 and count four more terminals.
6. Check the IF Motherboard (A4) Type number 791569 revision level stamped on board. For boards having revision level $G$ or later, reposition jumper plug JP1 at J3 for AGC Dump. For boards having earlier revision levels, perform the following modification to install AGC Dump. Refer to the WJ-8718A HF Receiver Instruction Manual.
a. Referring to Figure $\mathbf{5 - 8}$, locate A4 assembly.
b. Referring to Figure 6-3, install a wire by wire wrapping one end of it to XA6, pin 18. Solder the opposite end to E5. See the pin labels at XA1 and XA11 for guide to XA6 pin locations.
7. Refer to Figure 5-8 for A6 location. Check the I/O Motherboard (A6) Type number 791580 revision level. For boards having revision level $M$ or later, reposition jumper plug A6JP1 from X8 pin 3 to X7 pin 59. For boards having earlier revision levels, perform the following modification to configure the board for use with the MFP options. Refer to Figures 5-8 and 6-20 in the WJ-8718A HF Receiver Instruction Manual.
a. Connect E9 of the main chassis (Figure 5-8) to A6X7-59 (Figure 6-20).
8. Route the MFP-W1 cable through one of the cutouts in the chassis of the unit. Notice the red dot at one end of the MFP-XA3 connector. This red dot indicates the location of pin 1 on the connector. Connect MFP-XA3 to XA5 on the I/O Motherboard (A6). Refer to Figure 6-20. Ensure that pin 1 of connector MPP-XA3 is connected to pin 1 of XA5.
9. Place the unit with the top side facing up.
10. Remove the AGC Amplifier (A4A6) type number 78112 from slot A6 on the IF Motherboard (A4). Replace with the AGC Amplifier (MFP-A5) type number 796175. Install the new circuit card into slot A6 on the A4 board. Refer to Figure 5-7 in the WJ-8718A HF Receiver Instruction Manual.
11. Remove the Manual Tuning Up/Down Counter (A6A1) type number 791575 from slots X7 and X8 of the I/O Motherboard (A6).

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794275. Install the new circuit card into slots X7 and X8 on the A6 board. Refer to Figure 5-33 in the WJ-8718A HF Receiver Instruction Manual.
794276. Remove the Front Panel Interconnect (A6A2) from slots XI and X2 of the I/O Motherboard (A6). Replace with the IF Interface (MFP-A3) type number 794308 and cable MFP-W2. Install the new circuit eard into slots X1 and X2 on the A6 board. Refer to Figure 5-33 of the WJ-8718A HF Receiver Instruction Manual.
794277. Referring to Figure $\mathbf{5 - 8}$, perform the following steps to connect the ground lines for the new microprocessor front panel.
a. Solder one end of the MFP-P13 cable to the ground terminal C12 on the chassis.
b. Solder one end of the MFP-P12 to the ground terminal C11 on the chassis.
c. Screw the ground lug of MFP-P11 onto the chassis
794278. Install the new microprocessor front panel by connecting the following cable assemblies.
a. Connect MFP-P21 to the front panel connector A1A1J2.
b. Connect MFP-P1 (contains 16 sockets) to the front panel connector A1J2. Notice the brown dot at one end of the MPP-P1 connector. This brown dot indicates the location of pin 1 on the connector. Ensure that pin 1 of connector MFP-P1 is connected to pin 1 of A1J2.
c. Connect MPP-P2 (contains 14 sockets) to the front panel connector A1J1. Notice the red dot at one end of the MPP-P2 connector. This red dot indicates the location of the pin 1 of the connector. Ensure that pin 1 of connector MFP-P2 is connected to pin 1 of A1J1.
d. Connect MFP-P6 to the front panel connector A1A1J1. The red dot on MPP-P6 indicates the location of pin 1. Ensure that pin 1 of MFP-P6 is connected to pin 1 of A1A1J1.
e. Connect the Line Audio cable from the IF Motherboard (A4) to the line audio potentiometer on the front panel.
794279. Install the new mieroprocessor front panel in place and mount using the same six screws removed in step 3.
794280. Replace top and bottom covers using the same screws removed in step 2.

The information in this addendum is provided to correct and update the WJ-8718A/MFP Instruction Supplement.
1.1 Paragraph 5.6.2.1, Type 796056-1 Front Panel Encode parts list, page 5-16, should be corrected as follows:

1. From: R8*, Resistor, Fixed Film: $1.8 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$, Qty 1, Part No. CF1/4-1.8K/J, Mfr. Code 09021

To: R8*, Resistor, Fixed Film: 680 , 5\%, 1/4 W, Qty 1, Part No. CF1/4-680 OHMS/J, Mfr. Code 09021
2. From: U3, Integrated Circuit, Qty 2, Part No. DS8857N, Mfr. Code 27014

To: U3, Integrated Circuit, Qty 2, Part No. MM74C48N, Mfr. Code 27014
1.2 Revise Type 796056-1 Front Panel Encode (MFP-A1A1) schematic diagram, page 6-7, as shown.


To:

U3
MM74C48N
2. From:


To:
U7
MM74C48N

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Figure 1-1. WJ-8718A/MFP (Microprocessor Front Panel)

## SECTION I

GENERAL DESCRIPTION

### 1.1 ELECTRICAL CHARACTERISTICS

The WJ-8718A/MFP (Microprocessor Front Panel) Option (Figure 1-1) allows local digital control of all receiver parameters while providing push-button access to 100 programmable memory channels. Optional remote control, utilizing the EIA Standard RS-232/C format or the IEEE Standard 488-1975 format, can be accomplished by installing a 232 M or 488M I/O Interface.

Standard local control of a WJ-8718A/MFP Option equipped receiver is enhanced by the use of momentary-contact, push-button switches, with LED indicators. Selection of RF and BFO frequencies is achieved by a keypad or tuning wheel. RF frequency from 5 kHz to 29.99999 MHz is displayed in megahertz to a 10 Hz resolution; BFO frequency between -8 and +8 kHz is displayed in kilohertz to a 10 Hz resolution. The 7 -segment LED displays are available in a choice of three colors: yellow, red, or green.

Ninety-nine discrete (and one implicitly addressed) programmable memory channels allow for the selection of a method of receiver control to suit specific needs. Memory channels can be loaded with receiver parameter data and accessed one at a time or scanned sequentially. Automatic scan of a group of memory channels (with pre-selected start and stop addresses) tunes the receiver sequentially to the stored parameters.

The front panel functions are implemented through an 8-bit microprocessor, and the address, storage, and memory devices required to interface the front panel components with the signal processing circuits in the receiver. The interface circuits and the front panel components form the two key functional sections of the MFP Option.

### 1.1.1 TUNING ENCODER

The Tuning Encoder (MFP-A2) converts the rotation of the front panel tuning wheel into digital pulse trains which are monitored by the microprocessor through a logic circuit on the Front Panel Encode Board (MFP-A1A1).

### 1.1.2 FRONT PANEL SWITCH/ENCODE ASSEMBLY

The Front Panel Switch/Encode Assembly (MFP-A1) consists of a Type 796056-1 Front Panel Encode Board (MFP-A1A1), a Type 794309-X Front Panel Switchboard (MFP-A1A2), interconnecting track, and amplifiers for the headphone audios.

With the exception of the power switch, the switchboard contains all front panel switches, LED indicators, and LED display arrays. Logic circuits on the Front Panel Encode Board interface the switchboard with the microprocessor system. The front panel assembly is functionally dependent on the Type 794275-X and Type 794308-2 Interface Boards, which form the second functional section of the MFP Option. Multiconductor cables interconnect the front panel assembly to the interface boards.

### 1.1.3 IF INTERFACE

The IF Interface (MFP-A3) plugs into the WJ-8718A I/O Motherboard (A6) replacing the Front Panel Interconnect Board (A2).

The IF Interface contains storage registers for detection mode, IF bandwidth, and gain mode data. In addition, the interface couples signal strength voltage for the front panel meter to the Front Panel Switch Encode Board, and switches the line and ISB audio for the front panel phone jack. If an optional 232 M or 488 M I/O Interface is used, the IF Interface provides analog-to-digital conversion of signal strength voltage and digital-to-analog conversion of RF gain voltage.

### 1.1.4 SYNTHESIZER INTERFACE

The Synthesizer Interface (MFP-A4) plugs into the WJ-8718A I/O Motherboard (A6), replacing the standard Manual Tuning Up/Down Counter Board (A6A1). The Synthesizer Interface houses the Intel 8085A Microprocessor, EPROM for storage of microprocessor program and RAM for storage of program variables. The microprocessor, reset at power-up, executes the program stored in EPROM. The program is designed to implement the functions of the front panel by translating the switch settings into digital data for the signal processing circuits in the receiver. In addition, storage registers provide RF and BFO frequency data to the synthesizer section of the receiver.

### 1.1.5 WJ-8718A/488M OPTION

The 488 M Option extends the functional capability of the MFP Option to include a talk/listen compatibility with the IEEE Standard 488-1975 format. The 488M Option consists of the Type 796075 I/O Interface ( $488 \mathrm{M}-\mathrm{A} 3$ ) with the cables, connectors, and hardware to install the interface in the receiver. In addition, two EPROMs programmed to included remote operation are provided to replace the EPROM U1 and U2 on the Synthesizer Interface (MFP-A4).

The I/O Interface is plugged into the I/O Motherboard (A6) and is cable-connected to a digital connection (488-J1) on the rear panel and to the IF and Synthesizer Interface Boards (MFP-A3 and A4). The I/O board handles the bidirectional transfer of data on the 488 bus.

### 1.1.6 WJ-8718A/232M OPTION

The 232 M Option extends the functional capability of the MFP Option to include talk/listen compatibility with the EIA Standard RS-232/C format. The 232 M Option consists of the Type 796037 I/O Interface Board ( $232 \mathrm{M}-\mathrm{A} 3$ ) and the necessary cables, connectors, and hardware to install the board in the receiver. In addition, EPROMs programmed to include remote (232) operation as well as MFP operation are provided to replace EPROM U1 and U2 on the Synthesizer Interface (MFP-A4).

The I/O Interface Board ( $232 \mathrm{M}-\mathrm{A} 3$ ) is plugged into the I/O Motherboard (A6) and is attached via cable to the rear panel digital connections and the IF and Synthesizer Interface Boards (MFP-A3 and A4). The I/O Interface Board handles the bidirectional transfer of data on the RS-232/C interface, allowing remote control of up to 32 "daisy chain" configured receivers, as shown in Figure 1-2.

# Courtesy of http://BlackRadios.terryo.org 

### 1.1.7 WJ-8718A/CORM OPTION

The CORM (Carrier-Operated Relay) Option extends the functional capability of the MFP Option to allow a comparison between the input signal strength and a programmed threshold level. The COR threshold is set by first entering a single digit threshold code on the microprocessor front panel keypad, and then terminating the command by pressing the THRS switch once. The single digit code ( 0 through 9 ) entered with the instruction represents a corresponding signal strength level in dBm . Once the threshold has been established the relay will turn on when an incoming signal strength, greater than the set level, is detected. Consult the WJ-8718A/MFP Operator's Manual for the specific codes and their respective threshold levels.

If a CORM equipped receiver is also equipped with the WJ-8718A/232M or 488 M Option, the COR threshold can be controlled remotely. Four bits of threshold data can be commanded and monitored, and a fifth bit (COR ON/OFF flag) can be monitored. Refer to the WJ-8718/CORM Instruction Supplement for any additional information.

### 1.2 MECHANICAL CHARACTERISTICS

The front panel section of the MFP Option is an assembly formed by mounting the Type 791202-5 Encoder Assembly (MFP-A2) and the Type 794310 Front Panel Switch Encode Assembly (MFP-A1) on the rear of the microprocessor front panel. The completed assembly replaces the standard WJ-8718A HF Receiver front panel and the following standard front panel mounted boards: Manual Tuning Module (A7), Frequency Display (A8), BFO Switch (A9), and Front Panel Control (A10).

The front panel is constructed of aluminum with a photo-etched bezel overlay. Momentary-contact push buttons are mechanically arranged in functional blocks. Within each functional block, selection of a new operating parameter automatically deactivates a previously selected switch, preventing simultaneous activation of buttons with inverse functions. With the exception of the power switch and general purpose keys, each front panel switch contains an LED indicator which glows when the switch is activated.

Memory address, BFO frequency, and RF frequency are displayed by 7 -segment LEDs. Frequencies are selected via keypad or a front panel tuning wheel. In addition, there are three LED fault lights which glow red on various invalid conditions. The remaining standard WJ-8718A front panel features include a signal strength/line audio level meter, a stereo phone jack, and potentiometers for phone level and RF gain control.

The rear panel contains an adapter plate for the particular set of options installed in the receiver.
1.3 EQUIPMENT SUPPLIED

The MFP Option consists of the following circuit boards:
a) Type 794275-X Synthesizer Interface
b) Type 794308-2 IF Interface

## Courtesy of http://BlackRadios.terryo.org

c) Type 791202-5 Encoder Assembly
d) Type 794310-X Front Panel Switch/Encode Assembly
e) Type 796056-1 Front Panel Encode Board
f) Type 794309-X Front Panel Switchboard

The Encoder Assembly and the Front Panel Switch/Encode Assembly with its subassemblies, the Front Panel Switchboard and Front Panel Encoder, are pre-mounted on the rear of the MFP Option front panel. Two ribbon cables, MFP-W1 and MFP-W2, are supplied to interconnect the front panel assembly and the two interface boards.

### 1.4 EQUIPMENT REQUIRED BUT NOT SUPPLIED

A 600 ohm stereo headphone set is required to monitor the audio at the front panel phones jack. Mono headphones could be used, but LSB would not be monitored in the ISB mode. If either the 232 M or 488 M Option is used, compatible remote control equipment is required.

### 1.5 OPTIONAL EQUIPMENT

The 232M Option consists of the Type 796037 Asynchronous I/O Board, two cables (232-W2 and $232-$ W3), an adapter plate (with mounting hardware), two 25 -pin D-series connectors, and two EPROMs.

The 488M Option consists of the Type 796075 I/O Interface, one cable (488-W2), an adapter plate, one 24 -pin Amphenol connector (with mounting hardware), and two EPROMs.

# Courtesy of http://BlackRadios.terryo.org 

## SECTION II

## INSTALLATION AND OPERATION

### 2.1 UNPACKING AND INSPECTION

Paragraph 2.1 of the WJ-8718 Series HF Receiver Instruction Manual is applicable to the WJ-8718A/MFP Option.

### 2.2 PREPARATION FOR RESHIPMENT AND STORAGE

Paragraph 2.2 of the WJ-8718 Series HF Receiver Instruction Manual is applicable to the WJ-8718A/MFP Option.

### 2.3 INSTALLATION

Paragraph 2.3 of the WJ-8718 Series HF Receiver Instruction Manual is applicable to the WJ-8718A/MFP Option.

### 2.4 OPERATION

### 2.4.1 PREPARATION FOR OPERATION

A receiver equipped with an MFP Option can be readied for manual operation by consulting paragraphs 2.1 through 2.3 in the WJ-8718A HF Receiver Instruction Manual. Proper operation of the MFP Option can be tested by execution of a Built-In Test Equipment (BITE) program, described in paragraph 4.1.1.1.

If in addition to the MFP Option, a receiver is equipped with either the 232 M or 488 M remote interface option, it can be readied for remote operation by consulting the appropriate instructions in paragraph 2.4.2 (232 operation) or paragraph 2.4.3 (488 operation).

### 2.4.2 PREPARATION FOR EIA STANDARD 232 OPERATION

Before remote operation, a WJ-8718A/MFP Receiver with the 232M Option must be placed in the remote operating mode. The appropriate parameters must then be set on the I/O board switches and connections as described in the following paragraphs. Refer to Figure 6-8, Type 796037 I/O Interface Schematic Diagram, when reading the following descriptions.

### 2.4.2.1 Remote Operating Mode

A receiver designated as a listener must be placed in the remote operating mode by depressing the front panel REMOTE button or if the receiver has not previously been commanded by a remote device, the first remote message received after power on will place
the receiver in a remote state. The receiver can then be addressed to talk regardless of the REMOTE switch position. The local or remote operating modes cannot be established by remote command.

### 2.4.2.2 Baud Rate

The baud rate is established by precoded entries to the S1 switch assembly on the I/O Interface Board. An open switch denotes binary 1 and a closed switch denotes binary 0. Table 2-1 lists the baud rate codes. The switches are factory adjusted for a baud rate of 1200 .

Table 2-1. Baud Rate Codes

| Baud Rate | S1-4 | S1-3 | S1-2 | S1-1 |
| :---: | :---: | :---: | :---: | :---: |
| 50 | 0 | 0 | 0 | 0 |
| 75 | 0 | 0 | 0 | 1 |
| 110 | 0 | 0 | 1 | 0 |
| 134.5 | 0 | 0 | 1 | 1 |
| 150 | 0 | 1 | 0 | 0 |
| 300 | 0 | 1 | 0 | 1 |
| 600 | 0 | 1 | 1 | 0 |
| 1200 | 0 | 1 | 1 | 1 |
| 1800 | 1 | 0 | 0 | 0 |
| 2000 | 1 | 0 | 0 | 1 |
| 2400 | 1 | 0 | 1 | 0 |
| 3600 | 1 | 0 | 1 | 1 |
| 4800 | 1 | 1 | 0 | 0 |
| 7200 | 1 | 1 | 0 | 1 |
| 9600 | 1 | 1 | 1 | 0 |
| $19200 *$ | 1 | 1 | 1 | 1 |

### 2.4.2.3 Receiver Address, Parity, and Master/Slave

Up to 32 properly addressed receivers can be controlled by one remote device. As shown in Figure 2-2, the receivers are series connected, or "daisy chained", and only one receiver is interfaced directly to the remote control equipment. The serial data stream from the remote control equipment is actively repeated to all receivers in the chain, but only the addressed receiver "recognizes" its address and accepts the data.

Receiver address is established by five switch settings in the S2 switch assembly on the I/O board. Table 2-2 lists the functions of the switch assembly. A closed switch denotes a binary 1 and an open switch denotes a binary 0 . Valid receiver addresses are binary coded from 0 through 31, as listed in Table 2-3.

Table 2-2. Switch Assembly (S2) Functions

| S2-8 | S2-7 | S2-6 | S2-5 | S2-4 | S2-3 | S2-2 | S2-1 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Master/Slave | Parity Enable | Parity Bit | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |
|  |  |  |  | Receiver Address |  |  |  |

Switch S2-8 is opened (0) if the receiver is to be a slave and be commanded and monitored by a remote control device. Switch $\mathrm{S} 2-8$ is closed (1) if the receiver is to be a master and constantly transmit parameter data, allowing master-slaving of two receivers.

Switches S2-6 and S2-7 are used if the remote control equipment provides a parity bit. If S2-7 is closed, a parity bit is expected with the remote transmission; if S2-7 is open, a parity bit cannot be sent. S2-6 is closed (1) if even parity is desired and opened (0) if odd parity is desired.

Table 2-3. Receiver Address Codes

| Receiver Address | S2-5 | S2-4 | S2-3 | S2-2 | S2-1 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 0 | 1 |
| 2 | 0 | 0 | 0 | 1 | 0 |
| 3 | 0 | 0 | 0 | 1 | 1 |
| 4 | 0 | 0 | 1 | 0 | 0 |
| 5 | 0 | 0 | 1 | 0 | 1 |
| 6 | 0 | 0 | 1 | 1 | 0 |
| 7 | 0 | 1 | 1 | 1 | 1 |
| 8 | 0 | 1 | 0 | 0 | 0 |
| 9 | 0 | 1 | 0 | 0 | 1 |
| 10 | 0 | 1 | 0 | 1 | 0 |
| 11 | 0 | 1 | 1 | 0 | 1 |
| 12 | 0 | 1 | 1 | 0 | 0 |
| 13 | 1 | 1 | 1 | 1 | 0 |
| 14 | 1 | 0 | 0 | 1 | 1 |
| 15 | 1 | 0 | 0 | 0 | 0 |
| 16 | 1 | 0 | 0 | 1 | 1 |
| 17 | 1 | 0 | 1 | 1 | 0 |
| 18 | 1 | 0 | 1 | 0 | 0 |
| 19 | 1 | 0 | 1 | 0 | 1 |
| 20 | 1 | 0 | 1 | 1 | 0 |
| 21 | 1 | 1 | 0 | 0 | 1 |
| 22 | 1 | 1 | 0 | 0 | 0 |
| 23 | 1 | 0 | 1 | 0 |  |
| 24 |  |  |  |  | 0 |
| 25 |  |  |  |  |  |

Table 2-3. Receiver Address Codes (Cont'd)

| Receiver Address | S2-5 | S2-4 | S2-3 | S2-2 | S2-1 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 27 | 1 | 1 | 0 | 1 | 1 |
| 28 | 1 | 1 | 1 | 0 | 0 |
| 29 | 1 | 1 | 1 | 0 | 1 |
| 30 | 1 | 1 | 1 | 1 | 0 |
| 31 | 1 | 1 | 1 | 1 | 1 |
| NOTE: $0=$ open, 1 closed |  |  |  |  |  |

### 2.4.3 PREPARATION FOR IEEE STANDARD 488 OPERATION

Through the use of the WJ-8718/488M Option, up to 14 parallel-connected receivers can be controlled by one controller. Before remote operation, each receiver must be placed in the remote operating mode, and a valid receiver address must be established. Refer to Figure 6-7, Type 796075 I/O Interface Schematic Diagram, when reading the following information.

### 2.4.3.1 Remote Operating Mode

A receiver, designated as a listener, must be placed in the remote operating mode by depressing the front panel REMOTE button or if the receiver has not previously been commanded by a remote device, the first remote message received after power on will place the receiver in a remote state. The receiver can be addressed to talk regardless of the REMOTE switch position. The remote or local operating modes cannot be established by remote command, but can be monitored.

### 2.4.3.2 Receiver Address

Parallel data from the controller is applied to all receivers under control, but only the addressed receiver "recognizes" its address and accepts the data. The 5-bit address code is established by switch settings in the S1 switch assembly on the I/O Interface (488-A3). A closed switch denotes a logic 1 and an open switch denotes a logic 0 . A valid receiver address can be any binary-coded number in the 0 through 30 range, as listed in Table 2-3.

The logic levels established by the switch settings are inverted at the outputs of tri-state buffer U4, and are loaded on the microprocessor bus lines when the processor addresses U4 and initiates a read operation. The processor writes the receiver address into a register internal to the microprocessor interface, 488-A3U1, and, until U1 is reset, the interface (488-A3U1) determines if the receiver address on incoming data is valid. The ability of U 1 to perform this service ensures that the microprocessor is interrupted only when a valid data byte is waiting to be fetched, or a byte is needed for transmission.

### 2.4.4 MANUAL OPERATING PROCEDURE

Local control of the receiver is accomplished through momentary-contact pushbutton switches arranged in functional blocks on the WJ-8718A/MFP Option panel, illustrated in Figure 2-1 and described in paragraphs 2.4.4.1 through 2.4.4.10. Optional remote operation of the receiver is described in paragraph 2.4.5.

### 2.4.4.1 POWER PUSH ON/OFF Switch (S1)

The POWER switch is pressed alternately to energize and de-energize the receiver. In the ON position, the button is depressed and has the appearance of illumination due to a brightly colored mechanical insert visible through the transparent switch cover. In the OFF position, the button assumes the out position and the colored insert is retracted.

Before energizing the receiver, the rear panel voltage-select PC wafer should be examined as described in the WJ-8718A HF Receiver Instruction Manual.


Figure 2-1. Microprocessor Front Panel

### 2.4.4.2 General Purpose Keypad

The general purpose keypad is the 16 -button switch block located to the left of the tuning wheel. The multi-function, numerical selection buttons labeled 0 through 9 are utilized to enter RF and BFO frequencies, signal threshold level, BFO offset, dwell time, and memory address. A special function key (*) is used to place the receiver in the BITE mode, and to clear the numbered memory channels.

### 2.4.4.2.1 Terminator Functions

The software structure of the microprocessor in the receiver control block is such that an accepted numerical keypad entry is not acted upon until a termination switch is activated. The terminator informs the microprocessor how the number is to be used. Table 2-4 lists the termination switches and their associated function.

Table 2-4. Termination Switches

| Switch | Function |
| :--- | :--- |
| MHz | RF frequency |
| kHz | BFO or RF frequency |
| RECALL | Memory address |
| STORE | Memory address |
| EXAM | Memory address |
| DWELL | Scan pause, in seconds |
| THRS | Signal level threshold |
| LOCK OUT | Memory address |
| Special Function (*): | Access BITE (used with 17) |
|  | Receiver BITE Tests |
|  | Clears memory channels 1-16 |
|  | (used with 10) |

### 2.4.4.2.2 CLEAR Keypad Switch

The CLEAR key will remove an unterminated numerical entry from the receiver's internal circuits. Once the terminator is entered, the CLEAR switch does nothing. The CLEAR switch is also used to release the EXAM mode.

The EXAM mode is used to examine the contents of a memory channel without changing receiver parameters. During the examination of a memory channel, the front panel LEDs reflect the contents of the addressed channel, but the receiver's operating parameters remain as before the examining procedure. Activation of the CLEAR key, during the EXAM mode releases the EXAM mode and restores the current operating parameters to the front panel LEDs. Refer to paragraph 2.4.4.7.1 for more details on the EXAM mode.

### 2.4.4.2.3 Special Function Switch (*)

The * key is a special function key used, in this application, to access the BITE program mode, to access the Receiver BITE Tests, and to clear all numbered memory channels. The * key also turns turns on/off special operating modes, such as; handoff, master/slave and dwell.

The * key provides access to a Built-In Test Equipment (BITE) program which provides a diagnostic test of the front panel switches. Because the execution of the program is software dependent, the program also provides a test of the microprocessor, selected signals, EPROM, and data lines. The program is accessed by engaging, in order, the 1, 7, and * keys. The immediate effect of the BITE mode is illumination of all front panel LEDs and display of all segments in each 7 -segment display (except the BFO sign).

Entering 1, 8, and * in order on the keypad invokes the Receiver BITE Tests. These seven tests examine, in order, RAM, $+15 \mathrm{~V},-15 \mathrm{~V}$, 1st LO, 2nd LO, Bandwidth Selection and BFO Tuning, producing an error code in the process. The test in progress is displayed by number in the BFO OFS/THRS LVL display window. The error code is a binary-weighted code. Any errors detected are summed and displayed in the FREQUENCY/MHz display window at the end of the seventh test. The CLEAR key is used to exit BITE tests.

The * key is used to erase the contents of the numbered memory channels. Each memory channel ( 1 through 99) is cleared when the $9,9,9$, and * keys are pressed in the given sequence.

### 2.4.4.3 REMOTE and LOCAL Switches

The LOCAL switch places the receiver in local mode, allowing front panel control of receiver parameters. The REMOTE switch allows remote control operation through a data stream to the rear panel. Remote control is effective only if the receiver is equipped with an optional remote interface ( 232 M or 488 M ).

### 2.4.4.4 RF Frequency

The WJ-8718A HF Receiver is designed to receive and demodulate RF frequencies in the 5 kHz to 29.99999 MHz range. The tuned frequency of the receiver can be locally established via the general purpose keypad or the tuning wheel.

### 2.4.4.4.1 RF Frequency: Keypad Entry

To enter a tuned frequency via the keypad, the number is entered in order, from MSD to LSD, and terminated with kHz or MHz . An unterminated, undesired numerical entry can be cleared from the receiver's internal circuits by engaging the CLEAR switch. As soon as the terminator is entered, the operation is complete and the CLEAR switch has no effect. If the ENTRY FAULT LED glows after entry of an RF frequency, an invalid frequency has been entered. To disengage the FAULT light, re-enter the desired frequeney and terminate with kHz or MHz.

Use of the kHz or MHz terminator is a matter of choice. For example, 123 kHz could be entered $1,2,3, \mathrm{kHz}$ or decimal (.), $1,2,3, \mathrm{MHz}$. In either case, the display will read 00.12300 .

Entry of the decimal point (.) is unnecessary if a whole number is entered. For example, to enter 4 MHz , press the digit 4 key and the MHz key. The internal circuitry will take care of the decimal and the display will read 04.00000.

### 2.4.4.4.2 RF Frequency: Tuning Wheel Entry

The one-finger tuning wheel on the Microprocessor Front Panel can be used to dial the tuned RF frequency to a selected resolution. The tuning resolution switches are in the TUNE switch block to the right of the tuning wheel: FAST represents 1 kHz steps, MED represents 100 Hz steps, and SLOW represents 10 Hz steps. Selection of a tuning resolution disengages the LOCK switch and the tuning wheel can be used to dial the frequency.

The tuned RF frequency display is decremented at the selected tuning resolution if the tuning wheel is rotated counterclockwise. It is incremented, at the selected tuning resolution, if the tuning wheel is rotated clockwise. Tuning past the end of the range 00.00000 to 29.9999 MHz ) will cause the frequency to step to the other end of the band and continue tuning in the same increasing or decreasing direction. The display will lock on the registered frequency when the LOCK button is engaged and the tuning wheel will be ineffective. The LOCK switch has no effect on keypad RF frequency entry.

### 2.4.4.4.3 FREQUENCY/MHz Display

RF frequency is displayed by a 7-digit, 7-segment LED readout. As the frequency is entered, from MSD to LSD on the front panel keys, each digit is displayed in the LSB position and is displaced to the left one space with each new numerical entry. This calculator-style display reflects all numerical entries to the general purpose keypad, except BFO frequency, until the terminator is entered.

### 2.4.4.5 BFO Frequency

### 2.4.4.5.1 BFO Frequency: Keypad Entry

To enter a BFO frequency via the keypad, the CWV and $+/-$ BFO key must be engaged. An entered number between -8.00 and +8.00 is terminated with kHz , and is displayed by the BFO display.

### 2.4.4.5.2 BFO Frequency: Tuning Wheel Entry

The one-finger tuning wheel can be used to dial a BFO frequency if the BFO and CWV switches are engaged. Selecting a tuning resolution via the TUNE switch block to the right of the wheel disengages the LOCK switch and allows the tuning wheel to dial a BFO frequency. The only effective tuning resolutions for BFO tuning are the 100 Hz rate (FAST) and the 10 Hz rate (MED and SLOW). Rotating the wheel clockwise will increment the displayed frequency from a negative offset, through zero, to the upper limit of the range. Rotating the wheel counterclockwise will decrement the displayed frequency from a positive offset, through zero, to the lower limit of the range.

### 2.4.4.5.3 +/-BFO Offset Switch

The offset sign can be changed by the BFO $+/-$ key. Pressing the key causes the minus $(-)$ sign to be alternately displayed and removed. Absence of the sign on the display indicates positive offset.

### 2.4.4.5.4 BFO OFS/THRS LVL Display

The $+/-$ BFO frequency is displayed on a 3 -digit plus a minus sign (-), 7-segment LED display. The BFO frequency can be entered locally via a keypad or tuning wheel. The BFO
frequency range is $+/-8.00 \mathrm{kHz}$ and is displayed in kHz to a 10 Hz resolution. If the receiver is not equipped with the 10 Hz BFO Option, the LSD of the displayed BFO frequency is not connected to the synthesizer.

The LSD of the BFO OFS/THRS LVL display is also used to display dwell time and threshold levels, as deseribed in paragraphs 2.4.4.7.4 and 2.4.4.7.5.

### 2.4.4.6 Receiver Operating Parameters

Receiver operating parameters are entered on the Microprocessor Front Panel switches and transferred to the receiver automatically. Parameters can be stored in a memory using the STORE switch, and transferred to the receiver from memory using the RECALL switch. See paragraph 2.4.4.7 for a discussion of the memory channels.

Selectable parameters are detection mode, IF bandwidth, RF gain (MGC only), and gain mode. The parameters are discussed in paragraphs 2.4.4.6.1 to 2.4.4.6.4.

### 2.4.4.6.1 Detection Mode

Detection modes are selected by the AM, FM, USB, LSB, ISB, CWV (variable), and CWF (fixed) pushbuttons. The CWV switch is associated with the BFO switch and must be engaged to establish a new BFO frequency in the addressed receiver (see paragraph 2.4.4.5).

In the AM, FM, and CW modes, the line audio and phones audio are taken from the receiver's AM, FM, and CW detectors, respectively. One of the five available bandwidths can be selected for AM, FM, and CW modes. The CWV switch enables the variable BFO. The CWF switch enables the fixed BFO ( 455 kHz ). In the ISB mode, USB and LSB audios are available and can be monitored with stereo headphones.

### 2.4.4.6.2 IF Bandwidth

The IF BANDWIDTH/kHz switch block allows selection of . $3,1,3.2,6$, and 16 kHz IF bandwidths. In the AM, FM, and CW detection modes, any one of the five bandwidths should be selected. Bandwidth selection is ineffective in the USB, LSB, and ISB detection modes, when a sideband filter bandwidth of approximately 3 kHz is assumed. To indicate this, the LED on the 3.2 kHz IF bandwidth switch will glow.

### 2.4.4.6.3 Gain Mode

The GAIN MODE switches are MGC (Manual Gain Control), SLOW (AGC), and FAST (AGC). The differences in decay times of fast AGC and slow AGC make them useful for different kinds of signals. In the fast AGC mode, the gain of the receiver adjusts about as quickly for a rise in signal strength (attack time) as it does for a fall in signal strength (decay time). Fast AGC attack and decay time is approximately 15 milliseconds for each. In the slow AGC mode, attack time is also fast, but decay time is approximately 2 seconds slower. For AM and FM signals, the total power in the carrier and sidebands does not vary much with time at the transmitter. The main purpose for AGC is to compensate for atmospheric losses; therefore, fast AGC is best for AM and FM signals. For pulsed signals, such as telegraphy and single

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sideband voice signals, there are rapid fluctuations in transmitted power with recurring peaks. When this type of signal is received, it is desirable that AGC attacks quickly and decays slowly; therefore, slow AGC is desirable for USB, LSB, and ISB signals.

When it is desirable to fix the gain, to make comparisons of signal strength, or to eliminate signals or noise below a particular amplitude, the manual gain mode should be used. In this mode, the RF GAIN control can be engaged and should be adjusted so that the signal strength meter registers at the MAN SET line, as discussed in paragraph 2.4.4.6.4.

### 2.4.4.6.4 RF Gain Control

When the receiver is in the MGC (Manual Gain Control) mode, rotating the RF GAIN control clockwise approximates a linear increase in receiver gain. To obtain the greatest latitude for signal level change, engage the SIGNAL STR meter switch and set the gain control for an indication at the MAN SET mark on the meter.

### 2.4.4.7 Programmable Memory Channels

### 2.4.4.7.1 MEM ADRS Display

The MFP accesses 99 channels of addressable memory in locations 1 through 99. Memory location is entered on the keypad and is terminated with RECALL or STORE. As soon as the terminator is entered, the address is displayed on the 2-digit, 7-segment LED MEM ADRS display. When terminator STORE is selected, the memory location is loaded with the receiver parameters registered on the front panel. When terminator RECALL is selected, the parameters in the recalled memory channel are transferred to the receiver.

Use of the EXAM key allows the contents of the numbered memory channels to be examined without affecting receiver operation. The EXAM key can be used to examine a selected memory channel or can be used to step sequentially through the numbered memory channels.

To examine the contents of a single memory channel, enter the memory address on the general purpose keypad before pressing the EXAM key. The contents of the addressed channel will be reflected by the front panel LEDs, but the receiver parameters will remain unchanged. The MEM ADRS display will blink on/off for the memory chanel being examined.

To step sequentially through the numbered memory channels, press the EXAM key with no preceding memory address. The front panel LEDs will reflect the contents of a memory channel, and the address of that channel will appear in the MEM ADRS display. The channel that is initially accessed is determined by the conditions of the preceding memory access. Successive use of the EXAM key will access the memory channels in numerical sequence up to 99 , then back to 1 again until the EXAM mode is released.

Activation of the CLEAR key releases the EXAM mode and restores the front panel LEDs to the receiver operating parameters in effect before (and during) the EXAM procedure.

The * key is also used to clear the contents of the numbered memory channels. To perform this function, the activation of the * key must be preceded by entry of $9,9,9$ on the general purpose keypad.

### 2.4.4.7.2 100th Memory Channel

A 100th memory channel needs no keypad entry to be addressed. The channel is accessed by engaging the STORE or RECALL switches only (no numbered address). The MEM ADRS display goes blank to indicate the 100 th channel access. The 100 th memory channel provides the operator with a rapid access memory channel.

### 2.4.4.7.3 Scan Mode

The MFP memory scan capability allows the receiver to sequentially sean a selected group of memory channels. The memory locations are pre-loaded with receiver data (see paragraph 2.4.4.7.1).

During scan operation, the stored parameters of each memory channel are automatically recalled and transferred to the receiver. As each channel is recalled, the received signal strength is compared to a selected threshold level stored in that channel. When a channel is located where signal strength equals or exceeds the threshold level, the scanning operation stops for a selected dwell time, and then restarts automatically.

Pressing AUTO SCAN initiates a sequential scan of the memory channels. If none of the memory channels are locked out of the scan, the receiver scans through all channels except those reserved for frequency sector scan data storage (channels 87 through 99).

To omit a memory channel from the scan, enter the memory address on the keypad, then press the LOCK OUT key. To remove a group of locations, enter the first and last numbers of the group on the keypad, separated by a decimal point. To restore a channel or channels, use the same procedure, except press LOCK OUT twice.

### 2.4.4.7.4 Signal Level Threshold

The threshold level is a level selected for scan modes, which is compared to signal strength. The comparison determines whether a signal is strong enough to be monitored. Threshold level is selectable from 0 to 9 . To establish threshold level, the number is entered on the keypad and terminated with THRS. The threshold level will be displayed in the LSD of the BFO OFS/THRS LVL display. Threshold level must be selected and stored independently in each memory channel.

### 2.4.4.7.5 Dwell Time

If AUTO SCAN mode is to be utilized, a dwell time must be selected. Dwell time is selectable from 0.1 to 8 seconds. A zero entry to the keypad selects 0.1 seconds, a 1 selects 1 second, a 2 selects 2 seconds, etc. The desired number is entered on the keypad and terminated with the DWELL push button. The selected dwell time is automatically transferred to each memory channel. If dwell time is not selected, the 0.1 second dwell time will be assumed. In automatic scan mode, the pre-loaded memory channels are sequentially scanned
and the parameters transferred to the receiver. When an active channel is located (signal strength equal to or greater than threshold level) the scan locks at the active channel for the established dwell time and restarts automatically. If a dwell time of 9 is entered, the scan stops until restarted by pressing AUTO SCAN.

### 2.4.4.7.6 Hand Off Operation

The HAND OFF switch is utilized during Master/Slave hand off operations. Master/Slave operation is possible only with receivers that are equipped with the WJ- $8718 \mathrm{~A} / 232 \mathrm{M}$ or $\mathrm{WJ}-8718 \mathrm{~A} / 488 \mathrm{M}$ Remote Control I/O Option. In such cases, a master receiver can hand off front panel data to a number of slave receivers of the same type. In order to hand off front panel data from the master to the slaves, the master receiver must first be in master mode. This is accomplished by entering $25^{*}$ on the front panel keypad. Once the receiver is in master mode, its front panel parameters may be transferred to the slave receivers by depressing the master's front panel HAND OFF switch. The HAND OFF LED indicator light should light momentarily, indicating that the hand off has been accomplished. If the indicator light remains on, the hand off has not taken place. Entering 26* turns off the Hand Off Special Mode.

### 2.4.4.8 LINE AUDIO/SIGNAL STR Meter

The LINE AUDIO/SIGNAL STR meter has two scales. One scale displays the signal strength from 0 to 110 dB . A MAN SET mark on the scale indicates the proper signal strength for use in the manual gain mode. The second scale indicates the line audio level above 1 mV , referenced to 600 ohms . Meter function is determined by selection of one of the two meter switches.

### 2.4.4.8.1 LINE AUDIO Meter Switch

When the LINE AUDIO switch is engaged, the meter indicates the line audio level. The line audio (AUDIO 1) level is adjusted with the LINE AUDIO LEVEL potentiometer (R1) located on the rear panel.

### 2.4.4.8.2 SIGNAL STR Meter Switch

When the SLOW or FAST GAIN MODE switches (see paragraph 2.4.4.6.3) are engaged, the meter provides a logarithmic indication of signal strength. When the MGC switch is engaged, the meter provides a near-linear indication of receiver AM detector voltage. The RF GAIN potentiometer on the receiver front panel can then be used to control signal strength (see paragraph 2.4.4.6.4).

### 2.4.4.9 PHONES Jack

The audio signals from the receiver can be monitored at the PHONES jack on the front panel. The PHONES jack is designed to drive a 600 ohm stereo headphone set. When operating in the ISB mode, both USB and LSB audio can be monitored simultaneously: the USB

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components of the signal are on the phones jack tip contact and the LSB components of the signal are on the ring contact. In all other modes, only line audio is available and could be monitored with either mono or stereo headphones.

### 2.4.4.9.1 PHONE LEVEL Control

The PHONE LEVEL control is a concentric potentiometer. The outer ring of the control is rotated to vary the level of AUDIO 1 at the tip contact of the stereo PHONES jack. The inner shaft is rotated to vary the level of AUDIO 2 at the ring contact. Clockwise rotation increases audio level and counterclockwise rotation decreases audio level. In all detection modes except ISB, both audios are line audio. In ISB mode, AUDIO 1 is USB and AUDIO 2 is LSB. If mono headphones are in use, AUDIO 2 will not be monitored and rotation of the inner ring of the PHONE LEVEL control will have no audible effect.

### 2.4.4.10 Fault Lights

The Microprocessor Front Panel contains DATA, ENTRY, and RCVR fault indicators. If optional remote control is in use, the DATA fault indicator glows red to inform the operator that invalid data has been transferred in the remote data stream. The ENTRY fault indicator illuminates if an invalid number or out-of-sequence entry has been made to the front panel switches. An invalid numerical entry could be any number outside the range of the terminating function. An out-of-sequence entry could be a terminator entry preceding a numerical entry. In either case, the light ceases to glow when a valid entry is made. The RCVR fault indicator glows red during normal operation if a power supply fault or LO unlocked signal is detected.

### 2.4.5 REMOTE OPERATING PROCEDURES (OPTIONAL)

The remote operating procedures apply only to MFP Option equipped receivers, which are also equipped with a remote I/O interface ( 232 M or 488 M ).

### 2.4.5.1 232M Operation

A receiver equipped with the WJ-8718A/232M Option is ready to be interfaced with a remote control device compatible with the RS-232/C EIA Standard Interface. The option utilizes 8 signals: two for ground, two for data transfer, and four for handshake protocol. The interface point between receiver and remote control equipment is the 25 -pin, D-series connector at the REMOTE INPUT port on the receiver rear panel. A 25-pin connector at the receiver rear panel MONITOR OUTPUT port is the interface point for connecting the next receiver in the series chain. Up to 32 WJ -8718A HF Receivers, equipped with the WJ-8718A/MFP and 232M Options and configured as slaves (see paragraph 2.4.2.3), can be controlled by one remote control device, as shown in Figure 2-2.

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FIGURE 2-2


Figure 2-2. Daisy Chain Configuration

Table 2-5 lists the RS-232/C connection pin numbers with the respective interchange signals and functions. The connections between receivers will be pin-for-pin compatible if each receiver is configured as a slave as shown in Figure 2-3. If one receiver is configured as a master and the remainder are slaves, it is necessary to use reverse cable (modem bypass) with pin connections to the slave units as shown in Figure 2-4.

Table 2-5. RS-232/C Connector Pin Assignments

| Pin Number | Signal | Description |
| :---: | :--- | :--- |
| 1 | P-GND | Protective Ground |
| 2 | TxD | Transmitted Data |
| 3 | RxD | Received Data |
| 4 | RTS | Request to Send |
| 5 | CTS | Clear to Send |
| 6 | DSR | Data Set Ready |
| 7 | S-GND | Signal Ground |
| 20 | DTR | Data Terminal Ready |



Figure 2-3. Receiver-To-Receiver Cable Detail


Figure 2-4. Master/Slave Configuration

## Courtesy of http://BlackRadios.terryo.org

FIGURE 2-5

### 2.4.5.2 Control and Monitor Data Format

### 2.4.5.2.1 Byte Structure

The I/O software structure of the WJ-8718A/232M Option consists of a series of 11-bit binary coded words (if parity is used), presented in a pre-determined sequence. Each byte is composed of a start bit, eight data bits, a parity bit (if parity is enabled), and a stop bit. Figure 2-5 illustrates the byte structure. Binary coded logic levels are high ( -6 V ) or low ( +6 V ).


Figure 2-5. Byte Structure

### 2.4.5.2.2 Data Format: Tier 1

Full status parameter data of a standard WJ-8718A HF Receiver equipped with the WJ-8718A/MFP and 232 M Options are contained in seven bytes. These seven bytes, plus address and data definition bytes and a byte for Tier 2 access, comprise Tier 1 of the I/O sof tware structure. Tier 2 is discussed in paragraph 2.4.5.2.6. Tier 1 is structured as follows:

1. Byte 1: Receiver Address Byte.
2. Byte 2: Data Information Definition Byte (DID).
3. Bytes 3 through 9 (10): In monitor mode, Byte 3 is the address byte returned from the receiver to controller and Bytes 4 through 10 are data. In command mode, Bytes 3 through 9 are parameter data (controller to receiver).
4. Byte 10 or 11: Tier 2 Access Byte.

### 2.4.5.2.3 Receiver Address Byte

The data bits in the first byte (controller to receiver) determine which receiver is to be referenced. Up to 32 receivers can be controlled and monitored by one controller (such as the WJ-9644A Asynchronous Controller) via the WJ-8718A/232M remote interface. As shown in Table 2-6, five bits of the receiver address byte allow for 32 unique binary-coded addresses from 0 through 31 (Table 2-3). The three most significant bits of the address are a binary code (110) that uniquely identifies the function (receiver address) of the byte.

In the monitor mode, the receiver address byte is returned to the controller by the receiver prior to the transmission of data (receiver to controller).

Table 2-6. Receiver Address Byte


### 2.4.5.2.4 Data Information Definition (DID) Byte

Byte 2 is the Data Information Definition (DID) Byte, as shown in Table 2-7. The three most significant bits of the DID byte are a binary code (111) which uniquely defines Byte 2 as a DID byte. The $2^{4}$ bit of the DID byte is the command/monitor bit, which defines the function of the data bytes to follow Byte 2. If the $2^{4}$ bit is logic 1 , the controller desires to command the receiver and will transmit receiver parameter data in the byte(s) following Byte 2. If the $2^{4}$ bit is logic 0 , the controller desires to monitor the receiver, which will transmit the address byte (paragraph 2.4.5.2.3), and the byte(s) following, which contain the current status of the addressed receiver.

Table 2-7. DID Byte
$\left.\begin{array}{|ccc|c|l|lll|}\hline 2^{7} & 2^{6} & 2^{5} & 2^{4} & 2^{3} & 2^{2} & 2^{1} & 2^{0} \\ \hline 1 & 1 & 1 & & & & & \\ \begin{array}{c}\text { DID } \\ \text { Byte Code }\end{array} & \text { C/M } & \text { S/A } & \begin{array}{c}\text { Register: } \\ 0 \\ 0\end{array} & 0-1 & 1 & 0\end{array}\right]$

The $2^{3}$ bit of the DID byte defines the number of data bytes to succeed Byte 2: a logic 1 indicates a single byte, and a logic 0 indicates all bytes.

The remaining three bits of the DID byte contain a binary-coded register address. All receiver parameters are stored in seven 8 -bit registers, addressed in binary code from 0 through 6. If all parameter data are to be transmitted, the logic levels of the three register address bits are irrelevant. However, if one word of receiver parameter data is to be transmitted, the 3 -bit binary code for the desired register must be established in the $2^{0}, 2^{1}$, and $2^{2}$ bits of the DID byte.

### 2.4.5.2.5 Data Bytes

If all receiver parameters are to be controlled, the bytes following the DID Byte will contain data transmitted from receiver to controller or from controller to receiver, as listed in Table 2-8. If all receiver parameters are to be monitored, Byte 3 will be the address return byte followed by data, Bytes 4 through 10 . Transmission of less than complete parameter data requires that the register associated with a particular byte of data be addressed in the DID byte. The selected data will then be transmitted as Byte 3 or 4. Paragraphs 2.4.5.2.7 to 2.4.5.2 .9 contain examples of data formats for full and partial status transmissions, in monitor and command modes.

### 2.4.5.2.6 Tier 2 Access Byte

A second tier containing four pages of eight bytes each is available and is accessed by the byte shown in Table 2-9. The three most significant bits of the byte are the DID code (111), the $2^{4}$ and $2^{3}$ bits are a page code, and the remaining three bits are the address code for the second tier.

Tier 2 contains four pages of eight bytes each. At this time only 2 bytes of tier 2 information have been assigned. The first contains the $10^{0}$ digit (in BCD form) of tuned frequency if the WJ-8718A HF Receiver is equipped with the WJ-8718A/232M, WJ-8718A/MFP, and WJ-8718A/1 Hz Options. The second contains the COR Threshold and the COR Flag if the receiver is equipped with the WJ-8718A/CORM option. The remaining bytes in Tier 2 are available for future expansion. Tables 2-10a and $2-10 \mathrm{~b}$ are the data format for Byte 1 and Byte 2 on Page 1 of Tier 2. This byte must be preceded by a receiver address byte, a Tier 2 access byte, and a DID byte containing the register address 000 or 001. See paragraph 2.4.5.2.9 for an example of data format requiring access to Tier 2.

Table 2-8. Register Parameter Data


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TABLE 2-9, 10a, 10b

Table 2-9. Tier 2 Access Byte

| $2^{7}$ | $2^{6}$ | $2^{5}$ | $2^{4} \quad 2^{3}$ | 2 | $2^{1}$ | $2^{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | 1 | 1 |  | 1 | 1 | 1 |
|  | $\begin{gathered} \text { DID } \\ \text { CODE } \end{gathered}$ |  | $\begin{aligned} & \text { PAGE } \\ & \text { CODE } \end{aligned}$ |  | TIER 2 CODE |  |
|  |  |  | Page $2^{4} \quad 2^{3}$ |  |  |  |
|  |  |  | $\begin{array}{lll}1 & 0 & 0 \\ 2 & 0 & 1 \\ 3 & 1 & 0 \\ 4 & 1 & 1\end{array}$ |  |  |  |

Table 2-10a. Tier 2, Page 1, Byte 1

| $2^{7}$ | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $2^{3}$ | $2^{2}$ | $2^{1}$ | $2^{0}$ |  |  |  |  |
| $10^{0}$ |  |  |  |  |  |  |  |

Table 2-10b. Tier 2, Page 1, Byte 2

| $2^{7}$ | $2^{6}$ | $2^{5}$ | $2^{4}$ | $2^{3}$ | $2^{2}$ | $2^{1}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | $2^{0}$| COR X |
| :--- |
| FLAG |

### 2.4.5.2.7 Monitor Format

Table 2-11 contains examples of data format for full status monitor operations of a WJ-8718A HF Receiver, equipped with the WJ-8718A/232M and MFP Options. Table 2-12 contains examples of data format for partial status monitor operations. Operating parameters of the receiver to be monitored (at address 15) are given in Table 2-13.

### 2.4.5.2.8 Command Format

In the example in Table 2-14, the receiver at address 4 is commanded to establish the listed parameters. Table 2-15 lists the data format for full status command and Table 2-16 is the data for commanding only the BFO frequency.

Table 2-11. Full Status Monitor

| Byte Number | Binary Code |  |
| :---: | :---: | :---: |
| $\begin{array}{r} 1 \\ 2 \\ 3 \\ 4 \\ 5 \\ 6 \\ 7 \\ 8 \\ 9 \\ 10 \end{array}$ | $\left.\begin{array}{lllllllll} 1 & 1 & 0 & 0 & 1 & 1 & 1 & 1 \\ 1 & 1 & 1 & 0 & 0 & x & x & x \end{array}\right\}$ | Controller to Receiver <br> Receiver to Controller |

Table 2-12. Bandwidth/Gain Mode/Detection Mode Monitor

| Byte Number | Binary Code |  |
| :---: | :---: | :---: |
| 1 | 11001111 | Controller to Receiver |
| 2 | 11101100 |  |
| 3 | $\left.\begin{array}{lllllllll} 1 & 1 & 0 & 0 & 1 & 1 & 1 & 1 \\ 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 \end{array}\right\}$ | Receiver to Controller |

Table 2-13. Address 15 Receiver Parameters

| Tuned Frequency | 12.34567 MHz |
| :--- | :--- |
| BFO Frequency | -3.0 kHz |
| Operating Mode | Local |
| Bandwidth | 3.2 kHz |
| Gain Mode | Fast AGC |
| Detection Mode | AM |
| Signal Strength | Maximum |

Table 2-14. Command Parameters to Address 4 Receiver

| Tuned Frequency | 23.45678 MHz |
| :--- | :--- |
| BFO Frequency | +6.0 kHz |
| Operating Mode | Remote |
| Bandwidth | 16 kHz |
| Gain Mode | Manual |
| Detection Mode | CW |
| RF Gain | Maximum |

Table 2-15. Full Status Command

| Byte Number | Binary Code |
| :---: | :---: |
| 1 | 11000100 |
| 2 | $11110 \times \mathrm{x}$ |
| 3 | $\mathrm{xxxxx110}$ |
|  | 00110100 |
| 5 | $01010110$ |
| 6 | $011110000$ |
| 7 | $\begin{array}{llllllll} 0 & 0 & 0 & 0 & 1 & 0 & 1 & 0 \\ 0 & 1 & 1 & 0 & 0 & 0 & 0 & 0 \end{array}$ |
| 9 | $0 \times 000000$ |

Table 2-16. Command BFO Frequency

| Byte Number | Binary Code |  |
| :---: | :---: | :---: |
| 1 | 11000100 | Controller to Receiver |
| 2 | 11111000 |  |
| 3 | x xxxx 110 |  |
| 1 | 11000100 |  |
| 2 | 11111101 |  |
| 3 | 01100000 |  |

### 2.4.5.2.9 Access Tier 2 Format

In this example, shown in Table 2-17, it is assumed that the receiver is equipped with the 1 Hz Tuning Option (in addition to the MFP and 232 M Options) and the controller desires to monitor the status of the $10^{0}$ digit of tuned frequency. The receiver is at address 20 and the $10^{0}$ digit is 6 .

Table 2-17. Tier 2 Access Format

| Byte Number | Binary Code |  |
| :---: | :---: | :---: |
| 1 | $11010100$ | Controller to Receiver |
| 2 | $\begin{array}{lllllllll} 1 & 1 & 1 & 0 & 0 & 1 & 1 & 1 \\ 1 & 1 & 1 & 0 & 1 & 0 & 0 & 0 \end{array}$ |  |
| $\begin{aligned} & 3 \\ & 4 \\ & 5 \end{aligned}$ | $\begin{array}{llllllll} 1 & 1 & 0 & 1 & 0 & 1 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 0 & 0 & 0 \\ 0 & 1 & 1 & 0 & x & x & x & x \end{array}$ | Receiver to Controller |

### 2.4.5.3 488M Operation

The IEEE Standard 488-1975 Interface Bus is organized into three sets of signal lines: eight bidirectional data lines, three handshake lines, and five management control lines. The interface point between the bus and the receiver under control is a 24 -pin Amphenol receptacle mounted in the rear panel of the receiver at the REMOTE INPUT. Table 2-18 lists the 488 connector pin numbers with the respective interface signals and functions. Figure 2-6 illustrates the bus structure.

Table 2-18. 488 Connector Pin Assignments

| Pin Number | Signal | Description |
| :---: | :--- | :--- |
| $1-4$ | DIO1-4 | Data Input/Output Lines |
| $13-16$ | DIO5-8 | Data Input/Output Lines |
| 5 | EOI | End or Identify |
| 6 | DAV | Data Valid |
| 7 | NRFD | Not Ready For Data |
| 8 | NDAC | Not Data Aceepted |
| 9 | IFC | Interface Clear |
| 10 | SRQ | Service Request |
| 11 | ATN | Attention |
| 12 | SHIELD | Shield |
| 17 | REN | Remote Enable |
| $18-23$ | GND | Ground |
| 24 | GND, LOGIC | Logic Ground |

### 2.4.5.3.1 Data Lines

A set of eight data input/output lines, DIO1 through DIO8, carries all address and parameter data from the receiver rear panel REMOTE INPUT through 488-A3J1 to noninverting transceivers U2 and U3. A bit-parallel, byte-serial, negative logic form is used for the bidirectional transfer of data, received in one of two data modes dependent on the level of the ATN management line,

When the ATN line is low, data received is read as receiver address data which place the receiver in the listen, or talk, state. An unlisten data word can also be sent by the controller that unaddresses all previously addressed devices, allowing the devices to accept data when the DAV line is active (low) and the ATN line is high. Equipment address codes (Table 2-3) can be converted to standard ASCII (American National Standard Code for Information Interchange) characters. The address code is an 8-bit binary-coded word reflecting the levels established by the 488-A3S1 settings (paragraph 2.4.3.2) in the five least significant bits. The format of data transfers on the DIO lines is discussed in greater detail in paragraph 2.4.5.4.


Figure 2-6. 488 Bus Structure and MFP Interface Block Diagram

## Courtesy of http://BlackRadios.terryo.org

### 2.4.5.3.2 Handshake Lines

A three-wire handshake sequence facilitates the asynchronous, bidirectional transfer of each byte of data for up to 14 receivers and one controller. The data transfer rate adjusts to the slowest active device on the bus, allowing units with different input/output speeds to be interconnected. The handshake lines are DAV (Data Valid), NRFD (Not Ready For Data), and NDAC (Not Data Accepted). When reading the following description of each signal, refer to Figure 2-7, Handshake Timing Diagram.

DAV - The level on the DAV line is established by the transmitting device to indicate the availability and validity of information on the DIO data lines. When DAV is high, the data is considered invalid; when the signal level is low, the data is considered valid.

NRFD - The logic level of the NRFD signal indicates the condition of readiness of device(s) connected to the bus. When one or more devices are not ready for data, the NRFD line is logic low (true). However, when all devices on the bus are ready to accept data, the signal goes high (false: all are ready to accept data).

NDAC - The NDAC handshake signal indicates the condition of a device(s) to accept data. This line is logic low (true: data has not been accepted) until data are accepted by the device(s) on the bus. When the last device has accepted the data, the NDAC signal goes high.


Figure 2-7. Handshake Timing Diagram

# Courtesy of http://BlackRadios.terryo.org 

### 2.4.5.3.3 Management Control Lines

The management control signals manage an orderly flow of information across the bus. Three signals, ATN (Attention), REN (Remote Enable), and IFC (Interface Clear) are generated by the controller. The SRQ (Service Request) signal is generated by the device under control, and the EOI signal is generated by the transmitting device (controller or receiver).

The ATN line indicates how data on the DIO lines is to be interpreted. When ATN is low, data is read as receiver address data and, when high, as parameter data. The REN and IFC signals (active low) allow the controller to set and reset the interface system.

The SRQ signal indicates that the device under control requires attention. The EOI signal indicates the end of a multiple byte transfer sequence.

### 2.4.5.4 Parameter Data

The full operating status of the receiver is contained in 57 bits of data, stored in eight 8 -bit registers. The data associated with each register is listed in Table 2-15. Data transmission follows the required handshake signals. In the command mode, a register must be addressed before transmission of its data word. In monitor mode, the addressed receiver transmits an address header, followed by all bytes of parameter data, in the order listed in Table 2-19. The command and monitor formats are discussed in detail in paragraphs 2.4.5.4.1 and 2.4.5.4.2.

### 2.4.5.4.1 Command Data Format

In the command mode (receiver is a listener), three types of data words are transmitted by the controller: a receiver address word, a register address word, and a parameter data word corresponding to the register addressed (Table 2-19). To clarify the command data format, the following paragraph contains an example of programming technique using the Hewlett-Packard 9825A Calculator as the remote control device.

### 2.4.5.4.1.1 Remote Command Programming

The Hewlett-Packard 9825A is a typical remote control device compatible with the 488-1975 Interface Bus. The calculator converts decimal keyboard entries to a binary code which is applied to the receiver under control. Table 2-20 lists the binary/decimal conversions for all receiver parameter data.

The command program is entered on the calculator keyboard as a string of decimal numbers, separated by commas. For example, assume that a receiver at address 06 is to be commanded to the following parameters:

> Tuned frequency -12.34567 MHz
> Detection Mode - AM
> Bandwidth -1 kHz
> BFO Frequency - Not applicable in AM
> Gain Mode - Manual
> RF Gain - 30

Table 2-19. Register Address Data


The entire program is entered: $\mathrm{wtb}, 706,1,133,2,35,3,69,4,103,5,125,7,30$. Refer to Tables 2-19 and 2-22 to correlate the following discussion of each program entry.
"wtb" instructs the calculator to write binary entries for byte-serial transmission on the bus.
"706" contains two addresses: 7xx is the calculator's IEEE interface module address and the xx bits are the address code for the equipment on the bus with the calculator. Receiver address is a binary number in the 0 through 15 range (Table 2-3); in this example, address 06 has been assumed.
" 1 " is a register address. Register 1 (001) contains the $\mathrm{R} / \mathrm{L},+/-\mathrm{BFO}$, and gain mode data, plus the $2^{0}$ and $2^{1}$ bits of the MSD of the tuned frequency (BCD).
" 133 " is the decimal equivalent of the binary coded word 10000101, the data word for the addressed register (1). The word is decoded: remote operating mode, manual gain mode, and a tuned frequency MSD of 1 ( $1 \mathrm{X} . \mathrm{XXXXX} \mathrm{MHz}$ ). The BFO sign is irrelevant because the BFO is not applicable in the AM mode.
"2" is the address for Register 2 (010) which contains the $10^{5}$ and $10^{6}$ digits of the tuned frequency in $B C D$ format.
" 35 " is the decimal equivalent of binary 00100011 , the data word for the addressed register (2). The binary word represents BCD 23 , the $10^{5}$ and $10^{6}$ digits of the tuned frequency (X2.3XXXX).
" 3 " is the Register 3 (011) address. Register 3 contains the $10^{3}$ and $10^{4}$ digits of the tuned frequency.
" 69 " is the decimal equivalent of binary word 01000101, the data word for the addressed register (3). The word represents 45 in BCD or a tuned frequency of XX.X45XX.
"4" addresses Register 4 (100), the register for the $10^{1}$ and $10^{2}$ digits of the tuned frequency.
" 103 " is the decimal equivalent of binary 01100111 , the data word for the addressed register (4). In BCD the word is 67 or a tuned frequency of XX.XXX 67 MHz .
" 5 " is the address for Register 5 (101) which contains bandwidth and detection mode data.
" 125 " is the decimal equivalent of binary 01111101, the addressed register (5) data word. The word contains the codes for 1.0 kHz bandwidth and AM detection mode.
"7" addresses Register 7 (111), the RF gain mode register in command mode.
"30" is the RF gain code and needs no conversion because the RF gain is specified as decimal entries from 0 through 255.

In summary, a remote command program consists of a string of decimal entries alternating between register addresses and data. The parameter data, preceded by the register address, can be entered in any order and parameters that are not to be changed are left out.

# Courtesy of http://BlackRadios.terryo.org 

The MSB of the Register 1 data word is always high during remote control, indicating that the receiver is in remote operating mode. The bit is an indicator only, since remote command cannot establish remote operating mode in the receiver. Remote operating mode is established by engaging the front panel REMOTE button.

Table 2-20. Decimal Entries for Registers 0, 2, 3, 4, and 6

| BINARY | DEC | BINARY | DEC |
| ---: | :---: | :---: | :--- |
| $0 \rightarrow 9$ | $=0 \rightarrow 9$ | $30 \rightarrow 39$ | $=80 \rightarrow 89$ |
| $10 \rightarrow 19$ | $=16 \rightarrow 25$ | $60 \rightarrow 69$ | $=96 \rightarrow 105$ |
| $20 \rightarrow 29$ | $=32 \rightarrow 41$ | $70 \rightarrow 79$ | $=112 \rightarrow 121$ |
| $30 \rightarrow 39$ | $=48 \rightarrow 57$ | $80 \rightarrow 89$ | $=128 \rightarrow 137$ |
| $40 \rightarrow 49$ | $=64 \rightarrow 73$ | $90 \rightarrow 99$ | $=144 \rightarrow 153$ |

Table 2-21. Decimal Entries for Register 1

| $\begin{aligned} & \text { REM/LCL } \\ & +/-\mathrm{BFO} \end{aligned}$ | GAIN MODE | FREQ 10 | $\begin{gathered} \text { DEC } \\ \text { EQUIV. } \end{gathered}$ | $\begin{aligned} & \text { REM/LCL } \\ & +/-\mathrm{BFO} \end{aligned}$ | $\begin{aligned} & \text { GAIN } \\ & \text { MODE } \end{aligned}$ | FREQ $10^{7}$ | $\begin{gathered} \text { DEC } \\ \text { EQUIV } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | FAST | $0 \mathrm{X} . \mathrm{MHz}$ <br> 1X.MHz <br> $2 \mathrm{X} . \mathrm{MHz}$ | 128 129 130 | -BEO | FAST | 0X.MHz 1X.MHz 2X.MHz | $\begin{aligned} & 144 \\ & 145 \\ & 146 \end{aligned}$ |
|  | MAN | $\begin{aligned} & \text { 0X.MHz } \\ & \text { 1X.MHz } \\ & \text { 2X.MHz } \end{aligned}$ | $\begin{aligned} & 132 \\ & 133 \\ & 134 \end{aligned}$ |  | MAN | 0X.MHz 1X.MHz 2X.MHz | 148 149 150 |
|  | SLOW | $\begin{aligned} & \text { 0X.MHz } \\ & \text { 1X.MHz } \\ & \text { 2X.MHz } \end{aligned}$ | $\begin{aligned} & 136 \\ & 137 \\ & 138 \end{aligned}$ |  | SLOW | $\begin{aligned} & \text { 0X.MHz } \\ & \text { 1X.MHz } \\ & \text { 2X.MHz } \end{aligned}$ | $\begin{aligned} & 152 \\ & 153 \\ & 154 \end{aligned}$ |

Table 2-22. Decimal Entries for Register 5

| DETECTION MODES, DEC ENTRIES |  |  |  |  |  |  |
| :---: | ---: | ---: | ---: | ---: | ---: | ---: |
| BW, kHz | LSB | USB | ISB | CW | FM | AM |
| 16 | 0 | 1 | 2 | 11 | 28 | 29 |
| 6 | 32 | 33 | 34 | 43 | 60 | 61 |
| 3.2 | 64 | 65 | 66 | 75 | 92 | 93 |
| 1 | 96 | 97 | 98 | 107 | 124 | 125 |
| .3 | 128 | 129 | 130 | 139 | 156 | 157 |

# Courtesy of http://BlackRadios.terryo.org 

### 2.4.5.4.2 Monitor Data Format

A monitor program begins with the transmission of a receiver address word which activates the I/O circuits in the addressed receiver, causing an interrupt signal to be generated to the microprocessor in the receiver control block. The microprocessor branches to an area of memory containing the current receiver parameter data. Receiver parameters are stored in sequential memory locations in the order listed in Table 2-19. The processor begins to transmit data at the location where a previous data transmission ended. The controller receives the transmitted data in binary-coded form and has no way of knowing which data is represented by the binary-coded transmissions.

To synchronize the transmission of data, an address header precedes the parameter data words in memory. The address header is an 8 -bit word containing the 5 -bit receiver address in the lower-order bits. The $2^{6}$ and $2^{7}$ bits are high to uniquely identify the word as an address header ( $2^{5}$ bit is irrelevant). After the address header has been transmitted and identified, the eight data words that follow will be the parameter data listed in Table 2-19.

The programmer uses a comparison technique to identify or "find" the address header. The first two bits of the binary-coded word are high (11XXXXXX); therefore, the decimal equivalent of the address header word is always greater than 192. The parameter data binary codes are such that the 2 MSDs are never high at the same time; thus, the decimal equivalent of a parameter data word will always be less than 192. By successively comparing the decimal equivalent of each transmitted word to 192 , the address header can be isolated and the data words identified.

### 2.4.5.4.2.1 Remote Monitor Programming

Table 2-23 is a monitor program for the Hewlett-Packard 9825A Calculator. The first four program steps ( 0 through 3) establish a string of 15 spaces (A \$ [15]) for placement of parameter data, formatting the display readout, and loading zeros into locations I and A.

Program steps 4,5, and 6 establish two loops, or sets of program instructions, for the identification of the address header and parameter data words and their storage in consecutive locations. Step 4 instructs the calculator to read the binary-coded data (rdb) received on the 488 interface ( 7 XX ) from the equipment at address 06 (X06), and to place the data word in register rI or r0 $(\rightarrow \mathrm{rI})$. Step 5 examines registers $A$ and rI, and, if $A=0$ and rI is less than 192, repeats step 4 . This procedure (loop) will repeat until the decimal equivalent of the binary-coded data word stored in rI, is greater than 192 (the only data word greater than 192 is the address header word). At this point, the program jumps out of the loop to step 6.

In step $6, r I$ is incremented and $A$ is loaded with a 1 ; if $r I$ is less than 9 ( $r I$ is now 1 ), the program returns to step 4. The received data word is read and placed in rI (now in r1, one location away from the address header word in location r0) and the program advances to step 6. Again, rI is incremented $(\mathrm{I}+1=\mathrm{I}$ or $1+1=2=\mathrm{I}$ ) and the loop repeats until $\mathrm{I}=9$, when the program jumps out of the loop to step 7. The contents of r0 are reduced by 192: at this point r0 contains the address and r1 through 8 contain the receiver parameter data words in the order listed in Table 2-19.

Program steps 8 through 22 are a combination of binary AND, shift, and comparison operations which isolate the information in the parameter data words. For example, in step 14, the contents of r6 (in Table 2-19, the register 5 data) are ANDed with 31 (00011111),

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extracting the lower 5 bits of detection mode data, which are placed in Q. In step 15 , the r 6 contents are shifted right five places; the three LSBs of the data word now contain the bandwidth code and all other bits are zeros. This data word is placed in R.

Steps 23 through 42 of the program place the data (derived in steps 8 through 22) in the string reserved in step 0 . For example, in step 24, if $Q=29$, the selected detection mode is AM (Table 2-22); A is placed in the 3rd place and $M$ is placed in the 4th place in the string ("AM" $\rightarrow$ \$ 3, 4 ).

If the receiver under control is equipped with the 10 Hz BFO Option, Register 0 ( r 1 ) will contain the BCD $10^{1}$ frequency digit. In program step 43, the contents of r 1 are compared to 9 . If the register is used, the decimal equivalent of the binary-coded data word will be less than 9 , and the contents of r1 will remain intact. If the register is unused, all bits will be high, or greater than 9 , and a 0 will be loaded into r 1 (if $\mathrm{r} 1>9 ; 0>\mathrm{r}$ ).

In the remaining steps of the program, the calculator is instructed to display the receiver address and the parameters in the following order: $R / L$ status, detection mode, bandwidth, gain mode, BFO sign, three digits of BFO frequency, signal strength, and seven digits of RF tuned frequency.

Table 2-23. Monitor Program

| 0 : | dim A\$ [15] ; fxd 0 |
| :---: | :---: |
| 1: | prt "READ TEST" |
| 2: | fmt f1.0, "•", 2f1.0, " ", f3.0, " ", 2f1.0, "•", 5 f1.0 |
| 3 : | $0 \rightarrow \mathrm{I} ; 0 \rightarrow \mathrm{~A}$ |
| 4: | "loop":rdb (706) $\rightarrow$ rI |
| 5: | if $\mathrm{A}=0$ and rI<192;gto "loop" |
| 6: | $\mathrm{I}+1 \rightarrow \mathrm{I}$; $1 \rightarrow$ A, if $\mathrm{I}<9$;gto "loop" |
| 7: | r0-192 $\rightarrow$ r0 |
| 8: | band ( $\mathrm{r} 3,15$ ) $\rightarrow \mathrm{J}$ |
| 9: | shf ( $\mathrm{r} 3,4$ ) $\rightarrow$ I |
| 10: | band ( $\mathrm{r} 4,15$ ) $\rightarrow \mathrm{L}$ |
| 11: | shf ( $\mathrm{r} 4,4$ ) $\rightarrow$ K |
| 12: | band ( $\mathrm{r} 5,15$ ) $\rightarrow \mathrm{P}$ |
| 13: | shf ( $\mathrm{r} 5,4$ ) $\rightarrow$ M |
| 14: | band ( $\mathrm{r} 6,31$ ) $\rightarrow$ Q |
| 15: | shf ( $\mathrm{r} 6,5$ ) $\rightarrow$ R |
| 16: | band ( $\mathrm{r} 7,15$ ) $\rightarrow 0$ |
| 17: | $\operatorname{shf}(\mathrm{r} 7,4) \rightarrow \mathrm{N}$ |
| 18: | band ( $\mathrm{r} 2,3$ ) $\rightarrow \mathrm{H}$ |
| 19: | if $\mathrm{r} 2>=128 ; 1 \rightarrow \mathrm{~V}$;gto +2 |
| 20: | $0 \rightarrow$ V |
| 21: | band ( $\mathrm{r} 2,16$ ) $\rightarrow \mathrm{U}$ |
| 22: | band ( $\mathrm{r} 2,12$ ) $\rightarrow$ T |
| 23: | " " $\rightarrow$ A\$ [1,15] |
| 24: | if $\mathrm{Q}=29 ; \mathrm{AM}^{\prime \prime} \rightarrow \mathrm{A}$ \$ [3,4] |
| 25: | if $\mathrm{Q}=28 ; \mathrm{FMM} \rightarrow \rightarrow \mathrm{A} \$[3,4]$ |
| 26: |  |
| 27: | if $\mathrm{Q}=2 ;$ "ISB" $\rightarrow$ A \$ 3,5$]$ |
| 28: | if $\mathrm{Q}=1$; "USB" $\rightarrow$ A\$ $[3,5]$ |
| 29: | if $\mathrm{Q}=0 ; \mathrm{LSSB}{ }^{\text {a }}$ A A [3,5] |
| 30: | if $\mathrm{R}=0 ; " 16 \mathrm{kHz}{ }^{\prime \prime} \rightarrow \mathrm{A} \$[7,11]$ |
| 31: | if $\mathrm{R}=1 ;{ }^{\prime \prime} 6 \mathrm{kHz}{ }^{\prime \prime} \rightarrow \mathrm{A} \$[7,10]$ |
| 32: | if $\mathrm{R}=2 ;{ }^{\prime \prime} 3.2 \mathrm{kHz}$ " $\rightarrow \mathrm{A}$ \$ $[6,11]$ |
| 33: | if $\mathrm{R}=3 ; \mathrm{\prime} 1 \mathrm{kHz}$ " $\rightarrow \mathrm{A}$ \$ $[7,10]$ |
| 34: | if $\mathrm{R}=4 ; \mathrm{F} .3 \mathrm{kHz}{ }^{\prime \prime} \rightarrow \mathrm{A} \$[7,11]$ |
| 35: | if $\mathrm{R}=5$;"opt" $\rightarrow$ A\$ $[8,10]$ |
| 36: | if $\mathrm{T}=0 ; \mathrm{F} \mathrm{F} \rightarrow \rightarrow \mathrm{A}$ \$ $[13,13]$ |
| 37: | if $\mathrm{T}=4 ; \mathrm{M} \mathrm{M}^{\prime \prime} \rightarrow \mathrm{A} \$[13,13]$ |
| 38: | if $\mathrm{T}=8 ; \mathrm{S}$ " $\rightarrow \mathrm{A}$ \$ $[13,13]$ |
| 39: | if $\mathrm{U}=0 ;{ }^{\prime \prime}+\mathrm{H} \rightarrow \mathrm{A}$ \$ $[15,15]$ |
| 40: | if $\mathrm{U}=16 ;{ }^{\prime \prime}-\mathrm{H} \rightarrow \mathrm{A} \$[15,15]$ |
| 41: | if $\mathrm{V}=0 ; " \mathrm{~L} \\| \rightarrow \mathrm{A}$ \$ $[1,1]$ |
| 42: | if $\mathrm{V}=1 ; \mathrm{R}^{\prime \prime} \rightarrow \mathrm{A} \$[1,1]$ |
| 43: | if $\mathrm{r} 1>9 ; 0 \rightarrow \mathrm{r} 1$ |
| 44: | wrt 0, A\$, N, O,r1,r8,H,I,J,K,L,M,P |
| 45: | if $\mathrm{X}=0 ; \mathrm{prt}$ "RECEIVER ADDRESS",r0;spe 2 |
| 46: | $1 \rightarrow \mathrm{X}$;gto 3 |

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## SECTION III

## CIRCUIT DESCRIPTION

## 3.1

## MICROPROCESSOR FRONT PANEL ASSEMBLY CIRCUIT DESCRIPTIONS

Figure $3-1$ is a functional block diagram of the front panel section of the WJ-8718A/MFP Option which includes Encoder Assembly MFP-A2 and Front Panel Switch/Encode Assembly MFP-A1. The front panel assemblies interface all front panel features (except the power switch) with the I/O section of the receiver. Paragraph 2.4.4 of this manual contains a functional description of all front panel controls, indicators, displays, and connections. Refer to the schematic diagrams, Figures 6-1 through 6-8 when reading the following circuit descriptions.

### 3.1.1 TYPE 791202-5 ENCODER ASSEMBLY (MFP-A2)

### 3.1.1.1 Optical Encoder (U1)

The optical encoder (U1) converts the front panel tuning wheel rotation into electrical inputs to the Front Panel Encode Board (MFP-A1A1), described in paragraph 3.1.3. When the tuning wheel is rotated, both of the encoder output lines, CLK and DIR, swing repeatedly between 0 and +5 Vdc . Due to the internal mechanics of the encoder, the transitions of the two outputs are square wave trains staggered in time in relationship to each other, as shown in Figure 3-2. When the tuning wheel is rotated clockwise to increase tuned frequency, the square wave on the DIR line will appear to lead that on the CLK line. When the tuning wheel is rotated counterclockwise, the DIR square wave lags the CLK square wave. The two encoder inputs go through approximately 120 cycles per revolution of the input shaft, causing a tuning step for roughly each $3^{0}$ of shaft rotation.

### 3.1.2 TYPE 794310-X FRONT PANEL SWITCH/ENCODE ASSEMBLY (MFP-A1)

The Front Panel Switch/Encode Assembly (MFP-A1) is located directly behind the front panel (Figure 6-2). The module contains two subassemblies, interconnecting track, and the amplifiers for the front panel phones jack audios. The microprocessor referenced throughout the following circuit descriptions is located on the IF Interface Board, MFP-A3 (Figure 6-5).

### 3.1.2.1 Headphone Audio Amplifiers (MFP-A1, U1A, U1B)

Audio amplifiers U1A and U1B are low-power operational amplifiers with differential inputs. The non-inverting input to each amplifier is a de voltage varied by the front panel PHONE LEVEL potentiometer (R7). In all detection modes except ISB, the non-inverting amplifier inputs, AUDIO 1 and AUDIO 2, are the same audio signal. In ISB detection mode, the audio to U1A is USB information (AUDIO 1) and audio to U1B is LSB information (AUDIO 2). U1A outputs to the tip of the front panel stereo phones jack and U1B outputs to the ring.

Courtesy of http://BlackRadios.terryo.org


Figure 3-1. Front Panel Section, Block Diagram


Figure 3-2. Optical Encoder Timing Diagram

The amplifiers will drive standard 600 ohm stereo or mono headphones. If mono headphones are used, LSB in the ISB mode will not be monitored.

### 3.1.2.2 Type 794309-X Front Panel Switchboard (MFP-A1A2)

Figure 6-4 is the schematic diagram for the Front Panel Switchboard and contains the switch matrixes and all front panel LED displays and indicators. The switches, and the LED indicators and displays associated with the switches, are organized into two groups. Each group has a dedicated, programmable interface circuit on the MFP-A1A1 board, described in paragraph 3.1.3.

### 3.1.2.2.1 Group 1 Switch Matrix (S1 Through S20)

Keypad buttons S1 through S16, tuning resolution buttons S18 through S20, and LCK key S17 form the $4 \times 5$ Group 1 switch matrix. The KC0 through KC4 column scan lines and the KR0 through KR3 row output lines are connected to the Group 1 programmable keyboard/display interface circuit (MFP-A1A1U1). A closed switch connects the column scan line with the row line; the interface senses this closed switch, codes the position, and stores the code in a memory area which the microprocessor monitors during the execution of its program. A detailed description of MFP-A1A1U1 can be found in paragraphs 3.1.3.3 and 3.1.3.4.

### 3.1.2.2.2 Group 2 Switch Matrix (S21 Through S56)

The $8 \times 8$ Group 2 switch matrix ( S 21 through S 56 ) contains all front panel switches except the power switch and the switches in Group 1. The BR0 through BR7 row output lines
and the BC0 through BC7 column scan lines are connected to the Group 2 programmable keyboard/display interface (MFP-A1A1U2), described in paragraphs 3.1.3.3 and 3.1.3.5.

The interface senses a closed switch in the Group 2 matrix, codes the position and stores the code in a microprocessor accessible memory location.

### 3.1.2.2.3 LED Indicators

All front panel switches, except the power switch and keypad push buttons, have an associated LED indicator which glows when the switch is engaged. In addition, three LED FAULT lights glow red to indicate various invalid operating conditions (see paragraph 2.4.4.10).

The Group 1 LEDs are DS19 through DS23 and DS61 through DS63. The LEDs are forward biased by drivers associated with the Group 1 interface, MFP-A1A1U1. The drivers are discussed in paragraphs 3.1.3.5 and 3.1.3.7.

The Group 2 LEDs, DS24 through DS53, are forward biased by drivers associated with the Group 2 interface, MFP-A1A1U2. The Group 2 drivers are discussed in paragraphs 3.1.3.12 and 3.1.3.13.

### 3.1.2.2.4 LED Displays

DS1 through DS4, DS7 through DS13, DS15, and DS16 are 7-segment, commoncathode LED displays. DS1 through DS4 form the front panel BFO OFS/THRS LVL display; DS7 through DS13 form the FREQUENCY/MHz display (with DS60 providing the decimal point); DS15 and DS16 are the MEM ADRS display. The displays are functionally described in paragraphs 2.4.4.4.3, 2.4.4.5.4 and 2.4.4.7.1.

Each 7-segment array consists of 7 light-emitting diodes which are optically magnified to form 7 individual segments. The LEDs are arranged so that a numerical digit from 0 through 9 can be formed by illumination of selected segments. The cathodes of all LEDs in an array are tied to a common cathode input (cc) and the anode of each LED is connected to an a-through-g input. An LED is forward biased when the cathode becomes sufficiently more negative than the anode.

The LED biasing signals come from driver circuits associated with the Group 1 and Group 2 keyboard/display interface circuits, MFP-A1A1U1 and U2. The BFO OFS/THRS LVL and FREQUENCY/MHz displays are associated with Group 1 and the MEM ADRS display is associated with Group 2.

The negative sign in the BFO display is formed by illumination of only the g-segment (the center bar in the number "8") of DS1. Absence of the negative sign implies a positive BFO. The decimal point in the BFO frequency display is provided by the decimal segment of DS2, forward biased by the FDP level from the Front Panel Encode Board.

### 3.1.2.2.5 Voltage Regulator (U1)

Voltage regulator U1 provides a variable voltage of approximately +7 Vdc for the display drivers on the Front Panel Encode Board. The intensity of the front panel displays can be adjusted by varying potentiometer R2.

### 3.1.3 TYPE 796056-1 FRONT PANEL ENCODE (MFP-A1A1)

Figure 6-3 is the schematic diagram for the Front Panel Encode Board (MFP-A1A1); Figure 3-1, Front Panel Section Block Diagram, shows the interdependent relationship of all front panel boards.

The Front Panel Encode Board is powered by the following de voltages: +5 Vdc applied across the capacitive decoupling network formed by C1 through C10 and C14 through C16; -5 Vde derived from -15 V at J1 pin 35 and the VR1, C17, and R19 voltage regulating network; variable voltage derived from 10 V unregulated by U1 (paragraph 3.1.2.2.5) applied to the 7-segment drivers at A60 and B25.

### 3.1.3.1 Address Latch (U18)

Integrated circuit U18 is an octal D-type flip-flop. The CLR input is tied high, allowing the levels on the $D$ inputs to transfer to the $Q$ outputs on the positive-going edge of the clock (CLK) input. During the time the clock input remains either high or low, changes in D levels have no effect on the outputs. The clock signal to U18 is the inverted (by U13A) microprocessor ALE control signal. ALE transitions low when the microprocessor data bus contains address data, latching the AD0 through AD7 address bits to the U18 outputs. The AD0 bit is the A0 input to the programmable keyboard/display interfaces (U1 and U2) and is described in paragraphs 3.1.2.2.5 and 3.1.3.9.1. The remaining lower-order address bits are applied to address decoder U19, described in the following paragraph.

### 3.1.3.2 Address Decoder (U19)

U19 decodes the address latched by U18 to provide enabling or clock inputs to all addressable devices on the MFP-A1A1 board. Table 3-1 is the U19 truth table. The Y0 through Y3 outputs are normally high and go low only when the enabling and select inputs are as listed in the table.

Table 3-1. U19 Truth Table

|  |  | Enable | Inputs | Select |  |  |  | Outputs |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | G1 | G2A | G2B | C | B | A | Y0 | Y1 | Y2 | Y3 |  |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |  |
| 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 1 | 1 | 1 |  |
| 1 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 1 | 1 |  |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 0 | 1 |  |
| 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 | 1 | 0 |  |

# Courtesy of http://BlackRadios.terryo.org 

TABLE 3-2

### 3.1.3.2.1 U19 Enable

As shown in Table 3-1, U19 is enabled by a high logic level on G1 and a low logic level on both G2A and G2B. The G2A and G2B inputs are the signals carried on the AD4 and AD5 bits of the microprocessor data bus (latched by U18); both address bits must be low to enable U19. The G1 enable input is the output of AND gate U15B. The only time the U15B output is high is when both inputs, the AD6 and A12 address bits, are high. Table 3-2 lists the binary equivalent of the four hexadecimal address digits of all addressable devices on the Front Panel Encode Board, and shows that the U19 enabling conditions exist when any one of the hexadecimal addresses 1040, 1042 through 1046, or 1048 has been on the data bus and latched by U18.

Table 3-2. Front Panel Encode Address Data

| Bus Line |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| A |  |  |  |  |  |  |  | AD |  |  |  |  |  |  |  |
| 15 | 14 | 13 | 12 | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 0 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 1 | 1 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 1 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 1 | 0 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 0 | 0 |

### 3.1.3.2.2 U19 Select

U19 is a 3 -to-8 line decoder. When U19 is enabled, the levels on the $\mathrm{A}, \mathrm{B}$, and C select inputs determine which Y output goes low. The A, B, and C inputs carry the levels on the $\mathrm{AD} 1, \mathrm{AD} 2$, and AD 3 bus lines.

Refer to Table 3-1 and note that the conditions required for Y0 to go low are: $\mathrm{C}=0, \mathrm{~B}=0$, and $\mathrm{A}=0$. Table 3-2 shows that the only time that AD3 (C), AD2 (B), and AD1 (A) are low at the same time is when the hexadecimal address is 1040 . This is the address for readonly device U17, described in paragraph 3.1.3.16. Y4 goes low on a $\overline{\mathrm{RD}}$ to address 1048 , used by the microprocessor to reset U17.

Likewise, the U 19 truth table shows that Y 1 is low when $\mathrm{C}=0, \mathrm{~B}=0$, and $\mathrm{A}=1$. Table 3-2 shows that AD3 $(C)=0, A D 2(B)=0$, and AD1 $(A)=1$ when the address is either 1042 or 1043. The U19Y1 signal is the enabling pulse to U1, addressed by either 1042 or 1043 and described in paragraph 3.1.3.4.

The U19 truth table shows that Y 2 is low when $\mathrm{C}=0, \mathrm{~B}=1$, and $\mathrm{A}=0$. Table 3-2 shows that $\mathrm{AD} 3(\mathrm{C})=0, \mathrm{AD} 2(\mathrm{~B})=0$, and $\mathrm{AD} 1(\mathrm{~A})=0$ when the address is either 1044 or 1045. These are the addresses for U2, enabled by Y2 and described in paragraph 3.1.3.9.

Refer to Table 3-1 and note that $Y 3$ goes low when $C=0, B=1$, and $A=1$. These conditions exist when AD3 $(C)=0, A D 2(B)=1$, and $A D 1(A)=1$ or, from Table 3-2, when the hexadecimal address is 1046. Y3 provides the enabling pulse to the address 1046 write-only device, U20, described in paragraph 3.1.3.14.

### 3.1.3.3 Intel 8279 Programmable Keyboard/Display Interface (U1, U2)

The 8279 interface chip is designed for use with the Intel 8085A Microprocessor, located in the IF Interface (MFP-A3). The microprocessor is programmed to command the operating characteristics of the interface chip, enabling the interface to provide LED indicator and display drivers with information derived from front panel scans or data stream from a remote control device (if an optional remote interface is used). The interface automatically scans the front panel keyboard and refreshes the display, relieving the microprocessor of these tasks.

Table 3-3 lists the functions of the input and output lines of the 8279 interface which are used in this MFP application. Figure 3-3 is a block diagram of the interface chip showing key functional areas which are discussed in paragraphs 3.1.3.3.1 through 3.1.3.3.4. The functional and operational characteristics of the Group 1 interface (U1) and its associated external decode/driver circuits are discussed in paragraphs 3.1.3.3 through 3.1.3.8; the Group 2 interface (U2) and its associated external circuits are discussed in paragraphs 3.1.3.3 and 3.1.3.9 through 3.1.3.1.3.

Table 3-3. Pin Functions of Intel 8279 Interface

| Designation | Function |
| :---: | :---: |
| DB0 - DB7 | Bidirectional data bus lines which transfer data between the interface and the microprocessor bus. |
| CLK | Clock from external RC network which is used to generate internal timing. |
| RESET | Raised high by the microprocessor to reset the internal circuits of the interface. |
| $\overline{\mathrm{CS}}$ | A low level enables the receive or transmit functions of the interface. |
| A0 | Defines the functions of DB0 - DB7 signals: AO low, data; A0 high, command/status. |
| $\overline{\mathrm{RD}}$ | A low level enables data to be transmitted to the microprocessor bus via DB0-DB7. |
| $\overline{\mathrm{WR}}$ | A low level enables data to be received from the microprocessor bus via DB0-DB7. |
| SL0-SL3 | Used to scan switch matrix columns. |
| RL0 - RL7 | Return lines from switch matrix rows; lines connect to scan lines through closed switches. |
| OUT B0-B3 | Output lines synchronized to scan lines to provide data to the external display segment/indicator LED anode drivers. |
| $\overline{\mathrm{BD}}$ | Blanks the display during digit switehing or by command. |

### 3.1.3.3.1 I/O Control and Data Buffers

The bidirectional data buffers connect the internal bus to the DB0 through DB7 bus lines as directed by the levels of the $\overline{\mathrm{RD}}, \overline{\mathrm{WR}}, \overline{\mathrm{CS}}$, and A 0 signals to the I/O control circuits. The microprocessor enables the flow of data with the $\overline{\mathrm{CS}}$ signal, defines the character of input/output data with the A0 signal, and establishes the direction of data flow with the $\overline{\mathrm{RD}}$ and $\overline{\mathrm{WR}}$ signals.


Figure 3-3. Intel 8279 Interface Block Diagram

A high level on the A0 lines indicates that the information transferred through the data buffers is command or status, and a low level indicates data. When the interface is not selected ( $\mathrm{CS}=1$ ) the buffers are in a high impedance state. The data buffers input during the time $\overline{\mathrm{WR}}$ and $\overline{\mathrm{CS}}$ are low, and output when $\overline{\mathrm{RD}}$ and $\overline{\mathrm{CS}}$ are low. Refer to paragraphs 3.1.3.4.1 and 3.1.3.9.1 for information concerning the derivation of the $\overline{C S}, A 0, \overline{R D}$, and $\overline{W R}$ signals.

### 3.1.3.3.2 Control and Timing Sections

The internal control and timing registers store mode and operating conditions programmed by the microprocessor. Conditions are programmed by commands on the data lines when A0 is high and $\overline{W R}$ is low. The command word is latched into the interface when $\overline{W R}$ transitions to high, and the interface circuits decode and execute the command.

The timing circuits receive an external CLK signal generated by the external network of U13E and F, R7, and C11. The time constant of R7 and C11, in conjunction with the feedback and inversion characteristics of U13E and $F$, provides square wave pulses of 1 MHz . Internal counters divide the reference frequency as required, to synchronize internal timing with the programmed microprocessor cycle time.

The RESET input is set high, resetting the interface circuits, when the microprocessor completes its reset procedure following power-up.

The scan counter provides a 4-bit binary count to the SL0 through SL7 output lines. An external decoder drives the keyboard scan lines, and an external demultiplexer switches the $B C D$ code levels on the B0 through B3 pins to provide inputs to LED cathode drivers.

### 3.1.3.3.3 Return Line Circuits

The return line circuits include buffers, keyboard debounce and control circuits, and an $8 \times 8$ FIFO/Sensor RAM. The RL0 through RL7 lines are connected to the column scan lines through closed front panel switches. The return line circuits scan the lines, detect a closed switch, code the switch position, and store the code in the $8 \times 8$ FIFO/Sensor RAM. The microprocessor accesses the RAM locations to read the status of the switches (under program control).

### 3.1.3.3.4 Display Circuits

The display section includes registers which establish the levels on the B0 through B3 output lines. These lines are synchronized with the scan lines to provide a BCD output to an external decoder/driver which drives the front panel LED and display segment anodes. The interface automatically refreshes the displays.

Other display circuits are address registers and a $16 \times 8$ RAM. The microprocessor configures the display characteristies through the use of the display RAM. In this MFP application, the RF frequency display is configured for right or "calculator style" entry. The first numerical front panel keypad entry (for RF frequency, threshold level, dwell time, or memory address) is placed in the right most display character and is shifted left one character
as additional entries are made. The seven-character frequency display is used to display the threshold level, dwell time, and memory address only during numerical entry. When the appropriate terminator key (see paragraph 2.4.4.2.1) is depressed, the number appears in the appropriate display as described in paragraphs 2.4.4.4.3, 2.4.4.5.4, and 2.4.4.7.1. The $\overline{\mathrm{BD}}$ signal is generated by the timing circuits to blank the display during digit switching.

### 3.1.3.4 Group 1 Interface (U1)

The U1 interface scans the front panel switch matrix identified as Group 1, and interfaces the associated LED indicators and displays. Refer to paragraph 3.1.2.2.1 for a discussion of the switch matrix.

### 3.1.3.4.1 U1 Control Inputs ( $\overline{\mathrm{CS}}, \mathrm{A} 0, \overline{\mathrm{WR}}, \overline{\mathrm{RD}})$

The U1 interface is addressed by the microprocessor at hexadecimal 1042 or 1043. The address is latched by U18 when the ALE signal goes low, U19 is enabled, and the $\overline{\mathrm{CS}}$ signal (the Y1 output of U19) goes low to enable U1. The level on the A0 input to U1 is established by the LSB of the address. The microprocessor uses address 1042 ( $\mathrm{LSB}=0$ ) if the information to be transferred on the data bus is to be interpreted (by U1) as data. The address 1043 (LSB $=1$ ) is used if the data bus word is to be interpreted as command or status. The $\overline{W R}$ signal goes low to permit the microprocessor to load the bus with a word to be transmitted to U1; the word is latched into U1 when $\overline{\mathrm{WR}}$ transitions high. The microprocessor sets $\overline{\mathrm{RD}}$ low to request information from U1. The U1 data buffers load the information on the bus when $\overline{R D}$ is low and, when $\overline{\mathrm{RD}}$ transitions high, the information is latched by the microprocessor.

### 3.1.3.4.2 U1 Scan Line Outputs/Return Line Inputs

The internal circuits of U1 provide a binary-coded count on the SL0 through SL3 output lines. Three bits of the binary code ( 0 through 7) are decoded by U11 (described in paragraph 3.1.3.5) to become the column scan lines (KC0 through KC4). Each column scan line is connected to one of the five X planes in the Group 1 switch matrix. Each switch in the matrix is located at a point where the $X$ and $Y$ planes cross and can be identified by the unique position it occupies. The Y planes of the matrix are connected to the return lines (RL0 through RL3) of U1. When a switch is closed, a circuit is completed from the column scan line, through the closed switch, and through the return (row scan) line back to U1. The closed switch position is coded and stored in U1 RAM. During the execution of its program, the microprocessor initiates any further action necessary to incorporate the functions of the closed switch.

The binary count ( 0 through 15) on the SL0 through SL3 lines is applied to multiplexer U4, where one of 16 channels is switched as the count proceeds (refer to paragraph 3.1.3.6). The selected channel output goes high, resulting in (through U5 or U6) a negative cathode voltage to a front panel LED indicator or display array (see paragraphs 3.1.2-2.3 and 3.1.2.2.4). The LED is not forward biased and does not illuminate, unless its anode is sufficiently more positive than the cathode. The anode drive is provided through the B0 through B3 outputs of U1, described in paragraph 3.1.3.4.3.

### 3.1.3.4.3 U1 Output Ports B0 Through B3

The B0 through B3 lines provide a BCD code to display driver U3, described in paragraph 3.1.3.8. The lines are synchronized to the scan lines, allowing U1 to automatically scan the switches and update or refresh the front panel displays.

### 3.1.3.5 Column Driver (U11)

Integrated circuit U11 is a 3 -to-8 line decoder. The G1, G2A, and G2B enable inputs are tied to +5 Vdc and ground to hold the decoder in the enabled state. The levels on the A, B, and C inputs are decoded to provide a low level on one Y output as listed in Table 3-4. Refer to paragraph 3.1.3.4.2 for details of the function of the $Y$ outputs.

### 3.1.3.6 Multiplexer (U4)

U4 is a 4-to-16 line decoder. The SL0 through SL3 lines from U1 count in BCD as the front panel Group 1 switches are scanned. The BCD code is applied across pull-up resistors R11 through R14 to the A, B, C, and D select inputs of U4. Multiplexer U4 is a single digitally controlled 16 -channel switch, arranged so that the 4 -bit $B C D$ input selects the channel associated with the code, as listed in Table 3-5, the U4 Truth Table.

### 3.1.3.7 Cathode Drivers (U5, U6)

The outputs of multiplexer U4 are connected to the D inputs of cathode drivers U5 and U6. The output of the selected U4 channel goes high, and U5 or U6 inverts the high D input to a low level which is applied to the cathode of the selected LED or LED array. The display digits are enabled one at a time and refreshed at a rate undetectable by the human eye.

Table 3-4. U11 Truth Table

|  | Enable | Inpu | Select |  |  | Outputs |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| G1 | G2A | G2B | C | B | A | Y0 | Y1 | Y2 | Y 3 | Y4 | Y5 | Y6 | Y7 |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 1 | 1 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 1 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | 1 | 1 | 1 |
| 1 | 0 | 0 | 1 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 1 |
| 1 | 0 | 0 | 1 | 1 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |

Table 3-5. U4 Truth Table

| A | B | C | D | INH | Selected Channel |
| :---: | :---: | :---: | :---: | :---: | :---: |
| X | X | X | X | 1 | None |
| 0 | 0 | 0 | 0 | 0 | 0 |
| 1 | 0 | 0 | 0 | 0 | 1 |
| 0 | 1 | 0 | 0 | 0 | 2 |
| 1 | 1 | 0 | 0 | 0 | 3 |
| 0 | 0 | 1 | 0 | 0 | 4 |
| 1 | 0 | 1 | 0 | 0 | 5 |
| 0 | 1 | 1 | 0 | 0 | 6 |
| 1 | 1 | 1 | 0 | 0 | 7 |
| 0 | 0 | 0 | 1 | 0 | 8 |
| 1 | 0 | 0 | 1 | 0 | 9 |
| 0 | 1 | 0 | 1 | 0 | 10 |
| 1 | 1 | 0 | 1 | 0 | 11 |
| 0 | 0 | 1 | 1 | 0 | 12 |
| 1 | 0 | 1 | 1 | 0 | 13 |
| 0 | 1 | 1 | 1 | 0 | 14 |
| 1 | 1 | 1 | 1 | 0 | 15 |

### 3.1.3.8 Display Driver (U3)

LED driver U3 decodes the BCD code at the A, B, C, and D inputs to provide a high output level to the segments associated with the selected digit, forward biasing the segment anodes and causing display of the digit. The outputs of U3 also drive the LED indicators associated with Group 1: the tuning resolution, LOCK pushbutton, and fault lights.

### 3.1.3.9 Group 2 Interface (U2)

The U2 interface scans the front panel switch matrix defined as Group 2 and interfaces the associated LED indicators and displays. Refer to paragraph 3.1.2.2.2 for a discussion of the switch matrix.

### 3.1.3.9.1 U2 Control Inputs ( $\overline{\mathrm{CS}}, \mathrm{A0}, \overline{\mathrm{WR}}, \overline{\mathrm{RD}}$ )

The U2 interface is addressed by the microprocessor data bus at hexadecimal 1044 or 1045. The address is latched by U18 when the ALE signal goes low, U19 is enabled and the $\overline{\mathrm{CS}}$ signal (the Y2 output of U19) goes low to enable U2. The level on the A0 input to U1 is established by the LSB of the address. The microprocessor uses address 1044 (LSB $=0$ ) if the information to be transferred on the data bus is to be interpreted (by U2) as data. The address 1045 ( $\mathrm{LSB}=1$ ) is used if the data bus word is to be interpreted as command status. The WR signal goes low to permit the microprocessor to load the bus with a word to be transmitted to U 1 ; the word is latched into U 2 when $\overline{\mathrm{WR}}$ transitions high. The microprocessor sets $\overline{\mathrm{RD}}$ low to request information from U2. The U2 data buffers load the information on the data bus when $\overline{\mathrm{RD}}$ is low, and when $\overline{\mathrm{RD}}$ transitions high, the information is latched by the microprocessor.

# Courtesy of http://BlackRadios.terryo.org 

### 3.1.3.9.2 U2 Scan Line Outputs/Return Line Inputs

The internal circuits of U2 provide a binary-coded count on the SL0 through SL3 output lines. Three bits of the binary code ( 0 through 7) are decoded by U12 (described in paragraph 3.1.3.10) to become the column scan lines (BC0 through BC7). Each column scan line is connected to one of eight X planes in the Group 2 switch matrix. Each switch in the matrix is located at a point where the X and Y planes cross and can be identified by the unique position it occupies. The Y planes of the matrix are connected to the return lines (RL0 through RL7) of U2. When a switch is closed, a circuit is completed from the column scan line, through the closed switch, and through the return (row scan) line back to U2. The closed switch position is coded and stored in U2 RAM. During the execution of its program, the microprocessor initiates any further action necessary to incorporate the functions of the closed switch.

The binary count ( 0 through 15) on the SL0 through SL3 lines is also applied, to multiplexer U8, where one of 16 channels is switched as the count proceeds (refer to paragraph 3.1.3.11). The resultant high output from U8 results (through U9 and U10) in a negative cathode voltage to a front panel LED indicator or display array (see paragraphs 3.1.2.2.3 and 3.1.2.2.4). The LED, however, is not forward biased, and does not illuminate, unless its anode is sufficiently more positive than the cathode. The anode drive is provided through the B0 through B3 outputs of U2, described in the following paragraph.

### 3.1.3.9.3 U2 Output Ports B0 Through B3

The B0 through B3 lines provide a BCD code to display driver U7, described in paragraph 3.1.3.13. The lines are synchronized to the scan lines, allowing U2 to automatically scan the switches and update or refresh the front panel address displays. In addition, the BCD output can be programmed by the microprocessor to reflect the data from a monitored receiver or the contents of a recalled memory channel.

### 3.1.3.10 Column Driver (U12)

Integrated circuit U12 is identical to U11, described in paragraph 3.1.3.5. Refer to paragraph 3.1.3.9.2 for a discussion of the functions of the $Y$ outputs.

### 3.1.3.11 Multiplexer (U8)

Multiplexer U8 functions identically to U4, described in paragraph 3.1.3.6. The U8 multiplexer is associated with the Group 2 keyboard/display interface, U2.

### 3.1.3.12 Cathode Drivers (U9, U10)

Cathode drivers U9 and U10 function identically to U5 and U6, described in paragraph 3.1.3.7. U9 and U10 are associated with the Group 2 keyboard/display interface, U2.

### 3.1.3.13 Display Driver (U7)

Display driver U7 functions identically to U3, described in paragraph 3.1.3.8. The outputs of U7 drive the receiver and memory address displays and the LED indicators associated with the Group 2 switch matrix, described in paragraph 3.1.2.2.2.

### 3.1.3.14 Data Latch (U20)

U20 is an octal D-type flip-flop. The CLR input is tied high, allowing data at the D inputs to transfer to the $Q$ outputs when the clock (CLK) input transitions to a high. During the time clock remains at either a high or low level, the changes at the D inputs have no effect on the $Q$ outputs.

The clock input to U 20 is the output from U16, configured to provide an OR function. The active low signals on the U16 inputs are the Y3 output of U19 and the AND gate U15 output. In paragraph 3.1.3.2.2, it was established that U19Y3 goes low only when the address 1046 has been on the bus and latched by U18. The AND gate (U15) is configured (pin 12 tied high) so that the output follows the level of the $\overline{W R}$ signal. In summary, the clock signal to U20 transitions high only when the microprocessor initiates a write to address 1046, allowing the U16 output to go low, then high on the rising edge of $\overline{\mathrm{WR}}$. It should be noted that the U16 output is normally high and has no effect in clocking data to U20.

The microprocessor writes data to U20 on the four least significant bits of the address/data bus. The microprocessor establishes a high level on the AD3 line when BFO frequency is to be displayed. The level, latched to the Q4 output of U20, is the FDP (Frequency Decimal Point) output to the Front Panel Switchboard LED segment of the BFO frequency display which illuminates to provide the decimal point. The high FDP level ensures that the decimal point illuminates during all BFO frequency displays. The data on the remaining three bits are latched to the Q1, Q2 and Q3 outputs to establish the switehing characteristics of U14 (refer to paragraph 3.1.3.15). The levels on the three bits are determined by the operating mode of the controller.

In local mode, the RF gain of the receiver (in manual gain mode) is established by the front panel RF GAIN potentiometer. The microprocessor uses the LSB of the data bus to establish the level at Q3 of U20 which controls the switching of the RF gain voltage (by U14) to the VOLTAGE 2 output. VOLTAGE 1 is unused in this application.

The dual-purpose front panel meter indicates signal strength or audio level of the receiver. When the front panel LINE AUDIO or SIGNAL STR switch is engaged, the processor uses the AD1 and AD2 data bus bits to establish the levels at Q1 and Q2 which control the switching (by U14) of the appropriate voltage to the meter.

### 3.1.3.15 RF Gain/Meter Voltage Switch (U14)

Switch U14 is a triple two-channel multiplexer having separate digital control inputs ( $\mathrm{A}, \mathrm{B}$, and C ) and an inhibit input. The inhibit line is tied to ground, holding the multiplexer in an enabled state. Each control input operates one of a pair of channels which are connected in a single-pole, double-throw configuration. The A input controls the X channels, the B input controls the $Y$ channels, and the C input controls the Z channels.

A low level on a control input connects the related $\mathrm{X} 0, \mathrm{Y} 0$, or Z 0 line to the $\mathrm{X}, \mathrm{Y}$ or Z channel; a high level on a control input connects the related $\mathrm{X} 1, \mathrm{Y} 1$, or Z 1 line to the $\mathrm{X}, \mathrm{Y}$, or Z channel. For example:

| $\mathrm{A}=1$, | $\mathrm{X}=\mathrm{X} 1$ |
| :--- | :--- |
| $\mathrm{~A}=0$, | $\mathrm{X}=\mathrm{X} 0$ |
| $\mathrm{~B}=1$, | $\mathrm{Y}=\mathrm{Y} 1$ |
| $\mathrm{~B}=0$, | $\mathrm{Y}=\mathrm{Y} 0$ |
| $\mathrm{C}=1$, | $\mathrm{Z}=\mathrm{Z} 1$ |
| $\mathrm{C}=0$, | $\mathrm{Z}=\mathrm{Z} 0$ |

The processor establishes the levels on the select inputs through data latch U20, described in paragraph 3.1.3.14. The C input to U14 switches the variable RF gain voltage (RFG) at Z to the VOLTAGE 2 output from Z0. The A input switches the signal strength (SS) voltage to the Y0 channel where the B input selects either the signal strength or audio level voltage to be sent through Y0 to the front panel meter (from XA1A pin A33).

### 3.1.3.16 Tuning Wheel Register (U17)

Integrated circuit U 17 is a 4 -bit D-Type flip-flop register with tri-state outputs. Two of the flip-flops are used in this application. The flip-flop outputs, Q1 and Q2, tell the microprocessor if the front panel tuning wheel is being rotated and, if so, in which direction: clockwise to increase the displayed frequency (at the selected step size) or counterclockwise to decrease the frequency.

Table 3-6 is the U17 truth table. The table shows that the levels at the D inputs transfer to the $Q$ outputs on the positive-going transition of the clock pulse when G1, G2, and clear (CLR) are low; in addition, the $Q$ outputs must be enabled by low levels on the OD1 and OD2 inputs. A high level on the clear input resets the flip-flops and all Q outputs go low.

The low levels required to allow data at the $D$ inputs to be clocked to the Q outputs are provided as follows:

1. G1 and G2 are tied to ground.
2. OD1 is low (through AND gate U15C) when $\overline{\mathrm{RD}}$ is low.
3. OD2 is low when U19Y0 goes low or, as explained in paragraph 3.1.3.2, when the hexadecimal address 1040 has been on the microprocessor address/data bus and latched by U18.
4. Clear is the inverted level on U19Y1 (by U15D). When U17 is addressed, only one of the U19 outputs (Y0) is low, all others are high; therefore, clear is low when U17 is addressed by the microprocessor. When clear goes high, the Q outputs of U17 are reset. The processor resets U17 by a $\overline{\mathrm{RD}}$ to address 1048.

Table 3-6. U17 Truth Table


NOTES: 1. Q0 denotes level of $Q$ before the steady-state condition was established.
2. When either OD1 or OD2 is high, the output is disabled to the high-impedance state; however, sequential operation of the flip-flop is not affected.

In summary, the U17 outputs (Q1 and Q2) reflect the data on the D1 and D2 inputs when the microprocessor addresses U17 at address 1040 and initiates a read operation. The D1 input is tied to +5 Vdc and, therefore, is always high. This high level will be clocked to the Q1 output the first time the clock (CLK) input transitions high and Q1 will remain high until U17 is reset (clear high); a high level at Q1 is evidence that the clock has transitioned high. The clock input is the inverted (by 21 C ) clock output of optical encoder MFP-A2U1, described in paragraph 3.1.1.1.

The optical encoder converts the rotation of the front panel tuning wheel into electrical pulses, and outputs two trains of square wave pulses (on the clock and direct lines) which are applied to the Front Panel Encode Board at pins B19 and B20 of XA1B. The trains of pulses are phase-related to the direction of tuning wheel rotation; DIR leads clock if the rotation is clockwise, as shown in Figure 3-4, and lags clock if the rotation is counterclockwise, as shown in Figure 3-5.

The Q1 output of U17 is high only when clock transitions high; cloek transitions high only when the front panel tuning wheel is rotated (and clear is low). The microprocessor addresses U17 for a read operation approximately every 5 milliseconds during the execution of its program and "looks at" the Q1 output (on the LSB of the data bus). If Q1 is low, the processor knows that the tuning wheel is not being rotated and continues with the execution of its program.

If the microprocessor finds that Q1 is high, the tuning wheel is being rotated and the processor examines the Q2 output (on the AD1 data bus line) to determine the direction of rotation. At the same time the positive clock transition transferred the D1 input level to Q1, the D2 input level transferred to Q2. D2 is the inverted (by U21E) DIR line pulse train from the optical encoder. Figures 3-4 and 3-5 illustrate the timing for the clockwise and counterclockwise rotation of the tuning wheel. If the direction is clockwise, the D2 input is high each time the U17 clock transitions high; therefore, Q2 will go high (if it was low at the first
rising clock edge) and remain high until U17 resets or until the direction of rotation changes. If the direction is counterclockwise, the D2 input is low each time the U17 clock transitions high; Q2 will either be low from the time of U17 reset and will remain low, or will be high from clockwise rotation and will go low at the first rising clock edge.

The sequence of the microprocessor program is such that following a $\overline{R D}$ of U17, the processor addresses 1048 and U19Y4 goes low. This low level is inverted by U15D and clear goes high, resetting the Q1 and Q2 outputs to a low level. The reset procedure ensures that a high level on Q1 indicates that the tuning wheel has been rotated since the last time Q1 was examined and, furthermore, because only 5 milliseconds has elapsed, the tuning wheel is still being rotated and the microprocessor can proceed to increment or decrement the displayed frequency.


Figure 3-4. Timing Diagram For Tuning Wheel Clockwise Rotation

## 3.2 <br> MFP INTERFACE CIRCUIT DESCRIPTIONS

The interface section of the MFP Option consists of two circuit boards, which contain an Intel 8085A Microprocessor with memory, read-only, and write-only devices to support the microprocessor.

The discussions in paragraphs 3.2.1 through 3.2.3.5 concern the microprocessor and the functional areas directly related to the microprocessor. Detailed circuit descriptions of the IF and Synthesizer Interface/Memory Boards (MFP-A3 and A4) follow, in paragraphs 3.2.4 and 3.2.5.


Figure 3-5. Timing Diagram for Tuning Wheel Counterclockwise Rotation

### 3.2.1 MEMORY DEVICES

Memory provides locations to store instructions and data for the microprocessor. The processor deals with 8 -bit binary fields. The memory associated with the processor is organized to store eight bits in each location. Data and instructions are stored in memory as 8 -bit binary numbers, or as numbers that are integral (even) multiples of eight bits: 16 bits, 24 bits, and so on. This characteristic 8-bit field is referred to as a byte. Each byte is stored in a memory location that is numbered to distinguish it from all other locations. The number which identifies a memory location is called its address.

A group of logically related instruction words, or program, is stored in sequential memory locations. The microprocessor reads each instruction in memory in a logically determined sequence and uses it to initiate processing actions. The instruction program is stored in an EPROM integrated circuit on the Synthesizer Interface/Memory Board, MFP-A4.

An addressable area of memory is required to store receiver parameter data for easy access during transmission, when a remote control equipment request is received or for updating as locally or remotely selected receiver parameters are changed. Memory for this purpose is provided by RAM integrated circuits on the Synthesizer Interface/Memory Board, MFP-A4. The RAM also provides an area of memory which the microprocessor uses as a "scratch pad", to temporarily store data being processed. Another area of RAM is used as a stack, described in paragraph 3.2.3.1.1.

### 3.2.2 READ-ONLY/WRITE-ONLY DEVICES

Flip-flop and buffer integrated circuits on the MFP-A1, A3, and A4 boards serve as addressable read-only and write-only devices. The devices allow the microprocessor to monitor the status of the receiver's front panel controls and update receiver parameters. If the receiver is equipped with an optional remote interface, the microprocessor uses read-only and write-only devices to incorporate remote commands and to transmit monitored receiver parameter data.

### 3.2.3 MICROPROCESSOR THEORY

The microprocessor, located on MFP-A4, unifies the system by controlling the functions of memory, read-only, and write-only devices. The processor must be able to access, decode, and execute the binary-coded instructions in the program. During the execution of the program, the microprocessor must be able to reference memory, if necessary, and must be able to recognize and respond to interrupt signals. The microprocessor meets its obligations by utilizing internal circuits as illustrated in Figure 3-6, the Intel 8085A Microprocessor Block Diagram. A brief description of the operation of key functional areas within the microprocessor follows.

### 3.2.3.1 $\quad$ Registers

Registers are temporary storage units within the microprocessor. Some registers, such as the program counter, flag register, stack pointer, and instruction register have dedicated uses. The accumulator and all other general purpose registers are used for storage of intermediate data during the execution of instructions. The accumulator is also used during arithmetic operations.

### 3.2.3.1.1 Program Counter and Stack Pointer

The instructions that make up the program are stored in EPROM (Erasable Programmable Read Only Memory) on the MFP-A4 board. The microprocessor references the contents of the memory in order to determine what action is appropriate. This means that the processor must know which location contains the next instruction. The processor maintains a register which contains the address of the next program instruction. This register is the program counter. Because the microprocessor updates the program counter each time it fetches an instruction, the program counter is always pointing to the next instruction.

The program is stored in EPROM in numerically adjacent addresses (beginning at location zero). The lower order addresses contain the first instructions to be executed and the higher order addresses contain the later instructions. During the course of the instructions, the program may call, or transfer control to, a subroutine. A subroutine is a program within a program; of ten, it is a general purpose set of instructions which must be executed repeatedly during the program. Memory space is conserved by storing the instructions in an area apart from the main program and directing the processor to this area as necessary. The subroutine instructions are also stored in numerically adjacent addresses in EPROM. When a subroutine is called, the processor is instructed to jump to the lowest subroutine address. After the subroutine is finished, the processor must resume execution of the main program; therefore, the processor must remember the contents of the program counter at the time the call occurs.

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The microprocessor has a special way of handling subroutines in order to ensure an orderly return to the main program. When the processor receives a call instruction, it increments the program counter and stores the counter's contents in a reserved area in RAM on the MFP-A4 board. This memory area is called a stack. The stack saves the address of the instruction to be executed after the subroutine is completed. Next, the processor loads the subroutine address in its program counter, ensuring that the next instruction fetched will be the first step of the subroutine.

The last step of a subroutine is a return. After the processor fetches a return instruction, it replaces the current contents of the program counter with the address on the top of the stack and the program is resumed at the point immediately following the original call instruction.

The processor uses an area of RAM as a stack and maintains an internal pointer register which contains the address of the most recent stack entry. The external stack allows subroutine "nesting", a procedure during which one subroutine calls a second routine.

### 3.2.3.1.2 Instruction Register and Decoder

Each operation that the microprocessor can perform is identified by a unique byte of data known as an instruction or operation code, generally referred to as an opcode. An 8-bit binary-coded word used as an opcode can distinguish between 256 ( $2^{8}$ ) alternative actions.

The microprocessor fetches an instruction from memory (EPROM) in two distinet operations. First, the processor transmits the address in its program counter to the memory. Next, memory returns the addressed byte to the processor. The instruction byte is stored in a circuit within the processor called the instruction register and is used to direct the activities of the processor during the instruction execution.

The eight bits stored in the instruction register are decoded in the instruction decoder and machine cycle encoding network (Figure 3-6) and are used to activate the output lines of the decoder. The enabled lines are gated by timing signals in the timing and control block to develop electrical signals that initiate specific actions in the processor registers, ALU, and buffers. The outputs of the instruction decoder and internal clock generator provide the state and machine cycle timing signals (paragraph 3.2.3.2).

An 8-bit instruction may not always be sufficient to specify a particular processing action. If more than one byte is used for an instruction, successive instruction bytes are stored in sequentially adjacent memory locations. The microprocessor performs more than one fetch in succession to obtain the full instruction. The first byte retrieved from memory is placed in the instruction register while subsequent bytes are placed in temporary storage internal to the processor. The processor then proceeds with the execution phase.

### 3.2.3.1.3 Arithmetic Logic Unit (ALU) and Flag Register

The ALU is the portion of the microprocessor hardware which performs the arithmetic and logic operations on the binary data and is inaccessible to the programmer. (An example of the use of the ALU is when the program counter is incremented.)


Figure 3-6. Intel 8085A Microprocessor Block Diagram

The flag-bit register is associated with the ALU and the accumulator. The flagbits specify certain conditions that occur during the course of arithmetic and logical manipulations. The five Intel 8085A Microprocessor flag-bits are carry, auxiliary, sign, zero, and parity. The bits are important to the programmer in establishing the control of processor operation.

### 3.2.3.2 Internal Clock Generator and Timing

The activities of the microprocessor are cyclical. The processor fetches an instruction, performs the required operations, fetches the next instruction, and so on. This orderly sequence of events requires precise timing. Timing is provided by an external oscillator.

The combined fetch and execution of a single instruction is referred to as an instruction cycle and consists of a series of machine cycles, whose nature is determined by the opcode. The opcode is accessed in the first machine cycle of an instruction cycle. Each machine cycle consists of a series of clock, or timing, cycles determined by the type of instruction being executed and the machine cycle within the instruction cycle.

### 3.2.3.3 Interrupt Control

To the microprocessor, an interrupt signal is similar to a subroutine call except that it is initiated externally rather than by the program. If an optional remote interface has been installed, a high level on an interrupt line (RST 5.5 or RST 6.5) is generated by the I/O port on the I/O Interface Board ( $488 \mathrm{M}-\mathrm{A} 3$ or $232 \mathrm{M}-\mathrm{A} 3$ ) when a data word has been received from or requested by the controller. An interrupt request sets a processor interrupt enable flip-flop; the processor acknowledges the interrupt by suspending the execution of the main program and automatically branches to a subroutine to service the interrupt. The status of the main program is stored on the stack in RAM (paragraph 3.2.3.1.1) until the processor finishes the interrupt service and returns to the main program.

### 3.2.3.4 Data Bus Control Lines ( $\overline{\mathrm{RD}}, \overline{\mathrm{WR}}, \mathrm{ALE}$ )

At the beginning of a fetch machine cycle, the processor places the contents of the program counter (a memory address) on the 16 -bit address bus. The high-order byte of address data is placed on the A8 through A16 lines and will remain there for several clock cycles. The low-order byte is placed on the AD0 through AD7 lines and the microprocessor line drivers are enabled. Unlike the upper address lines, the information on the lower address lines will remain there for only one clock cycle after which the drivers will go to their high-impedance state. This is necessary because the AD0 through AD7 lines are multiplexed between the address and data. During the first clock cycle of a machine cycle, AD0 through AD7 output the eight lowest bits of the address, after which the lines either output the desired data for a write operation or the drivers will float, allowing the external device to drive the lines for a read operation.

The address information on the AD0 through AD7 lines is transitory, therefore, it must be latched into selected external 8 -bit latches. To facilitate the latching of the lowest eight bits of data, the microprocessor provides a special timing signal, Address Latch Enable
(ALE), during the first clock state of each machine cycle. After the eight address bits have been latched, the processor initiates either a read or write operation by providing a low level on the $\overline{\mathrm{RD}}$ or $\overline{\mathrm{WR}}$ control line.

Figure 3-7 illustrates the timing of a read cycle. A low level on the $\overline{R D}$ line enables the addressed memory device and, after a period of time (the access time of the memory), valid data will be present on the AD0 through AD7 lines. The processor next loads the data into its instruction register (paragraph 3.2.3.1.2) and raises $\overline{R D}$ high, disabling the addressed memory device.

After ALE drops, a low level on the $\overline{W R}$ line causes data to be placed on the AD0 through AD7 lines by the processor. The data is loaded into the addressed memory when $\overline{W R}$ goes high. Figure 3-8 illustrates a write timing pulse.

The A8 through A15 data lines are used to identify the MSBs of a memory or I/O location for a data transfer cycle. Selected bits provide enabling pulses to address decode circuits which, in turn, provide enabling pulses to read-only and write-only registers on the MFP-A1, A3, and A4 boards, and, if the 232 M or 488 M Option is in use, on the $232 \mathrm{M}-\mathrm{A} 3$ or 488M-A 3 boards.


Figure 3-7. $\overline{\mathrm{RD}}$ Timing Diagram

### 3.2.3.5 $\quad$ Reset In/Reset Out

The microprocessor is reset when the RESET $\overline{\text { IN }}$ line is low and commences execution of its program when the line goes high. The RESET OUT line goes high to reset the MFP-A1 display interface circuits and, if the 232 M or 488 M Option is in use, the $232 \mathrm{M}-\mathrm{A} 3$ or 488M-A3 I/O port.


Figure 3-8. $\overline{W R}$ Timing Diagram
3.2.4

TYPE 794275-X SYNTHESIZER INTERFACE AND MEMORY BOARD (MFP-A4)
The Synthesizer Interface Board replaces the Manual Tuning Up/Down Counter (A6A1) in the WJ-8718A HF Receiver. The board contains the Intel 8085A Microprocessor, which is the control for the MFP Option. Random Access Memory (RAM) provides temporary storage, and Erasable Programmable Read Only Memory (EPROM) contains the microprocessor's instruction program. The board interfaces the microprocessor data bus with the BFO and LO Synthesizers through data latches. Figure 3-9 is a simplified block diagram of the MFP-A4 board and Figure 6-6 is the schematic diagram.

### 3.2.4.1 Microprocessor (U18)

The Intel 8085A Microprocessor is an 8-bit general purpose microprocessor that requires a single +5 V supply. The microprocessor contains eight addressable 8 -bit general purpose registers and two 16-bit non-addressable registers which make up the CPU.

A multiplexed data bus allows the microprocessor to communicate with external devices. The execution of the microprocessor program consists of a series of machine cycles. Each machine cycle contains a series of clock cycles. During the first clock cycle of a machine

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Figure 3-9. Synthesizer Interface and Memory Board, Simplified Block Diagram

## Courtesy of http://BlackRadios.terryo.org

cycle, an external device address is placed on the data bus. The address is divided between the higher-order 8-bit address (A8 through A15) and the lower-order 8-bit address/data bus (AD0 through AD7). The eight lower-order address bits are latched into external devices by the Address Latch Enable (ALE) signal. During the remaining portion of the machine cycle the bus is used for data.

The microprocessor provides $\overline{\mathrm{RD}}, \overline{\mathrm{WR}}$, and ALE outputs for bus control; and RST 5.5 , RST 6.5 , and RST 7.5 for interrupts. RST 7.5 is used for timing events and is driven by an RC oscillator circuit U11B and U11C. RST 5.5 and RST 6.5 are only used when an optional remote I/O is installed.

The microprocessor is driven by oscillator circuit U11E and U11D at a frequency of approximately 2 MHz ; this is internally divided by 2 to give the bus an operating speed of 1 MHz .

### 3.2.4.2 Power Up/Down Circuit (U11)

The Power Up/Down Detect circuit drives the $\overline{\text { RESET IN }}$ input to the Intel 8085A Microprocessor. When this input goes low all operations stop and all bus outputs become a high impedance state. The input to the circuit is unregulated 10 V , dropped to approximately 1.5 V by VR1, R2, CR1, and R3. This voltage keeps the U11A input slightly above threshold and enables the RESET IN input to be high, allowing the microprocessor to operate.

During power-on, R6 and C9 provide an RC time constant to allow the microprocessor to become fully energized before being released to start operation. CR2 isolates U11 from the RC circuit. During power-down, the U11A input goes below threshold causing the U11F output to go low; this biases CR1 and discharges C9, resetting the microprocessor in the process.

### 3.2.4.3 Read-Only Memory (U1, U2)

The Synthesizer Interface Board houses two 32,768-bit ( $8 \mathrm{~K} \times 8$ ) ultraviolet erasable and electrically programmable read-only memories (EPROMs), U1 and U2.

Prior to shipment, the EPROMs are programmed with the microprocessor instruction set. Before programming, all bits in the EPROM are in a 1 state. Data is introduced by selectively programming 0 s into the desired bit locations (see Table 3-7). The only way a 0 can be changed to a 1 is through ultraviolet erasure. Erasure occurs when the EPROM window is exposed to light with wavelengths shorter than approximately 4000 Angstroms. Exposure to sunlight, and certain types of fluorescent lamps with wavelengths in the 3000 to 4000 Angstroms range, should be avoided to prevent unintentional erasure.

### 3.2.4.3.1 EPROM Operation

Power required for the operation of each EPROM is +5 Vdc . The programmed read-only memory devices, U1 and U2, have two modes of operation: read and standby. All inputs are TTL levels.

Table 3-7. RAM/EPROM Address Data

| Address bit |  | $2^{3}-2^{0}$ | $2^{3}-2^{0}$ | $2^{3}-2^{0}$ | $2^{3}-2^{0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- |
| Address Line | $15-12$ | $11-8$ | $7-4$ | $3-0$ |  |
| EPROM 1 |  |  |  |  |  |
| From 0000 | $=$ | 0000 | 0000 | 0000 | 0000 |
| To 0FFF | $=$ | 0000 | 1111 | 1111 | 1111 |
| EPROM 2 |  |  |  |  |  |
| From 4000 | $=$ | 0100 | 0000 | 0000 | 0000 |
| To 4FFF | $=$ | 0100 | 1111 | 1111 | 1111 |
| RAM |  |  |  |  |  |
| From 2000 | $=$ | 0010 | 0000 | 0000 | 0000 |
| To 27FF | $=$ | 0010 | 0111 | 1111 | 1111 |

### 3.2.4.3.2 Control Inputs

The $\overline{\mathrm{CE}}$ inputs to U1 and U2 are chip-select pulses from decode circuit U7. The low enable pulse to U 1 is provided when decoder U7 senses that the hexadecimal-coded address on the microprocessor data bus is in the range of 0000 through 0 FFF ; the low enable pulse to U2 is provided when the hexadecimal address is in the range of 4000 through 4 FFF . The microprocessor addresses location 0000 at power-up and consecutively reads and executes the instructions stored in EPROM.

The $O E$ inputs are tied to the microprocessor $\overline{\mathrm{RD}}$ signal, which is pulled low when the processor initiates a read operation. Output Enable (OE) is used to take the outputs out of tri-state.

### 3.2.4.3.3 Read Mode

When the microprocessor initiates a read operation, the 16 -bit memory address is placed on the data bus, ALE goes low, and if the memory addressed is RAM or EPROM, U7 is enabled. If the hexadecimal coded address is between 0000 and $0 F F F$, U7 pulls the U1 Chip Enable ( $\overline{\mathrm{CE}}$ ) line low, and the AD0 through AD7 address bits (from low-order address latch U5) and the A8, A9, A10, and A11 address bits are applied to the EPROM device through Address Inputs A0 through A11. After an address access time, followed by the falling edge of the microprocessor RD signal, the data stored in the addressed memory location are available at the 00 through 07 EPROM Data Outputs. In the same manner, if the hexadecimal address is between 4000 and 4 FFF , U7 enables EPROM U2.

The data are loaded on the AD0 through AD7 address/data bus lines and latched into the microprocessor's internal instruction register (paragraph 3.2.3.1.2), on the rising edge of $\overline{R D}$.

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## CIRCUIT DESCRIPTION

### 3.2.4.3.4 Standby Mode

The EPROM devices are in reduced power standby mode when the $\overline{\mathrm{CE}}$ enable lines are held at TTL high logic level. During the standby mode, the outputs are in a high impedance state, independent of the OE input, allowing the shared data bus to be driven by other devices addressed for a read operation.

### 3.2.4.4 Random Access Memory (U3)

The Synthesizer Interface Board houses a single addressable read/write memory device, consisting of a 16,384 ( $2 \mathrm{~K} \times 8$ ) CMOS RAM integrated circuit, (U3). RAM stores data, current receiver status, and serves as a stack for the temporary storage of the microprocessor's internal registers during a program interrupt. Refer to Table 3-7 RAM/EPROM Address Data for the binary equivalent addresses of the various chips.

### 3.2.4.4.1 RAM Operation

Power required for the operation of RAM U3 is +5 Vdc . The modes of operation consist of read, write, high-Z, and low power.

### 3.2.4.4.2 Control Input

The $\overline{\mathrm{CE}}$ input from decoder U7 goes low when the hexadecimal address on the microprocessor data bus is in the range of 2000 through 27 FF .

The read and write inputs are connected to the microprocessor $\overline{\mathrm{RD}}$ and $\overline{\mathrm{WR}}$ bus control signals and are active (low) for their respective bus cycles.

### 3.2.4.4.3 Read Mode

When the decoded address is in the 2000 to 27 FF range and the microprocessor initiates a read operation, all bus cycles are identical to the read operation previously described for EPROM (see paragraph 3.2.4.3.3).

### 3.2.4.4.4 Write Mode

When the decoded address is in the range of 2000 to 27 FF and the microprocessor initiates a write operation, decoder U7 output Y2 becomes active (low) and enables U3 address inputs A0-A10. During the data portion of the bus cycle, $\overline{\mathrm{WR}}$ line becomes active (low) and a rising edge of $\overline{\mathrm{WR}}$ causes data to be written into the internal address selected by A0-A10.

### 3.2.4.4.5 High-Z Mode

Any microprocessor bus cycle that does not enable U1, U2, or U3 leaves all outputs in a high-Z (high impedance) state. This disables the devices from having any effect on the bus.

### 3.2.4.4.6 Low Power Mode

RAM U3, being CMOS, can be kept powered on to maintain all data stored in the memory by holding Vec/Vdd and $\overline{\mathrm{CE}}$ to a 2 V level. Buffer U10D provides isolation for $\overline{\mathrm{CE}}$ with power off; its open collector output is pulled up by R14.

### 3.2.4.4.7 Battery Backup

An inherent quality of RAM is that it is volatile, meaning that data is not retained when power is removed. To prevent the loss of data, a battery (BT1) is used to power the RAMs when power is down.

Transistor Q1, forward biased by Vee when power is on, charges the 2.4 V battery through R4 and R5. When power is down, Vec drops to zero and the transistor becomes reverse biased, allowing current to flow only to the RAM circuits connected to Vdd.

### 3.2.4.5 Bidirectional Bus Transceiver (U4)

Transceiver (U4) is made up of 16 high-speed CMOS buffer drivers, only eight of which are enabled at the same time. The direction of the data (or which eight buffers are enabled) is controlled by the DIR input and is connected to the $\overline{R D}$ line of the microprocessor through NAND gate U9D. NAND gate U9Ds $\overline{\text { INTA }}$ input is not used in this design. When $\overline{R D}$ is active (low), the eight buffers are enabled to allow bus data to be input to the microprocessor. The rest of the time these buffers are tri-state. The opposite buffers are then enabled to allow data to be output from the microprocessor to the bus.

### 3.2.4.6 Address Latch (U5, U6)

High-speed CMOS U5 and U6 are octal D-type flip-flops. The data on the inputs of U5 and U6 contain the low-order and high-order address portion of the bus cycle. When ALE goes low indicating a valid address on the bus, this signal is inverted by U9C. The resultant positive-going edge, when applied to the clock inputs of $U 5$ and $U 6$, clocks the $D$ inputs to the $Q$ outputs. These outputs remain in this state until the next bus ALE cycle. This demultiplexes the address from the data portion of AD0-AD7 (low-order) lines and provides a bus cycle of equal duration on the A8-A15 (high-order) bus address lines. The EN input is tied to the HLDA output of the microprocessor and is not used in this design.

### 3.2.4.7 Address Decoders (U7, U8)

Table 3-8 is a truth table for U7, the high-order address decode operation, and Table 3-9 is the truth table for the low-order address decode operation. When enabled by applying a high to input G1 and a low to inputs G2A and G2B, these decoders provide a low logic level on one of seven $Y$ outputs.

The U7 inputs are the most significant four bits of the address bus A12 through A15. A15 must always be low in order for either of these devices to be enabled. U7 outputs, Y0-Y4, are enabled for ROM1, U8 enables G2B (necessary for frequency and BFO data lateh strobes), RAM, and ROM2.

Table 3-8. U7 Truth Table

| Inputs |  |  |  |  |  | Outputs |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| G1 | Enable G2B | G2B | C | $\begin{aligned} & \text { Select } \\ & \text { B } \end{aligned}$ | A | Y0 | Y1 | Y2 | Y3 | Y4 |
| VCC | GND | A15 | A14 | A13 | A12 |  |  |  |  |  |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 | 1 | 0 |

Table 3-9. U8 Truth Table

| Inputs |  |  |  |  |  | Outputs |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| G1 | Enable G2B | G2A | C | $\begin{gathered} \text { Selec } \\ \text { B } \end{gathered}$ | A | Y0 | Y1 | Y2 | Y3 | Y4 | Y5 |
| $\overline{\mathrm{WR}}$ | * U7Y1 | A4-A5 | A2 | A1 | A0 |  |  |  |  |  |  |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 0 | 1 | 1 | , | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 | 0 | 1 | 1 |
| 1 | 0 | 0 | 1 | 0 | 0 | 1 | 1 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 0 | 1 | 1 | 1 | 1 | 1 | 1 | 0 |

The U8 inputs are the three least significant address bits of the bus, address bits 4 and 5, Y1 decode output from U7, and $\overline{W R}$. Any microprocessor write cycle, to address $1 \times 3 x$, causes G2A and G2B to be low and G1 to be high through Nand gates U9B and U9A. The U8 outputs Y0-Y5 are clock strobes which latch bus data into the write-only registers used for storing frequency data for the LO synthesizers and BFO synthesizers. Table 3-10 lists the binary equivalent of the four hexadecimal address digits, 1030 through 1035.

Table 3-10. Addresses 1030 Through 1035

| Address Bit | $2^{3} 2^{2} 2^{1} 2^{0}$ |  |  |  | $2^{3} 2^{2} 2^{1} 2^{0}$ |  |  |  | $2^{3} 2$ |  | $2^{2}$ |  |  | $2^{3}$ | $2^{2}$ | $2^{1}$ |  | $2^{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Data Line |  | 14 | 13 |  | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 5 | 4 | 3 | 2 | 1 |  | 0 |
| U16 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 0 | 0 | 0 |  | 0 |
| U15 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 0 | 0 | 0 |  | 1 |
| U14 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 0 | 0 | 1 |  | 0 |
| U13 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 0 | 0 | 1 |  | 1 |
| U12 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |  | 1 | 0 | 1 | 0 |  | 0 |
| U17* | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 |  |  | 1 | 0 | 1 | 0 |  | 1 |

NOTE: Data Lines 0-7 = AD0-AD7
Data Lines 8-15 = A8-A15
The 26 -bit RF frequency data words are provided in Table 3-11 and the 12-bit BFO frequency data words are provided in Table 3-12.

Table 3-11. RF Frequency Data Words

| Frequency Digit | $10^{1}$ | $10^{2}$ | $10^{3}$ | $10^{4}$ | $10^{5}$ | $10^{6}$ | $10^{7}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Frequency Bits | $2^{0}-2^{3}$ | $2^{0}-2^{3}$ | $2^{0}-2^{3}$ | $2^{0}-2^{3}$ | $2^{0}-2^{3}$ | $2^{0}-2^{3}$ | $2^{0}, 2^{1}$ |
| Register Outputs | Q7-Q4 | Q3-Q0 | Q7-Q4 | Q3-Q0 | Q7-Q4 | Q3-Q0 | Q7, Q6 |
| Register | U 16 | U 16 | U 15 | U 15 | U 14 | U 14 | U 13 |

Table 3-12. BFO Frequency Data Words

|  | Sign |  | Frequency Digit |  |  |  |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: |
| Frequency Data | + | - | $10^{1}$ | $10^{2}$ | $10^{3}$ |  |
| Frequency Bit | NA | $2^{3}-2^{0}$ | $2^{3}-2^{0}$ | $2^{3}-2^{0}$ |  |  |
| Register Outputs | Q5 | Q4 | Q3-Q0 | Q7-Q4 | Q3-Q0 |  |
| Register | U13 | U13 | U13 | U12 | U12 |  |
| NOTE: NA denotes not applicable. |  |  |  |  |  |  |

### 3.2.4.8 Frequency Registers (U12-U17)

U12 through U17 are octal D-type flip-flops and are identical to the U5 and U6 latches described previously, except they are standard CMOS parts. These registers serve as latches for the receiver LO synthesizer and BFO synthesizer circuit.

The clock input to each register is tied to one of the Y outputs of decoder U8. When a hexadecimal address between 1030 and 1035 is on the data bus and a write cycle is executed, one of the Y outputs goes low for the duration of the inverted write input on G1. This positive-going edge of the $Y$ output causes its data on the bus to be clocked from the $D$ input to the Q output. These signals are then used by the necessary inputs to its LO and BFO circuits.

### 3.2.5 TYPE 794308-2 IF INTERFACE (MFP-A3)

The IF Interface Board replaces the Front Panel Interconnect (A6A2) in the WJ-8718A HF Receiver. The board provides signals that control appropriate devices in the IF section of the receiver. These functions include gain mode, bandwidth, detection mode, RF gain control, AGC dump, sideband audio switching, audio squelch, signal strength monitoring, synthesizer tuning voltages, $\pm 15 \mathrm{~V}$, and audio outputs.

The IF Interface Board also provides circuitry for interfacing to a standard front panel WJ-8718A HF Receiver. These circuits are omitted for use when installed in a MFP equipped receiver; therefore, the Type 794308-1 IF Interface Board, which contains these circuits, is not covered in this discussion.

Figure $3-10$ is a simplified block diagram and Figure 6-5 is the schematic diagram for the MFP-A3 board.

### 3.2.5.1 Bidirectional Bus Transceiver (U20)

Transceiver U20 is made up of 16 high speed CMOS buffer devices. Only eight buffers are enabled at one time. The direction, or the specific buffers that are enabled, is controlled by the DIR input at pin 1. The pin 1 input is connected to the $\overline{\mathrm{RD}}$ line through U13B and U19A for any address decode of 101X (through U10D) or 103X (through U10A). Refer to Table 3-13 for additional 101X addressing information. (Besides decoding, the 103 X is used for devices on the Type 794308-1 Board for a standard receiver, and is therefore not mentioned again in this description.)

When the $\overline{\mathrm{RD}}$ is active for one of the decoded addresses, the buffers are enabled to allow data to be transferred from the board to the external circuits. In all other cases, these buffers are disabled and their opposites are enabled to allow data transfer from the external circuit to the board.


Figure 3-10. IF Interface Board, Simplified Block Diagram

TABLE 3-13

Table 3-13. Addresses 1010 through 1017


### 3.2.5.2 Address Latch (U8)

The data on the inputs to U8 contain the low order address portion of the bus cycle from Transceiver U20. When ALE goes low, indicating a valid address on the bus, this signal is inverted by U10B. The resultant positive-going edge when applied to the clock input of U8 latches the data into the Q outputs. These outputs remain in this state until the next ALE cycle. This demultiplexes the address from the data portion of AD0-AD7 low order lines.

### 3.2.5.3 Decoder Select (U11)

The select inputs to U11 (A0, A1, and A2) are the three least significant bits of the address latehed by U8. As shown in Table 3-14, a Y output goes low only when U11 is enabled and a particular 3-bit code is on the select inputs. Comparison of the select input codes listed in Table 3-14, with the three least significant bits of the addresses listed in Table 3-13, shows that one of the Y0 through Y7 outputs goes low when one of the addresses in the hexadecimal 1010 through 1017 range is on the bus. Each Y output is tied to an enable input of a read-only device or the clock input of a write-only device on the IF Interface Board (MFP-A3).

When U11 is enabled, the Y output goes low during the time that the $\overline{\mathrm{RD}}$ or $\overline{\mathrm{WR}}$ signal is low. If the $\overline{R D}$ signal enables U11, the addressed device is a tri-state buffer (unused in this application) which is enabled by the low pulse from a U7 output. However, if the WR signal enables U7, the resultant high-to-low level from U7 to the clock input of the addressed device
does not change the status of the device. After the microprocessor replaces the address on the bus with data, the WR line goes high and U11 is disabled. The U11 output that went low when the $\overline{\mathrm{WR}}$ line was low now changes state; this positive-going clock transition latehes the data into the addressed device.

Table 3-14. U11 Truth Table

| Enable |  |  | Select |  |  | Outputs |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| G1 | G2A | G2B | A0 | A1 | A2 | Y0 | Y1 | Y2 | Y3 | Y4 | Y5 | Y6 |
| H | L | L | L | L | L | L | H | H | H | H | H | H |
| H | L | L | H | L | L | H | L | H | H | H | H | H |
| H | L | L | L | H | L | H | H | L | H | H | H | H |
| H | L | L | H | H | L | H | H | H | L | H | H | H |
| H | L | L | L | L | H | H | H | H | H | L | H | H |
| H | L | L | H | L | H | H | H | H | H | H | L | H |
| H | L | L | L | H | H | H | H | H | H | H | H | L |

### 3.2.5.4 RF Gain Conversion D/A (U23)

Digital-to-analog (D/A) converter U23 at address 1013 is used during scan operation during BITE tests, or if the receiver is under remote control. The 8 -bit binary-coded $R F$ gain word is latched into its inputs when the WR1 input is pulsed low during a $\overline{W R}$ cycle from the Y3 output of address decoder U11.

The converter provides an output current directly related to the coded RF gain word. The current is converted to a positive voltage by amplifier U24C through feedback input RFB. The output voltage is derived from a $R-2 R$ ladder network which can be deduced mathematically by the equation:
binary input $\times \frac{V R E F}{256}=$ output voltage.
Where $V^{\text {REF }}$ is derived from VR1 zener diode through the -15 V supply. The symbol $\mathrm{V}^{\text {REF }}$ equals -5.1 V , so a binary input bit multiplied by .019 V is an approximation of the output voltage.

### 3.2.5.5 Audio/Gain Switch (U22)

Analog switch U22 is a triple two-channel multiplexer, housing three separate digital control inputs: A, B, and C. Each input controls a pair of channels connected in a single-pole double-throw configuration.

Input A controls channel X which selects the RF gain voltage. Input A is tied to the remote/local latch U21B. When the receiver is in the local mode, the A input is held low and the RF gain potentiometer voltage is fed through channel X0 to X . When the receiver is in the remote mode, the A input is held high and the digital-to-analog RF gain voltage is fed through channel X1 to X.

# Courtesy of http://BlackRadios.terryo.org 

TABLE 3-15

Input B controls the output of the manual RF gain voltage. When manual AGC is selected, input B is held high and the voltage from channel X is output to the amplifier U24D through channel Y1 to Y. When the receiver is in either fast or slow AGC, input B is held low and the RF gain voltage is turned off by the ground on channel Y0 to Y.

Input C controls the audio output connected to the PHONES jack. In any detection mode but ISB, the audios on the ring and tip of the front panel stereo PHONES jack are the same. In ISB detection mode, U22 switches the audio so that the USB audio is on the tip of the jack and LSB audio is on the ring, allowing both audios to be monitored with a 600 ohm stereo headphone. Input C is held high by the latch U9 ISB output, connecting the ISB audio to feed through ehannel Z1 to Z.

### 3.2.5.6 Registers (U5, U9, U21)

Three registers serve as storage devices for the control of the IF section of the receiver. Each register latches the data on its $D$ inputs to the $Q$ outputs on a positive-going transition of its clock pulse.

To initiate a write operation, the microprocessor places the address of the device on the data bus and issues the ALE signal, latching the lower-order bits of the address into U8. The processor pulls the $\overline{\mathrm{WR}}$ line low and address decoder U11 is enabled. U11 decodes the address latched by U8, and the Y output connected to the clock or clock pulse input of the addressed device goes low. Y0 is the clock input U21B, and Y1, Y2, and Y7 are the clock pulse inputs to U5, U9 and U21A respectively.

The processor then pulls the $\overline{W R}$ line high, signaling that the data is on the bus. At this point, the high level on $\overline{W R}$ disables U11, and the $Y$ output that was low makes a positive transition. This positive-going pulse latches the data into the addressed device.

### 3.2.5.7 Bandwidth, BFO Inhibit, and AGC Dump (U5)

The bandwidth storage register U5 holds the bandwidth code for the particular bandwidth and detection mode selected. Table 3-15 lists the bandwidth code.

The Q5 output controls the AGC Dump circuitry used during scan operation to hold off any energy in the AGC circuitry. When held high, all signal strength is shut off. The Q7 output inhibits the BFO input when not in CW operation. Refer to Table 3-16 for the U5 data word.

Table 3-15. Bandwidth Code

Selected Bandwidths

| Selected Bandwidths | 16 kHz | 6 kHz | 3.2 kHz | 1 kHz | .3 kHz |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Active U5 Outputs | Q6 | Q6 | Q6 | Q2 | Q2 |
|  | Q4 | Q3 | Q2 | Q1 | Q0 |

Table 3-16. Address 1011 (U5) Data Word

| Output | Function |
| :---: | :---: |
| Q4 | BFO Inhibit |
| Q6 | 16 kHz (A) IF Bandwidth |
| Q5 | AGC Dump |
| Q4 | 16 kHz (B) IF Bandwidth |
| Q3 | 6 kHz IF Bandwidth |
| Q2 | 3.2 kHz IF Bandwidth |
| Q1 | 1 kHz IF Bandwidth |
| Q0 | .3 kHz IF Bandwidth |

### 3.2.5.8 Gain/Detector Mode (U9)

Table 3-17 is the data word for U9. The microprocessor updates receiver gain and detection mode parameters by writing the data into register U9. The Q outputs connect to the detection mode and AGC circuits in the IF section of the receiver. A high logic level indicates the selected gain or detection mode and activates the IF circuit to initiate the parameter change.

Table 3-17. Address 1012 (U9) Data Word

| Output | Function |
| :---: | :---: |
| Q7 | SLOW AGC |
| Q6 | MAN AGC |
| Q5 | LSB |
| Q4 | ISB |
| Q3 | USB |
| Q2 | CW |
| Q1 | FM |
| Q0 | AM |

### 3.2.5.9 Local/Remote Select (U21B)

Register U21B (address 1010) provides a select level to gain voltage switch U22. The microprocessor reads the status of the front panel REMOTE and LOCAL switches, during a read operation on the Front Panel Encode Board (MFP-A1A1), to determine if the receiver is in the local or remote operating mode. The local/remote status is contained in the LSB of the local/remote data byte in a remote monitor operation. The bit is low if the receiver is in local mode and high if the receiver is in remote mode. The logic level is applied to the D input of the

## Courtesy of http://BlackRadios.terryo.org

flip-flop U21B via data line AD0. The bit level is latched to the Q output of U21B when the clock input makes a positive transition. From U21B, the pulse representing the local/remote code is applied to the select input of U22. U22 switches between manual and remote RF gain voltage.

### 3.2.5.10 AUDIO ON/OFF (U21A, U16A, U16B)

Register U21A (address 1017) provides a select level to analog switches U16A and U16B which, in turn, provide audio squelching during scan operation. The output of U21A is low in normal operation allowing both audio channels to be routed to the front panel. During scan operation, the microprocessor provides a high level on the AD7 input to the register. The bit is latched to the $Q$ output on the positive-going transition of the clock input. When their select inputs are held high, U16A and U16B become an open circuit and provide a squelched audio output.

### 3.2.5.11 A/D Converter (U25)

Analog-to-digital converter U25 (address 1014) is made up of an eight-channel, multiplexed analog switch, an 8-bit A/D converter, address decode circuitry, and a tri-state output buffer.

Seven of the eight channel inputs to the multiplexer are connected to receiver signals and are monitored during receiver operation. Signal strength is monitored if scan, COR, or remote operation is being utilized. All other inputs are used during Built-In Test Equipment routines if the software program provides this feature.

To select an input, a select address code must be written to the $A, B$, and $C$ inputs of the A/D converter. Table $3-18$ shows the input states for the three address lines which are connected to AD4, AD5 and AD6 of the microprocessor. To select a channel, the address lines are latched into the address decoder by the low-to-high transition of the A/D ALE input. The A/D ALE input goes low when a $\overline{W R}$ to address 1014 occurs and decoder U11 output Y4 goes low. When the $\overline{\mathrm{WR}}$ signal returns to a high state, the output of U19C transitions high, latching in the address data into the address decoder.

After a channel has been selected, the data can be read from the output buffer by placing a microprocessor read instruction to address 1014 on the bus. This causes decoder U11 output Y4 to go low, as well as the output of U19D. The low from U19D is then inverted by U14C to enable the tri-state output buffer of the A/D converter. This places the conversion data in the buffer onto the bus for the microprocessor to accept.

The converter is the heart of the single-chip multiplexed A/D. The converter comprises three major parts: the 256R ladder network, a successive approximation register (SAR), and a comparator. Continuous conversion is accomplished by tying the EOC (end of conversion) output to the ST (start conversion) input. The SAR is reset on the positive edge of the ST pulse and initiates conversion on the falling edge. A reset is applied on power-up through U19B to the ST input, to ensure conversion start.

## Courtesy of http://BlackRadios.terryo.org

Table 3-18. Input Select Address Codes

| Selected | Address Line |  |  |
| :---: | :---: | :---: | :---: |
| Analog Channel | C | B | A |
| IN0 | 0 | 0 | 0 |
| IN1 | 0 | 0 | 1 |
| IN2 | 0 | 1 | 0 |
| IN3 | 0 | 1 | 1 |
| IN4 | 1 | 0 | 0 |
| IN5 | 1 | 0 | 1 |
| IN6 | 1 | 1 | 0 |
| IN7 | 1 | 1 | 1 |

All analog inputs are referenced to Vec ( +5 V ). The binary output is derived by the equation:

$$
\text { binary output }=\frac{\mathrm{v}^{\mathrm{IN}}}{\mathrm{~V}^{\mathrm{REF}}} \times 256
$$

Table 3-19 lists the analog inputs and their functions.

Table 3-19. U25 Analog Inputs and Functions

| Input | Function |
| :--- | :--- |
| IN0 | BFO Tuning Voltage |
| IN1 | Not Used |
| IN2 | Signal Strength |
| IN3 | Combined Audio |
| IN4 | 1st LO Tuning Voltage |
| IN5 | -15 V supply |
| IN6 | +15 V supply |
| IN7 | 2nd LO Tuning Voltage |

Clock circuit U18B and U18C provide timing and control of the A/D conversion rate. Circuit U18B is configured as an RC oscillator that provides an approximately 700 kHz clock that is conditioned by U18C before entering the A/D clock input.

### 3.2.5.12 Analog Inputs

IN0 is the input for the BFO tuning voltage. This voltage is tested to determine if the BFO synthesizer is locked. If unlocked, an error condition will occur in testing.

IN2, the signal strength input, is taken from the AGC circuit, inverted and amplified by U24A. Potentiometer R7 sets the multiplication ratio along with R8. Refer to paragraph 4.1.2.2 for R7 adjustment procedures. MAN signal streng'th is also passed through isolation amplifier U24B to protect the meter circuit from interfering with the AGC signal strength circuit.

## Courtesy of http://BlackRadios.terryo.org

## CIRCUIT DESCRIPTION

IN3, the audio monitoring circuit, creates a DC level from the audio for the A/D to convert. Resistor R19 provides isolation; C22 couples the AC; CR1 and CR2 half-wave rectify; and C14, R21, and R20 integrate the signal to make it an input level for the A/D.

IN4, the 1st LO tuning voltage input is tested to see if the first local oscillator synthesizer is locked. If unlocked, an error condition will occur in testing. Resistor R18 helps to pull down the tested voltage.

IN5 is the test point for the -15 V power supply. Voltage dividers R14 and R15 provide a positive reference, while CR3 protects the point from going negative at the input of the A/D.

IN6 is the test point for the +15 V power supply. Voltage dividers R16 and R17 provide a positive reference to the input of the A/D.

IN7 is the test point for the 2nd LO tuning voltage. No signal conditioning is necessary.

### 3.3 488M OPTION CIRCUIT DESCRIPTION

### 3.3.1 TYPE 796075 I/O INTERFACE (488M-A3)

The $488 \mathrm{M}-\mathrm{A} 3$ board interfaces the microprocessor in the interface section of the MFP Option with the remote control equipment. All board components are powered by +5 Vdc , applied across bypass capacitors C2 through C5. Refer to Figure 6-7, Type 796075 I/O Interface.

### 3.3.1.1 $\quad$ Address Latch (U6)

Integrated circuit U6 is an octal D-Type flip-flop. The clear (CLR) input is held high by Vec, causing information at the D inputs to be latched to the Q outputs on the positivegoing edge of the clock pulse. The clock pulse to $U 6$ is the microprocessor control line ALE, inverted by U8E. ALE goes low, and is inverted when a valid address has been placed on the data bus. The resultant positive-going clock pulse to U6 latches the lower-order address bits to the Q outputs. From the outputs of U6, four of the address bits are decoded as necessary to clock or enable the addressable circuits on the I/O board: U1, discussed in paragraph 3.3.1.6, and U4, discussed in paragraph 3.3.1.4. The three least significant bits of the address are described in paragraph 3.3.1.6.1.

### 3.3.1.2 Address Decoder (U7)

Integrated circuit U7 is a three-to-eight line decoder, enabled by one high and two low logic levels on the G inputs, as shown in Table 3-20. When U7 is enabled, the three select inputs at A, B, and C are decoded and one of the eight outputs goes low.

Table 3-20. U7 Truth Table

|  | Enable | Inputs | Select |  |  | Outputs |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| G1 | G2A | G2B | A | B | C | Y0 | Y1 | Y2 |
| H | L | L | L | L | L | L | H | H |
| H | L | L | L | L | H | H | L | H |
| H | L | L | L | H | L | H | H | L |

### 3.3.1.2.1 U7 Enable

The G1 enable input to U7 is the A12 data line. Table 3-21 lists the addressable devices on the I/O board. The A12 bit is high when the least significant bit of the MSD of the hexadecimal address is 1 (0001). G2B is the inverted (by U8A) level of the AD5 address bit. To provide the two low levels to enable U7, AD5 must be high and AD4 must be low. As shown in Table 3-21, the enabling conditions exist when one of the hexadecimal addresses 1020,1028 through 102 F , or 1060 has been placed on the data bus.

### 3.3.1.2.2 U7 Select

The select input A to U7 is the address bit carried on the AD3 data line. (Note that the three least significant bits of the address, carried on AD0 through AD2, are not decoder inputs. These bits have a unique use and are discussed in paragraph 3.3.1.6.1.

Three of the eight available outputs of U7 are used to enable the addressable devices on the I/O board. Comparison of Tables $3-20$ and $3-21$ shows that the Y1 output goes low when one of the eight U1 addresses is latched by U6, and Y0 goes low when the U4 address is latched, Y2 goes low when U10 is addressed.

### 3.3.1.3 Switch Assembly (S1)

The eight-position switch assembly (S1) is used, prior to remote operation, to establish receiver address. A high logic level indicates a closed switch, and a low logic level indicates an open switch. Paragraph 2.4.3.2, part of the preparation for 488 operation procedure, contains a description of the switch settings.

### 3.3.1.4 Tri-State Buffer/Inverter (U4)

Octal tri-state buffer U4 allows the microprocessor to read the settings of the S1 switch assembly. The buffer inputs the logic levels of the switch settings from the pull-up resistive network, U5. The outputs of U4 are tri-stated if the logic levels at the enable inputs (pins 1 and 19) are high. When the enable inputs go low, the outputs of U4 are taken out of tristate and the data from the S1 switch assembly are inverted and placed on the lower-order address/data bus lines. U 4 is enabled when the $\overline{\mathrm{RD}}$ signal from the microprocessor goes low (after the address is latched) and when U7Y0 is low. As discussed in paragraph 3.3.1.2.2, U7Y0 is low only when address 1020 is latehed by U6.

The microprocessor initiates a read operation to address 1020 (U4) and loads the equipment address into U 1 . The $488 \mathrm{M}-\mathrm{A} 3$ interface device (U1) is programmed by the microprocessor to determine if the address on incoming data matches the switch-established address.

Table 3-21. Addresses 1020, 1028 Through 102F, and 1060

| Address Bit |  |  |  |  |  | 2 |  | 2 | 2 | 2 | 2 |  |  | 2 | 2 | $2^{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Data Line | 15 | 14 | 13 |  | 11 | 10 | 9 | 8 | 7 | 6 | 5 | 4 | 3 | 2 | 1 | 0 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 0 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 0 | 1 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 0 | 1 | 1 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 0 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 0 | 1 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 1 | 0 |
| U1 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 1 | 1 |
| U4 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | - |
| U10 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 0 | 0 | 0 |
| NOTE: D |  |  |  |  | Data Lines 0-7 = AD0 - AD7 <br> Data Lines $8-15=$ A8 - A15 |  |  |  |  |  |  |  |  |  |  |  |

### 3.3.1.5 Transceiver Network (U2 and U3)

Bidirectional bus transceivers, U2 and U3, are octal transmitter/receivers which interface the TTL logic of U1 with the IEEE-488 Bus. Either the transmitter or receiver in each of the eight channels in both U2 and U3 is enabled, forcing the disabled channel to a high impedance state.

Transceiver U2 interfaces the DIO data lines. When the signal level at pin 1 is high, data on the DIO lines transfer from the D to B ports and, when low, from the B to D .

Transceiver U3 interfaces the handshake and management control signals. The levels on the TE (Talk Enable) and DC (Direction Control) inputs, as well as the bus signal ATN, determine the direction of signal flow. (See Table 2-18 for an explanation of the handshake and management control line abbreviations used in the following discussion.)

TE controls the direction of DAV, NDAC, and NRFD. When TE is high, DAV=T, NDAC $=$ R, and $N R F D=R$. ( $T=$ transmit; $R=$ receive.)

DC controls the direction of IFC, REN, SRQ, and ATN. When DC is high, IFC $=\mathrm{R}$, REN $=R, S R Q=T$, and $A T N=R$.

ATN is a normal transceiver channel that functions additionally as an internal direction control or talk enable for EOI, whenever the TE and DC inputs are in the same state.

### 3.3.1.6 GPIB (General Purpose Interface Bus), (U1)

The GPIB (IEEE-488) Interface, U1, is a peripheral device designed to interface the microprocessor in the receiver control block with the IEEE-488 bus. Functions of the GPIB Interface include data transfer, handshake protocol, talk/listen addressing procedures, device clearing and triggering, service requests, and both serial and parallel polling schemes. Refer to Figure 3-11, the GPIB Interface Block Diagram to correlate the following functional descriptions of U1 components.


Figure 3-11. GPIB Interface (U1) Block Diagram

# Courtesy of http://BlackRadios.terryo.org 

TABLE 3-22
WJ-8718A/MFP
TABLE 3-23

### 3.3.1.6.1 Read/Write Registers

The GPIB Interface structure includes eight registers that the microprocessor writes to, and eight registers that the processor reads from. One read register and one write register are used for the transfer of data. The remaining write registers are used by the microprocessor to establish operating characteristics of U1. The remaining read registers provide the processor with a monitor of bus and controller conditions.

The registers are accessed by the levels on the $\overline{\mathrm{CS}}, \overline{\mathrm{RD}}, \overline{\mathrm{WR}}$, and RS0 through RS2 lines as listed in Table 3-22. Table 3-23 lists the RS0 through RS2 codes and functions of the read/write registers. The RS0 through RS2 codes are established by the three least significant bits of the address on the microprocessor bus lines. The $\overline{C S}$ signal level is established by decoding other address bits, as described in paragraphs 3.3.1.1 and 3.3.1.2. The $\overline{\mathrm{CS}}$ signal goes low to enable U1 when a hexadecimal address between 1028 and 102F (Table 3-21) has been placed on the bus and latched by the ALE signal. The $\overline{\mathrm{RD}}$ or $\overline{\mathrm{WR}}$ signal is activated when the processor initiates a read or write operation to the addressed U1 register.

Table 3-22. Read/Write Register Address Codes

| Register | $\overline{\mathrm{CS}}$ | $\overline{\mathrm{RD}}$ | $\overline{\mathrm{WR}}$ | RS0 - RS2 |
| :--- | :---: | :---: | :---: | :---: |
| All read registers | 0 | 0 | 1 | Register Select Code |
| All write registers | 0 | 1 | 0 | Register Select Code |
| Don't care | 1 | x | x | xxx |

Table 3-23. Register Select Codes

| Read Registers | RS2 | RS1 | RS0 | Write Registers |
| :--- | :---: | :---: | :---: | :--- |
| Data-In | 0 | 0 | 0 | Data-Out |
| Interrupt Status 1 | 0 | 0 | 1 | Interrupt Mask 1 |
| Interrupt Status 2 | 0 | 1 | 0 | Interrupt Mask 2 |
| Serial Poll Status | 0 | 1 | 1 | Serial Poll Mode |
| Address Status | 1 | 0 | 0 | Address Mode |
| Command Pass Through | 1 | 0 | 1 | Auxiliary Mode |
| Address 0 | 1 | 1 | 0 | Address 0/1 |
| Address 1 | 1 | 1 | 1 | EOS |

The data-in register is used to move data from the 488 bus to the microprocessor in the receiver control block when the receiver is addressed to listen. The data-out register is used to move data onto the 488 bus. The GPIB Interface (U1) handles and initiates the required handshake protocol.

The write interrupt mask registers are configured by the microprocessor to select the events that will cause an interrupt signal to be generated by the corresponding bit in the interrupt status registers. The status registers are read by the microprocessor to determine which event has occurred so that the proper service routine can be executed.

# Courtesy of http://BlackRadios.terryo.org 

The address mode register is used by the microprocessor to configure the addressing mode, establishing the way that the information in the address 1 and 0 registers is used. The configuration of addressing is such that processor intervention is not required to recognize addressing by the controller. The address $0 / 1$ register determines the talker or listener functions. The address status register is used for monitoring the address state.

The remaining registers provide the means for the processor to establish and monitor U1 operating characteristics which include requests for service, decoding of incoming messages, loading of "hidden" U1 auxiliary registers, auxiliary commands, delays and finishes to handshake signals, internal time delays on the data lines, and parallel/serial poll protocol.

### 3.3.1.6.2 Reset Procedure

The GPIB Interface is reset to an initialization state by the microprocessor, either by a high level on the U1 RESET input line or by an auxiliary command written into the auxiliary mode register internal to U1. The reset state pulse or command is followed by initial condition configurations written to the interrupt mask, serial poll mode, address, EOS, and auxiliary registers. The initialization state is released by a microprocessor auxiliary command.

### 3.3.1.6.3 488 Handshake and Management Signals

The GPIB Interface implements all functions of the 488 handshake and management signals, described in paragraphs 2.4.5.3.2 and 2.4.5.3.3. The bidirectional handshake signals, $\overline{\mathrm{DAV}}, \overline{\mathrm{NRFD}}$, and $\overline{\mathrm{NDAC}}$, effect the transfer of data on the DIO lines. The $\overline{\mathrm{ATN}}, \overline{\mathrm{REN}}$, and $\overline{\mathrm{IFC}}$ signals originate from the controller, $\overline{\mathrm{SRQ}}$ is generated by the GPIB Interface, and $\overline{\mathrm{EOI}}$ is bidirectional. The $\overline{\text { ATN }}$ signal level specifies how data are to be interpreted. The IFC signal places the interface in a quiescent state and $\overline{\text { REN }}$ establishes remote control. The EOI signal indicates the end of a multiple byte transfer sequence. The GPIB Interface generates the SRQ signal to indicate the need for attention. All signals on the bus are specified as negative logic; however, on the U1 pins, logic is positive.

### 3.3.1.6.4 Direct Memory Access (DMA)

The $\overline{\text { DMA }} \overline{\mathrm{ACK}}$ is pulled up and not used in this design.

### 3.3.1,6.5 Data Ports

The bidirectional D0 through D7 data bus ports connect to the lower order microprocessor data bus. The ports input data from the receiver and output data from the controller. Direction of data transfer depends on the $\overline{R D}$ and $\overline{W R}$ signals.

The $\overline{\text { DI01 }}$ through $\overline{\text { DI08 }}$ ports transfer bidirectional data bytes between U1 and the external transceiver, U2 (See paragraph 3.3.1.5). The direction of transfer depends on the level of control signal TR/1.

### 3.3.1.6.6 Clock Input

A clock pulse is generated by the RC time constant of R1 and C1 and applied to the clock input. The time constant, coordinated with the feedback and inversion characteristics of U8B and F , provides square wave pulses of 1 MHz to a time delay generator internal to U .

### 3.3.1.6.7 Transmit/Receive Control Lines

The TR/1 signal is generated by U1 to control the operation of the external transceivers, U2 and U3 (paragraph 3.3.1.5). The TR/1 signal is set high to indicate output to the 488 bus. Data/signals are on the DI01 through DI08, EOI and DAV lines. Input (bus to U1) signals are on the NRFD and NDAC lines. The TR/1 signal is set low to indicate input signals on the DI01 through DI08, EOI and DAV lines, and output signals on the NRFD and NDAC lines. The TR/2 line is not used.

### 3.3.1.7 Master/Slave Hand Off (U9, U10)

When the receiver is programmed as a master device, it is necessary to control the ATN bus signal for sending commands over the bus. Register U10 is used to control the DC input of transceiver U3. The DC input determines the direction of the ATN transceiver.

To send a command out over the bus, the U10 register $\bar{Q}$ output is set to a low state to reverse the control of the ATN transceiver channel. By enabling its driver, the ATN input is then held low by R7 which sets ATN active on the GPIB. With ATN active, any data sent through the U2 data buffers is treated as a command instruction by any device connected to the master receiver. Resistors R5 and R6 keep REN and SRQ inactive, while the DC input is in the command state.

When address 1060 is placed on the bus and latched by U6, decoder U7 output is then enabled. During a $\overline{W R}$ bus cycle, the U9D output goes low. On the low-to-high transition at the end of the $\overline{\mathrm{WR}}$ cycle, the data (from AD0) on the D input of latch U10 is clocked into the Q output.

The U10 Q output is low during normal operation, allowing the normal ATN signal from U3 to be passed through U9B to U1.

U10 is reset on power-up by the inverted reset signal from U8D.

## 3.4 <br> 232M OPTION CIRCUIT DESCRIPTION

### 3.4.1 TYPE 796037 (ASYNCHRONOUS) I/O INTERFACE BOARD (232M-A3)

The (Asynchronous) I/O Interface Board ( $232 \mathrm{M}-\mathrm{A} 3$ ), when installed in an MFP equipped receiver, interfaces the microprocessor in the MFP Option with the remote control equipment. Figure 6-17 is the $232 \mathrm{M}-\mathrm{A} 3$ schematic diagram.

The majority of board components are powered by +5 Vdc , applied across bypass capacitors C1, C6, C7 and C8. The zener voltage reference circuit, formed by VR1, VR2, R1, R2, and C2 through C5, provides potentials of +12 and -12 Vdc for the U7 and U8 logic gates and +12 Vdc for the baud rate generator, U 5 .

### 3.4.1.1 $\quad$ Address Latch (U2)

Integrated circuit U2 is an octal D-Type flip-flop. The output enable input is grounded, causing information at the $D$ inputs to be latched to the Q outputs on the positivegoing edge of the clock pulse. The clock pulse to U 2 is the microprocessor control line ALE, inverted by U6B. ALE goes low (and is inverted), when a valid address has been placed on the data bus. The resultant positive-going clock pulse to U2 latches the lower-order address bits to the Q outputs (Q6 and Q7 are unused). From the outputs of U2, five of the address bits are decoded as necessary to clock or enable the two addressable circuits on the I/O board: U1, discussed in paragraph 3.4.1.7, and U3, discussed in paragraph 3.4.1.4. The LSB of the address is described in paragraph 3.4.1.7.5.

### 3.4.1.2 Address Decoder (U4)

U4 is a three-to-eight line decoder, enabled by one high and two low logic levels on the G inputs, as shown in Table 3-24. When U4 is enabled, the three select inputs at A, B, and C are decoded and one of the eight outputs goes low. In this application, only two outputs, Y0 and Y 1 , are used.

### 3.4.1.2.1 U4 Enable

The G1 enable input to U 4 is the bit that is carried on the A12 data line. Table 3-25 lists the addressable devices on the I/O board. The A12 bit is high when the least significant bit of the MSD of the hexadecimal address is 1 (0001). G2B is the inverted (by U6A) level of the AD5 addresss bit. To provide the two low levels to enable U4, AD5 must be high and AD4 must be low. As shown in Table 3-25, the enabling conditions exist when one of the hexadecimal addresses 1020,1021 , or 1022 has been placed on the data bus.

Table 3-24. U4 Truth Table

|  | Enable | Inputs | Select |  |  | Outputs |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| G1 | G2A | G2B | A | B | C | Y0 | Y1 |
| H | L | L | L | L | L | L | H |
| H | L | L | L | L | H | H | L |

Table 3-25. Addresses 1020, 1021, and 1022

| Address Bit |  | $2^{2}$ | 2 |  |  | ${ }^{3}$ |  | 2 | $2^{3}$ | 2 |  |  | $2^{0}$ |  |  |  | $2^{0}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Data Line | 15 | 14 | 13 |  |  | 10 | 9 | 8 | 7 | 6 | 5 |  | 4 | 3 | 2 | 1 | 0 |
| U1 | 0 | 0 | 0 |  |  | 0 | 0 | 0 | 0 | 0 | 1 |  | 0 | 0 | 0 | 0 | 0 |
| U1 |  | 0 | 0 |  |  | 0 | 0 | 0 | 0 | 0 | 1 |  | 0 | 0 | 0 | 0 | 1 |
| U3 |  | 0 | 0 |  |  | 0 | 0 | 0 | 0 | 0 | 1 |  | 0 | 0 | 0 | 1 | 0 |
| NOTE: |  |  |  |  | Data Lines 0-7 = AD0-AD7 <br> Data Lines $8-15=$ A8 - A15 |  |  |  |  |  |  |  |  |  |  |  |  |

### 3.4.1.2.2 U4 Seleet

The select inputs to U 4 at $\mathrm{A}, \mathrm{B}$, and C are the address bits carried on the AD1, AD2, and AD3 data lines. (Note that the least significant bit of the address, carried on AD0 is not a decoder input. This bit has a unique use and is discussed in paragraph 3.4.1.7.5.

Two of the eight available outputs of U4 are used to enable the two addressable devices (U1 and U3) on the I/O board. Comparison of Tables 3-24 and 3-25 shows that the Y0 output goes low when one of the two U1 addresses is latched by U2, and Y1 goes low when the U3 address is latched.

### 3.4.1.3 Switch Assembly (S2)

The eight-position switch assembly (S2) is used, prior to remote operation, to establish a receiver address for receiver master/slave configuration, and to enable parity check. If parity is enabled, the switch establishes odd or even parity. Table 2-2 lists the switch functions. A high logic level indicates a closed switch and a low logic level indicates an open switch.

Paragraph 2.4.2.3, part of the preparation for operation procedure, contains a description of the switch settings.

### 3.4.1.4 Tri-State Buffer/Inverter (U3)

Octal tri-state buffer U3 allows the microprocessor to read the settings of the S2 switch assembly. The buffer inputs the logic levels of the switch settings from the pull-up resistive network, U11. The outputs of U3 are tri-stated if the logic levels at the enable inputs (pins 1 and 19) are high. When the enable inputs go low, the outputs of U3 are taken out of tristate and the data from the S2 switch assembly are inverted and placed on the lower-order address/data bus lines. U3 is enabled when the $\overline{\mathrm{RD}}$ signal from the microprocessor goes low (after the address is latched) and when U4Y1 is low. As discussed in paragraph 3.4.1.2.2 U4Y1 is low only when address 1022 is latched by U2.

The microprocessor initiates a read operation to address 1022 (U3) to determine if the equipment address on incoming data is valid.

### 3.4.1.5 Baud Rate Network (U5, S1)

Baud rate is established by entering a baud rate select code in the four-position switch assembly, S1, described in paragraph 2.4.2.2. Baud rate generator U5 is an N-Channel LSI device capable of generating 16 selected frequencies (see Table 2-1) from a single input frequency.

The baud rate generator is basically a programmable 15 -stage feedback shift register capable of dividing any module up to 215 less 1 ( 32,767 ). The device is driven by external crystal Y1 with a resonating frequency of 5.0688 MHz . Pin functions are listed in Table 3-26.

Table 3-26. Baud Rate Generator Pin Functions

| Pin No. | Symbol | Name | Function |
| :---: | :--- | :--- | :--- |
| 1 | X1 | Crystal input 1 | One pin of Y1 |
| 2 | X2 | Crystal input 2 | One pin of Y1 |
| 3 | VCC | Power supply | +5 Vdc |
| 5 | GND | Ground | Common return |
| 8 | FO | Reference frequency | Reference output at <br>  <br> 9 |
|  | STB | Strobe | 1/4 F IN |
| A high logic level loads |  |  |  |
| $10-13$ | D, C, B, A | Input address | inputs A, B, C, and D <br> into U5. |
|  | Fogic levels select |  |  |
| 14 | F OUT | Output frequency | F OUT. <br>  |
|  |  | Frequency selected by |  |
| input address. |  |  |  |

### 3.4.1.6 Active Repeater Network

The J1 and J2 connections on the I/O board allow up to 32 receivers to be joined in a daisy chain and placed under control of one RS-232 compatible controller. J1 connects to the rear panel REMOTE INPUT and J2 connects to the rear panel MONITOR OUTPUT. In the daisy chain configuration, the monitor output of a receiver is connected to the REMOTE INPUT of the next receiver in the chain (see Figure 2-2). The remote control equipment connects to the REMOTE INPUT of the first receiver in the chain.

Signal and data transmissions are applied to each unit in the chain and are accepted by the addressed unit or units. Each I/O board actively repeats the transmissions, in both directions, assuring signal integrity.

Each handshake signal (RTS, DTR, DSR, and CTS) and each data bit (Rx and Tx data) passes through a repeater network consisting of a NAND gate circuit. The function of each line and the repeater network logic is discussed in the description of USART, U1 (paragraph 3.4.1.7).

### 3.4.1.7 Universal Synchronous/Asynchronous Receiver/Transmitter (USART, U1)

The USART (U1) is a peripheral device designed for data communications. The USART receives and transmits data between the microprocessor on the IF Interface Board and external control equipment. The functional format of the USART is programmed by the microprocessor prior to data transmission.

Internal to the USART, an 8-bit data bus coordinates the exchange of data between buffers, logic, and control circuits in response to control and handshake signals applied to the USART by external devices. Basically, the USART provides two services in response to the externally applied signals: it converts parallel data from the microprocessor to a serial data stream for transmission to the remote controller, or it converts a serial stream of data from the controller to a parallel format for transmission to the microprocessor. The USART generates signals which inform the microprocessor and the controller when it is ready to transmit or receive, ensuring the orderly transfer of data. Data are transmitted from the USART on the TxD line and received by the USART on the RxD line (paragraphs 3.4.1.7.14 and 3.4.1.7.15).

The activities of the USART are timed by the CLK, $\overline{\mathrm{RxC}}$, and $\overline{\mathrm{TXC}}$ inputs, and regulated by the logic levels on the control input lines (RESET, $\overline{C S}, C / \bar{D}, \overline{R D}$, and $\overline{W R}$ ) and the handshake input lines ( $\overline{\mathrm{CTS}}$ and $\overline{\mathrm{DSR}}$ ). The signals generated by the USART are Rx RDY and Tx RDY to the microprocessor, and $\overline{\mathrm{RTS}}$ and $\overline{\mathrm{DTR}}$ to the remote controller. The input and output signals are described in paragraphs 3.4.1.7.1 through 3.4.2.7.15.

### 3.4.1.7.1 RESET Signal

A high logic level on the RESET line resets the internal circuits of the USART, The RESET signal is generated by the microprocessor when power is applied. Following reset and prior to data transmission, the USART must be programmed by the microprocessor.

The programming procedure establishes general operating characteristics of the USART and must follow each USART reset.

### 3.4.1.7.2 Clock Input

The Clock (CLK) input is provided to establish the time base for internal activities of the USART. Clock rate is established by the baud rate generator, U5.

### 3.4.1,7.3 Receiver Clock/Transmitter Clock

The Receiver Clock ( $\overline{\mathrm{RxC}}$ ) and the Transmitter Clock ( $\overline{\mathrm{TxC}}$ ) establish the rate of data transmission. Characters are shifted out of the USART on the falling edge of $\overline{T x C}$. Input data are sampled on the rising edge of $\overline{\mathrm{RxC}}$. The rate of the two clock signals is established by coded entries to the S1 switch assembly (paragraph 3.4.1.5), whose 16 selectable rates are between 50 and 19,200 baud. $\overline{\mathrm{RxC}}$ and $\overline{\mathrm{TxC}}$ are 16 times the baud rate.

# Courtesy of http://BlackRadios.terryo.org 

### 3.4.1.7.4 Chip Enable Input (CS)

The $\overline{\mathrm{CS}}$ signal must go low to enable the USART. The signal is generated by decoder U4 (paragraph 3.4.1.2) when one of two USART addresses has been placed on the bus by the microprocessor and latched by U2 (paragraph 3.4.1.1). The two addresses, hexadecimal 1020 and 1021, differ only in the logic level of the LSB which determines the level of the C/ $\overline{\mathrm{D}}$ signal, described in the following paragraph.

### 3.4.1.7.5 Control/Data Signal (C/D)

The C/D line is logic high when the USART address 1021 is used. The microprocessor uses this address to indicate to the USART that the data to follow are mode or command instructions. Following a USART reset, the first control word the microprocessor must issue is a mode instruction which defines the general operating characteristics of the USART. The second control word is a command instruction which defines a status word used to control the operation of the USART.

The USART address 1020 places a logic low level on the least significant bit (the $\mathrm{C} / \overline{\mathrm{D}}$ line), and is used by the microprocessor to inform the USART that a read or write data transmission is to follow.

### 3.4.1.7.6 Read/Write Signals ( $\overline{\mathrm{RD}} / \overline{\mathrm{WR}}$ )

A $\overline{R D}$ or $\overline{W R}$ signal is pulled low by the microprocessor when a read or write operation is to be initiated at the addressed device. The $\overline{\mathrm{RD}}$ and $\overline{\mathrm{WR}}$ signals have no effect on the USART unless the $\overline{C S}$ (enable) signal is low. Refer to paragraph 3.2.3.4 and Figures 3-7 and 3-8 for information concerning read and write operations.

### 3.4.1.7.7 Data Input/Output Lines (D0 Through D7)

The D0 through D7 input/output ports of U1 are connected to the lower-order address/data lines on the microprocessor data bus. Data are loaded on the bus by the microprocessor and enter the D ports of U1 during a write operation. During a read operation, data are loaded on the bus from the D ports of the USART.

### 3.4.1.7.8 Receiver Ready (Rx RDY)

The Rx RDY signal, generated by the USART, goes high to inform the microprocessor that an entire character has been received from the remote controller and is ready to be fetched by the microprocessor. The microprocessor treats the Rx RDY signal as an interrupt request, halts its main program, and branches to a subroutine to execute the instructions there. During the course of these instructions, the data in the USART are read and the Rx RDY automatically resets (goes low).

### 3.4.1.7.9 Transmitter Ready (Tx RDY)

The Tx RDY signal is raised high by the USART to inform the microprocessor that the USART is ready to receive data from the microprocessor. The microprocessor interprets the Tx RDY signal as an interrupt request, branches from the main program to a subroutine, and executes the subroutine instructions. The Tx RDY signal automatically resets (goes low) when the microprocessor writes data into the USART. Data written into the USART are automatically transmitted via the USART as long as CTS is active (low).

### 3.4.1.7.10 Data Set Ready (DSR)

The $\overline{\mathrm{DSR}}$ signal into the USART goes low when the remote controller is ready to send data. The signal is applied to an active repeater network in each receiver under control. Figure 3-12 illustrates the flow of a signal ( DSR or CTS ) from the control equipment through circuits representative of the repeater networks in three receivers. Additional receivers (up to 32) could be connected to the receiver chain in the manner depicted in the illustration. The logic levels illustrated are typical of an activated $\overline{\mathrm{DSR}}$ signal; the opposite logic levels are the steady state condition.

### 3.4.1.7.11 Data Terminal Ready ( $\overline{\mathrm{DTR}}$ )

The $\overline{\mathrm{DTR}}$ line from the USART goes low to inform the remote controller that the USART is ready to receive. This signal will go low only from the addressed receiver (or receivers, if more than one receiver has the same address) and will be actively repeated by the I/O boards in the receivers along the path to the controller. Figure 3-12 illustrates a typical network of one controller and three receivers under control. The control signal into the controller represents the $\overline{\mathrm{DTR}}$ (or $\overline{\mathrm{RTS}}$ ) signal. Circuits shown are equivalent to the repeater networks. In the figure, the logic levels on the receiver-to-controller lines are steady-state conditions.

### 3.4.1.7.12 Clear to Send (CTS)

A low logic level on the $\overline{\text { CTS }}$ input to U1 tells the USART that the controller is ready to receive a transmission from the USART. The signal reaches the USART through circuits identical to the $\overline{\mathrm{DSR}}$ signal path, described in paragraph 3.4.1.7.10. The USART will not transmit data unless $\overline{\text { CTS }}$ is active (low).

### 3.4.1.7-13 Request to Send ( $\overline{\mathrm{RTS}}$ )

A low logic level on the $\overline{R T S}$ output from the USART notifies the controller that the USART is now ready to transmit. The signal reaches the controller through circuits identical to the circuits discussed in the $\overline{\mathrm{DTR}}$ signal description in paragraph 3.4.1.7.11.


Figure 3-12. Equivalent Circuits for RxD, TxD, and Handshake Signals

## Courtesy of http://BlackRadios.terryo.org

### 3.4.1.7.14 Received Data (RxD)

The serial data stream on the RxD input to the USART contains the instructions from the remote control equipment. The data word is an 11-bit character consisting of a start bit, a parity bit (if parity is enabled), eight data bits, and a stop bit. Data format is discussed in paragraph 2.4.5. Figure 3-12 illustrates a remote controller with three receivers under control. Circuits equivalent to the repeater networks in the receiver illustrate the flow of receive and transmit data through the receivers. The logic levels are based on the assumption that receiver 3 is the addressed unit. Logic gates in the repeater network invert the logic sense of received data between the USART and the RS-232/C bus. The gates also switch signal levels between TTL and 232 logic.

### 3.4.1.7.15 Transmitted Data (TxD)

The information on the TxD line is a serial data stream transmitted from the USART to the remote control equipment. The data comprises current receiver parameters as requested by the controller. See paragraph 2.4.5 for information about data format. Figure 3-12 illustrates the flow of transmit data through a controlled network of three receivers. The circuits in the illustration are representative of the repeater network which assures the integrity of the data signals. The logic gates in the repeater network invert the logic sense of transmitted data between the USART and the RS-232/C bus. The gates also switch signal levels between TTL and 232 logic.

## SECTION IV

MAINTENANCE

### 4.1 GENERAL

The maintenance information which follows supplements Section IV (Maintenance) in the WJ-8718 Series HF Receiver Instruction Manual. In that manual, when reference is made to a standard circuit board which has been replaced by an MFP Option board, the referenced circuit should be changed to the equivalent MFP board, as listed in Table 4-1.

Table 4-1. MFP Equivalent Circuits

| Standard |  |  | MFP Equivalent |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| Type | Ref. Desig. | Board | Type | Ref. Desig. | Board |
| 791828 | A6A2 | Front Panel Interconnect | 794308-2 | MFP-A3 | IF Interface |
| 791575 | A6A1 | Manual/Up/Down Counter | 794275-X | MFP-A4 | Synthesizer Interface |
| 791684 | A10 | Front Panel Control | 794310 | MFP-A1 | Front Panel Switch/ Encode Assembly |
| 791578 | A8 | Frequency Display | 794310 | MFP-A1 | Front Panel Switch/ Encode Assembly |
| 791874 | A7 | Manual Tuning Module | 794310 | MFP-A1 | Front Panel Switch/ Encode Assembly |
| 791827 | A9 | BFO Switch | 794310 | MFP-A1 | Front Panel Switch/ Encode Assembly |
| 34836 | A7U1 | Encoder Assembly | 791202-5 | MFP-A2 | Encoder Assembly |

### 4.1.1 PERFORMANCE TESTS

Performance tests should be utilized to confirm operating specifications during initial inspection, during periodic checks, during troubleshooting, and after repairs. A receiver equipped with the MFP Option can be tested in the manual mode, following the procedures outlined in the WJ-8718 Series HF Receiver Instruction Manual, which establishes the required receiver parameters for using the MFP operating procedures in paragraph 2.4.4 of this supplement. The MFP Option is programmed with a diagnostic Built-In Test Equipment (BITE) program which verifies the operation of the front panel switches and LEDs, as well as the operation of microprocessor circuits involved in the execution of the program. The BITE program is discussed in paragraph 4.1.1.1.

The MFP Option is also programmed with diagnostic Receiver BITE Tests which aid in the troubleshooting of the receiver's IF and Synthesizer Sections, $\pm 15$ power supplies, and the random access memory (RAM) associated with the microprocessor. The Receiver BITE Tests are discussed in paragraph 4.1.1.2.

# Courtesy of http://BlackRadios.terryo.org 

## NOTE


#### Abstract

The purpose of both diagnostic programs is for ascertaining receiver readiness prior to operation, and for serving as a troubleshooting starting point for the maintenance technician. They are not intended for use on initial receipt of equipment, to determine if a receiver is defective and subject to return. That decision rests with the maintenance technician, only after examining the receiver for loose connections and other probable causes for malfunction.


### 4.1.1.1 BITE Program

The BITE (Built-In Test Equipment) program is a diagnostic program which verifies the operation of the front panel switches and LEDs and, in the process, verifies operation of the microprocessor and the following microprocessor support circuits:

1. The microprocessor bus lines AD0 through AD7 and A8 through A14.
2. Microprocessor signals ALE, $\overline{\mathrm{RD}}$, and $\overline{\mathrm{WR}}$.
3. The $\overline{R A M} \overline{E N A B L E}$ line and the EPROM devices on the Synthesizer Interface (MFP-A4).

The receiver is placed in the BITE mode through the front panel keypad. The access procedure is outlined to provide a complete check of the front panel switches and LEDs. In some situations, such as troubleshooting, the maintenance technician may prefer to validate one or more selected switches. Once the BITE program mode has been established, the switches can be verified in any order or number.

### 4.1.1.1.1 Access To BITE Program

1. Energize the receiver under test and press the front panel LOCAL key.
2. Press three general purpose keypad buttons: 1, 7, and * (special function).
3. All LEDs should be illuminated and all 7-segment displays (except the BFO sign) should display the number eight.
4. Refer to Table 4-2 and verify the operation of each Group 1 switch on the front panel by pressing the key and noting the RC code displayed in the $100 \mathrm{~Hz}(\mathrm{R})$ and $10 \mathrm{~Hz}(\mathrm{C})$ digits $(00.000 \mathrm{RC}$ ) of the FREQUENCY/MHZ display. The RC code is derived from the position a switch occupies in the Y (row or R ) and X (Column or C ) planes of the switch matrix. Refer to paragraph 4.1.1.1.2 for a detailed discussion of the RC code.

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5. Refer to Table 4-3 and verify the operation of each Group 2 switch on the front panel by pressing the key and noting the RC code in the left ( R ) and right (C) digit of the MEM ADRS display. Switches can be verified in any order or number.
6. To take the receiver out of the BITE mode (and restore normal operation), press the CLEAR key.

Table 4-2. Group 1 Switch Codes


Table 4-3. Group 2 Switch Codes

| Switch | $\frac{\text { Code }}{\mathrm{R} \quad \mathrm{C}}$ | Switch | Code |
| :---: | :---: | :---: | :---: |
| AM | 00 | 6 | 23 |
| FM | 01 | 16 | 24 |
| USB | 02 | LOCAL | 31 |
| LSB | 03 | RECALL | 35 |
| ISB | 04 | REMOTE | 41 |
| CWV | 05 | STORE | 45 |
| CWF | 06 | LINE AUDIO | 50 |
| MGC | 10 | SIGNAL STR | 51 |
| SLOW | 11 | EXAM | 60 |
| FAST | 12 | HAND OFF | 61 |
| BFO (TUNE) | 15 | AUTO SCAN | 62 |
| . 3 | 20 | LOCK OUT | 65 |
| 1 | 21 | THRS |  |
| 3.2 | 22 | DWELL | 67 |
| $\text { NOTE: } \begin{aligned} & R \text { (Row Code) }=\text { MEM ADRS Display (MSD) } \\ & C \text { (Column Code) }=\text { MEM ADRS Display (LSD) } \end{aligned}$ |  |  |  |
|  |  |  |  |

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FIGURE 4-1

### 4.1.1.1.2 Switch Codes (RC)

The front panel is organized into two switch matrixes, as described in paragraph 3.1.2.2. The Group 1 switches are on the right side of the front panel, and the Group 2 switches are on the left. Figure 6-4 is the schematic diagram illustrating both switch matrixes.

The X and Y planes of each switch matrix are identified as rows ( Y ) and columns (X). Each switch can be identified by the unique position it occupies in a row and a column. Figure 4-1 illustrates a 2 by 2 switch matrix. The RC code is derived from the row and column position as shown in the Figure $4-1$ table. S1 is in Row 0, Column 0 and is coded 00 ; S2 is in Row 0, Column 1 and is coded 01; S 3 is in Row 1, Column 0 and is coded 10; S 4 is in Row 1, Column 1 and is coded 11. This same principle is used for coding the 4 by 5 Group 1 matrix and the 8 by 8 Group 2 matrix.

The BITE program can be used during troubleshooting to isolate switch problems to a single switch, a row, a column, or a group. Refer to paragraph 4.1 .5 for a discussion of switch troubleshooting.


Figure 4-1. RC Code Diagram

### 4.1.1.2 Receiver BITE Tests

The Receiver BITE Tests are seven tests that examine (in order) RAM, +15 V , $-15 \mathrm{~V}, 1$ st LO, 2nd LO, Bandwidth Selection, and BFO Tuning, producing an error code in the process. Table 4-4 lists these seven tests along with their error codes.

Table 4-4. Receiver BITE Tests

| Number | Test | Error Code |
| :---: | :--- | :---: |
| 1 | RAM | 0001 |
| 2 | +15 V | 0002 |
| 3 | -15 V | 0004 |
| 4 | 1st LO | 0008 |
| 5 | 2nd LO | 0016 |
| 6 | Bandwidth Selection | 0032 |
| 7 | BFO Tuning | 0064 |

The test in progress is displayed by number in the BFO OFS/THRS LVL display window. The error code is a binary-weighted code. Any detected errors are summed and displayed in the FREQUENCY/MHz display window at the end of the seventh test. (A blank FREQUENCY/MHz display indicates no errors.) If an error is detected, the program should be cleared and run again to confirm this error. A displayed error does not necessarily indicate that the receiver is defective. (It may be an indication of a loose board, for instance.) Therefore, the error code should be used solely as a troubleshooting starting point for the maintenance technician.

### 4.1.1.2.1 Access To Receiver BITE Tests

1. Energize the receiver under test and press the front panel LOCAL key.
2. Press three general purpose keypad buttons: 1, 8, and * (special function).
3. The tests run continually in order, and the test in progress is displayed by number (see Table 4-4) in the BFO OFS/THRS LVL display window.
4. Any detected errors (binary-weighted code) are summed and displayed in the FREQUENCY/MHz display window at the end of the seventh test.
5. If an error is detected, the program should be cleared and run again to verify this error.
6. To take the receiver out of the Receiver BITE Test mode (and restore normal operation), press the CLEAR key.

### 4.1.1.3 Remote Interface Test Procedure

The information in this paragraph applies only to MFP Option equipped receivers which are also equipped with an optional remote interface ( 232 M or 488 M ). The interface can be tested by following the test procedures in the WJ-8718 Series HF Receiver Instruction Manual, supplemented with the guidelines listed below. It is assumed that the receiver has been tested in the manual mode and found to be operational, that the appropriate remote control equipment has been properly connected to the receiver under test, and that the switches in the remote interface board have been configured as required by the conditions of remote operation (reference paragraph 2.4.2, for 232 M operation; or paragraph 2.4 .3 for 488 M operation).

1. Energize the receiver under test and depress the front panel REMOTE button.
2. Use remote control command instructions (paragraph 2.4.5) to establish the receiver parameters listed in the test procedure.
3. Proceed to carry out the test procedure instructions, using remote commands to change the receiver parameters as required during the test. If test results are satisfactory, the receiver is responding properly to the remote commands. If a problem becomes evident during a performance test procedure, refer to the troubleshooting section of the WJ-8718 Series HF Receiver Instruction Manual and to paragraph 4.1.3 of this supplement.
4. Initiate monitor operations and verify that the monitored data corresponds to the most recently commanded parameters.

## NOTE

When the receiver under test is in any sideband mode (ISB, LSB, or USB), the sideband filter is automatically selected and the IF bandwidth is approximately 3 kHz . To indicate this, the 3.2 kHz IF bandwidth is automatically displayed.
5. Verify that the receiver status can be monitored with the front panel LOCAL pushbutton engaged.

### 4.1.2 ALIGNMENT PROCEDURES

### 4.1.2.1 Variable Voltage (VR) Adjustment (MFP-A1A2)

Adjustable potentiometer R2 on the Front Panel Switchboard, external to voltage regulator U1, is adjusted to set the level of the variable voltage output (VR) at J2. The input to U1 is an unregulated 10 V . The VR output should be approximately 7 Vdc . The VR voltage drives TTL logic circuits (LED anode drivers) on the Front Panel Encode Board. If VR is too high, the drivers will overheat and some display segments may not illuminate. If this problem
occurs, R2 should be adjusted to decrease VR voltage until all segments are of the same intensity. To properly adjust R2, connect an oscilloscope (reference equipment list, WJ-8718 Series HF Receiver Instruction Manual) to J2, and adjust R2 for an oscilloscope display of approximately 7 Vde with a maximum ripple of $500 \mathrm{mV} \mathrm{p}-\mathrm{p} . \mathrm{R} 2$ is adjusted clockwise to increase voltage and counterclockwise to decrease voltage.

### 4.1.2.2 Scan Threshold Adjustment (R7)

Potentiometer R7 controls the range of the signal strength levels, so that they fall within the required specifications. Table 4-5 lists the threshold level codes and their corresponding signal strength levels in dBm.

Table 4-5. Threshold Level Codes

| Keypad Entry <br> Code | Threshold <br> Level |
| :---: | :---: |
| 1 | -110 dBm |
| 2 | -100 dBm |
| 3 | -90 dBm |
| 4 | -80 dBm |
| 5 | -70 dBm |
| 6 | -60 dBm |
| 7 | -50 dBm |
| 8 | -30 dBm |
| NOTE: $\pm 10 \mathrm{dBm}$ tolerance. |  |

To locate R7, consult Figure 5-10 (location of components) and Figure 6-5 (schematic diagram). Before alignment, the potentiometer should be turned completely counterclockwise. The scan threshold adjustment itself is done with the receiver in scan mode. To align R7 to the specified range, use the following procedures:

1. Connect the test equipment as shown in Figure 4-2 and energize all equipment.
2. Tune the signal generator to provide a $15 \mathrm{MHz}, \mathrm{CW}$ signal at -30 dBm .
3. Store the following receiver parameters in memory channel 87:

| Detection Mode | CW |
| :--- | :--- |
| Bandwidth | 16 kHz |
| Gain Mode | AGC FAST |
| Tuned Frequency | 14.9 MHz |
| Threshold | 8 |
| Dwell | 1 |

4. Store 15.1 MHz in channel 88 .

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5. Press 89 and then LOCK OUT, to lock out Sector B from the scan.
6. Press 87 and then AUTO SCAN, which establishes an $f_{1}$ to $f_{2}$ single sector scan. The FREQUENCY/MHz window will display frequencies between 14.9 MHz and 15.1 MHz .
7. Slowly adjust R7 clockwise until the FREQUENCY/MHz display stops and dwells on a signal for a dwell time of one second.
8. Press AUTO SCAN to terminate the scan mode.
9. Enter a threshold level code of 1 and press AUTO SCAN to restart scan.
10. First adjust the signal generator low, so that the FREQUENCY/MHz display continues scanning. Gradually increase the output of the signal generator, until the FREQUENCY/MHz display stops and dwells on the signal. Signal generator output level should correspond to the specified level in Table 4-5 ( $\pm 10 \mathrm{dBm}$ tolerance).
11. Press AUTO SCAN to terminate the scan mode.
12. Enter the next threshold level code to be verified.
13. Press AUTO SCAN and repeat comparison of the signal generator output to specified threshold level in Table 4-5.


Figure 4-2. Scan Threshold Adjustment Setup

### 4.1.3 GENERAL TROUBLESHOOTING PROCEDURES

The MFP Option consists mostly of integrated circuits (ICs); therefore, the maintenance technician should become proficient at isolating defective ICs. Paragraphs 4.1.4 through 4.1.7 contain specific information for troubleshooting the key functional areas of the controller: microprocessor system, front panel switches, audio signal path, and power supply. The following general information is provided to aid the technician during all troubleshooting efforts, and should be used in conjunction with the specific information in paragraphs 4.1.4 through 4.1.7.

### 4.1.3.1 Digital Test Equipment

In many cases, the most efficient method of testing and troubleshooting digital ICs is to use a Logic Troubleshooting Kit, containing probes for supplying and detecting digital voltage levels and pulses. A recommended set of probes, the Hewlett-Packard 5022A MultiFamily Logic Troubleshooting Kit, contains the following devices:

| 545A | Logic Probe |
| :--- | :--- |
| 546A | Logic Pulser |
| 547A | Current Tracer |
| 548A | Logic Clip |

### 4.1.3.1.4.

### 4.1.3.1.1 Logic Probe

The logic probe detects logic levels and pulses without the need for a voltmeter or oscilloscope. The lamp in the probe tip shows a bright light for a logic high, no light for a logic low, and a dim light for a floating or faulty logic level. When a pulse train of up to 50 MHz is detected, the lamp blinks at a rate of 10 Hz . Power for the probe is supplied by the unit under test.

### 4.1.3.1.2 Logic Pulser

The logic pulser probe provides pulses that are compatible with digital circuits. When the probe tip is placed against a circuit point, the probe senses the logic level present. When the pulse button is pressed, the circuit is briefly pulsed to the opposite state. By using the logic probe and logic pulser in combination, a known input can be supplied to an IC and the effect can be observed at the output.

### 4.1.3.1.3 Current Tracer

The current tracer probe senses the magnetic field produced by current flow in the circuit. It can be used to trace current from the source to the sink; therefore, it can help find short circuits, or determine if an IC output is actually sourcing current, and if it is, find out where the current is flowing.

### 4.1.3.1.4 Logic Clip

The logic clip indicator displays the logic level present at all of the terminals of an IC at the same time. The readout is 16 LEDs: the individual LED glows if the terminal is high, and remains dark if the terminal is low. When used in combination with the logic pulser, the simultaneous effect of input pulses on all of the terminals of the IC under test can be observed.

### 4.1.3.2 Symptoms of Integrated Circuit Failure

1. Open Input - The output of a digital IC that has an opened input (internal) will usually behave as if the input level is a static high. The output of this IC will not respond to input pulses.
2. Opened Output - The output level of a digital IC that has an opened output (internal) will not respond to input pulses and will usually assume a faulty level of approximately 1.5 volts. This faulty level will often have the same effect as a high level on the input of the next IC in the circuit.
3. Short Circuit between Input and Ground or Vee - If the input is shorted to ground, the device will react as though the input is low. If the input is shorted to Vce, the device will react as though the input is high. Remember that either of these conditions can affect the output of the stage prior to this. A digital voltmeter (DVM) or oscilloscope can be used to distinguish between normal logic levels, and ground or Vec. The logic pulser cannot override the Vee or ground potential.
4. Short Circuit between Output and Ground or Vec - As in the above cases, the output will not respond to pulses at the input. A DVM can be used to distinguish between normal logic levels, and ground or Vec.
5. Short Circuit between Two Inputs - This failure is not as easy to detect. The device will behave normally unless one input tries to go high while the other is low. If this occurs, the short circuit will hold both inputs low.
6. Internal Failure of Logic Circuits - The effect of this trouble is to lock the output in either the high or low state. The output will not respond to input pulses.

### 4.1.3.3 In-Circuit Tests

Digital ICs rarely have partial or intermittent failures; thus, failures can usually be located by simple static tests. The circuit troubleshooting procedure will normally isolate a failure down to the few ICs that could be involved in the circuit failure. Checking the various nodes (branches common to one or more inputs and one or more outputs) is part of this procedure. The following series of steps will serve to locate most faulty ICs.

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1. Refer to Section III of this manual for detailed discussions of ICs, included in the appropriate detailed circuit descriptions. The schematic diagrams in Section VI provide pin diagrams.
2. Use the logic pulser to stimulate the inputs, while using the logic probe or logic clip to observe the response at the outputs. Make a note of each node that fails to respond to the stimulus, in the manner predicted by the truth table. The logic clip is especially useful for counters. (The term node is used to emphasize the fact that the failure of an IC to respond properly may be due to short-circuited tracks, or other failures external to the IC under test.)

## NOTE

When an IC fails, it usually fails completely. Thus, it is not normally necessary to observe the timing of the output signals. If an IC performs correctly under static test, it can be assumed to be good.
3. When one or more failing nodes have been located, use the logic probe to check the output of the IC that is driving the node. The logic probe will indicate a faulty level by showing a dim light. If a faulty level is indicated, the IC driving the node should be replaced.
4. Check for a short circuit to ground or Vee by placing both the logic probe and logic pulser on the terminal. The logic pulser can override normal TTL output levels, but cannot override power supply potentials. Therefore, the presence of a pulse indicates that the node is not short-circuited, and the absence of a pulse indicates the node is shorted to Vee (if it is high) or to ground (if it is low). If a short circuit is indicated, proceed to step 5. If no short circuit to ground or Vec is indicated, proceed to step 6.
5. A node short-circuited to ground or Vec could be the result of either a short-circuited track on the board or an internal short in an IC. First, examine all branches of track connected to the node with the current tracer. Use the current tracer to isolate a defective IC: the circuit that draws the most current is generally defective.
6. If neither an open output (step 3), nor a node shorted to ground or Vee (step 4) is indicated, look for a short between two nodes. This can be done more easily by turning OFF the unit's power, and using an ohmmeter to test for continuity between nodes that should not be common. (Refer to the schematic diagrams.) If a short circuit between nodes is indicated, proceed to step 7. If a short circuit is not evident, proceed to step 8.
7. The most probable cause of a short circuit between two nodes is a solder bridge on the board. If the two short-circuited nodes are connected to the same IC, the short circuit might be

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internal to the IC. If no external short circuit can be found, replace the common IC.
8. When open outputs and short circuits have been eliminated, the problem could be an open input on the IC being driven from the node, an internal IC failure, or an open signal path. The current tracer, or an ohmmeter, can be useful in finding an open signal path. For an opened input or an internal failure, the IC must be replaced.

### 4.1.4 MICROPROCESSOR SYSTEM TROUBLESHOOTING

### 4.1.4.1 Modular Concept

The microprocessor system consists of a number of modules, parallel-connected to the microprocessor data bus. A module can be described as a block of functionally related components capable of transmitting or receiving data on the bus. The IF Interface (MFP-A3), the Synthesizer Interface (MFP-A4), and the Front Panel Encode Board (MFP-A1A1) each contain modules which are a part of the microprocessor system. If an optional remote interface is in use, the interface board is also a part of the modular structure. The practical approach to troubleshooting the system is to isolate the trouble to a specific circuit board and then to a specific module.

Each module typically consists of address decode and functional circuitry. In order to troubleshoot within the module, the technician must be able to locate a module and identify the address and function circuits.

The microprocessor accesses each module in the system for a data transfer by a unique hexadecimal address. The address is placed on the bus at the beginning of each data transfer cycle. Typically, the address decode components in a module provide the following two services:

1. Because the eight lower-order bits of the data bus are multiplexed, the address data cannot remain on the bus throughout the entire data transfer. Therefore, a circuit is required which will capture and hold the address data. The typical address capture circuit in the MFP system is a D-Type flip-flop which latches the address information on the negative-going transition of a special microprocessor signal (ALE).
2. The address information must be decoded by a circuit which can provide clock or toggle type output signals. Generally, address decode circuitry is shared by more than one module. The decoder must utilize the address data inputs to provide one unique toggle output for each address input. The address decode circuits in the system are typically BCD-to-decimal integrated circuits with select and enable inputs tied directly or through logic gates to the address capture circuit, to the microprocessor data bus, or to selected processor signals.

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The functional portion of a module is typically fed by the microprocessor data bus and is placed on or off the bus by a toggle-type clock or enabling signal from the address decode circuits. It is the enabling signal which causes the function circuit to appear visible or invisible to the microprocessor. In the disabled state, the function circuit is an open circuit. The function circuitry can be a logic gating network, a memory device, or a complex chip designed for a specific purpose. The circuitry often consists of a primary-purpose integrated circuit with one or more support circuits. In some cases, the primary function circuit is always on and must be interfaced to the bus by an external tri-state buffer which guarantees that the processor looks at the circuit at the proper time. Other primary function ICs contain internal tri-state buffers.

The modular concept simplifies troubleshooting by providing a pattern of similarity in circuit characteristics throughout the microprocessor system, allowing application of generalities which apply to each module. These generalities are listed below.

1. During the execution of its program, the microprocessor communicates with each module on the bus, enabling one functional circuit at a time through address capture/decode circuits. If an enable signal is held high throughout the run of the program, the microprocessor cannot communicate with the module. If the signal is held low, the module will always be on the bus, receiving data intended for other devices. A logical approach to modular troubleshooting is to verify that the enabling outputs of the address capture/decode network are toggling.
2. If the address decode outputs do not toggle, the enable or clock inputs to the address decode and capture ICs should be verified. These inputs are typically gated combinations of the ALE, $\overline{\mathrm{RD}}$, and $\overline{\mathrm{WR}}$ microprocessor signals and selected upper-order address lines. Absence of the proper levels on the clock or enabling inputs is often the result of defects in the solder connections or logic gates in the path of the signals. Successful execution of the BITE program (paragraph 4.1.1.1) verifies that the $\overline{\mathrm{RD}}, \overline{\mathrm{WR}}, \mathrm{ALE}$, and upper order data lines are operational.
3. The BITE program also verifies that the AD0 through AD7 address/ data lines are functional, and it can be assumed that the address capture IC is receiving address information. Troubleshooting efforts should next be directed toward the address capture and decode ICs.
4. If the functional IC enable input is toggling, but the IC function is not operational, support ICs should be examined. Support ICs are integrated circuits which perform services for the functional circuit. Services are varied and include signal level inversion, generation of timing signals, buffering, and multiplexing.
5. If a functional IC is mounted on sockets, the circuit can be replaced by an identical IC known to be operational. Observation of changes in operating characteristics will aid in isolation of a defective IC.

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6. If an IC in a particular location repeatedly requires replacement, the conditions external to the IC should be checked: resistors, Vec, ground connections, solder joints, and connections to unused sections of the ICs.
7. The information in paragraph 4.1.3.2, Symptoms of Integrated Circuit Failure, can be referenced to aid in locating defective ICs.

### 4.1.4.2 IF Interface Modules (MFP-A3)

There are four IF Interface modules to support the microprocessor system. Troubleshooting is indicated when a failure occurs in an area of operation within the functional boundaries of one of the modules. Primarily, these modules serve to store data associated with the IF bandwidth, gain mode, detection mode and $A / D$ conversion for monitoring receiver operations. The data is utilized by the signal processing circuits in the receiver, and is used for switching the RF gain voltage and audios. If the receiver is equipped for optional remote operation, board functions include analog-to-digital conversion of signal strength voltage and digital-to-analog conversion of remote RF GAIN voltage. The board also contains modules which when used with a standard front panel, interface the microprocessor with front panel components through J2, J3, and J4. In this application, front panel interface is accomplished through the Front Panel Encode Board (MFP-A1A1). The front panel interface modules and output jacks are unused.

Tables 4-6 and 4-7 list all information pertinent to the modular structure. The tables are to be used in conjunction with the information in paragraph 4.1.4.2.1.

Table 4-6. IF Interface Modules

| Data | Module |  |  |  |
| :--- | :--- | :--- | :--- | :--- |
|  | 1 |  | 3 | 4 |
| Description | D/A: RF gain | Programmable | IF register (1) | IF register (2) |
|  |  | A/D |  |  |
| Functional Circuit | U23 | U25 | U5 | U9 |
| Tri-State Buffer | NA | Internal (U25) | Internal (U5) | Internal (U9) |
| Buffer Clock | NA | NA | NA | NA |
| Address Capture Latch | U8 | NA | U8 | U8 |
| Address Capture Clock | U10B, pin 6 | U19C, pin 8 | U10B, pin 6 | U10B, pin 6 |
| Address Decode | U11, pin 12 | Internal (U25) | U11, pin 14 | U11, pin 13 |

Table 4-7. IF Interface Support Circuits

| Support Circuit | Supports | Function |
| :---: | :---: | :---: |
| U1 <br> U2 <br> U3 <br> U4 <br> U6 <br> U7 <br> U10A <br> U10C <br> U10D <br> U12 <br> U13A <br> U13B <br> U13C <br> U13D <br> U14A <br> U14B <br> U14C <br> U14D <br> U14E <br> U14F <br> U15A <br> U15B <br> U16A <br> U16B <br> U16C <br> U17 <br> U18A <br> U18B <br> U18C <br> U18D <br> U18E <br> U18F <br> U19A <br> U19B <br> U19D | ```* * * * * * U12 NU U11 ** U4* U20 U2*, U17* U1* U19A NU U25 U16 NU U10D U20 U11, U12 Receiver audio circuitry Receiver audio circuitry Receiver BFO circuitry * NU U25 U25 U17* U17* U17* U20 U25 U25``` | Address encode <br> Address encode <br> Transceiver direction control <br> Inverter <br> Inverter <br> Inverter <br> Inverter <br> Transceiver direction control $\overline{\mathrm{RD}}, \overline{\mathrm{WR}}$ Buffer <br> Audio squelch switch <br> Combined audio squelch switch <br> BFO switch (used only on a standard front <br> panel receiver) <br> RC oscillator circuit <br> RC oscillator circuit <br> Transceiver direction control Conversion start control gate <br> Output enable control gate |
| NOTE: 1. <br> 2. <br> 3. | notes not used. ponents belong iver front panel ves to maintain | dules used only when a standard use. rity of the board. |

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TABLE 4-7

Table 4-7. IF Interface Support Circuits (Cont'd)

| Support Circuit | Supports | Function |
| :---: | :---: | :---: |
| U20 <br> U21A <br> U21B <br> U22 <br> U24A <br> U24B <br> U24C <br> U24D | U1*-U4*, U5, U6* U7*, U8, U9, U21 U23, U25 U16A, U16B U22 U16A, U16B U25 U25 U22 Manual gain circuitry | Bus transceiver <br> Select level register <br> Remote/local latch <br> Audio gain switch <br> MGC signal strength amplifier <br> AGC signal strength amplifier <br> Current-to-voltage converter amplifier <br> RF gain amplifier |
| NOTE: 1. NU denotes not used. <br> 2. * Components belong to modules used only when a standard receiver front panel is in use. <br> 3. ** Serves to maintain integrity of the board. |  |  |

4.1.4.2.1 IF Interface Troubleshooting Procedure

1. Reference the following data to support Tables 4-8 and 4-9.
a. Circuit descriptions in paragraph 3.2.5.
b. Location of Components Diagram, Figure 5-10.
c. Schematic Diagram, Figure 6-5.
d. Modular troubleshooting concept described in paragraph 4.1.4.1
2. Recommended test equipment for troubleshooting IF Interface components are an oscilloscope, digital voltmeter, ohmmeter (reference the equipment list in the WJ-8718 Series HF Receiver Instruction Manual), and general purpose digital troubleshooting equipment described in paragraph 4.1.3.1 of this supplement.
3. The adjustable potentiometer R7 is unique to the IF Interface modular structure and should be considered during troubleshooting efforts. Potentiometer R7 controls the range of the signal strength levels, so that they fall within the required specifications and is discussed in paragraph 4.1.2.2.

### 4.1.4.3 Synthesizer Interface Modules (MFP-A4)

The Synthesizer Interface contains the microprocessor and a power-up network to reset the processor. Proper operation of the microprocessor can be validated by execution of the BITE program, described in paragraph 4.1.1.1. The Synthesizer Interface contains six modules which interface the microprocessor with the RF and BFO Frequency Synthesizers, one RAM module which stores transitory data, and two EPROM modules which store the microprocessor instruction set.

Tables 4-8 through 4-11 list the Synthesizer Interface modules, while Table 4-12 lists the support ICs. The tables are to be used in conjunction with the information in paragraph 4.1.4.3.1 to troubleshoot the Synthesizer Interface.

### 4.1.4.3.1 Synthesizer Interface Troubleshooting Procedure

1. Reference the following manual data to support Tables 4-8 through 4-12.
a. Circuit descriptions in paragraph 3.2.4.
b. Location of Components Diagram, Figure 5-11.
c. Schematic Diagram, Figure 6-6.
d. Modular troubleshooting concept, described in paragraph 4.1.4.1.
2. Execution of the BITE program will effectively verify the operation of the EPROMS on the Synthesizer Interface.
3. Recommended test equipment for Synthesizer Interface troubleshooting to a component level consists of the general purpose digital test equipment described in paragraph 4.1.3.1 of this supplement, as well as an oscilloscope, digital voltmeter, and ohmmeter (refer to the equipment list in the WJ-8718 Series HF Receiver Instruction Manual).

Table 4-8. Synthesizer Interface Modules (EPROM)

| Data | Module |  |
| :--- | :--- | :--- |
|  | 1 | 2 |
| Description | Program memory | Program memory |
| Functional Circuit | U1 | U2 |
| Tri-State Buffer | Internal (U1) | Internal (U2) |
| Buffer Clock | NA | NA |
| Data Capture Latch | NA | NA |
| Data Capture Clock | NA | NA |
| Address Capture Latch | U6 | U6 |
| Address Capture Clock | U9C, pin 8 | U9C, pin 8 |
| Address Decode | U7, pin 15 | U7, pin 11 |
| NOTE: NA denotes not applicable. |  |  |

Table 4-9. Synthesizer Interface Module (RAM)

| Data | Module |
| :--- | :--- |
|  | 3 |
| Description | Storage memory |
| Functional Circuit | U3 |
| Tri-State Buffer | Internal (U3) |
| Buffer Clock | NA |
| Data Capture Latch | NA |
| Data Capture Clock | NA |
| Address Capture Latch | U6 |
| Address Capture Clock | U9C, pin 8 |
| Address Decode | U10D, pin 11 |
| NOTE: NA denotes not applicable. |  |

Table 4-10. Synthesizer Interface Modules (RF Frequency)

| Data | Module |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | 4 | 5 | 6 | 7 |
| Description | RF frequency register | RF frequency register | RF frequency register | RF frequency register |
| Functional Circuit | U14 | U15 | U16 | U17 |
| Tri-State Buffer | Internal (U14) | Internal (U15) | Internal (U16) | Internal (U17) |
| Buffer Clock | NÁ | NA | NA | NA |
| Data Capture Latch | NA | NA | NA | NA |
| Data Capture Clock | NA | NA | NA | NA |
| Address Capture Latch | U5 | U5 | U5 | U5 |
| Address Capture Clock | U9C, pin 8 | U9C, pin 8 | U9C, pin 8 | U9C, pin 8 |
| Address Decode | U8, pin 13 | U8, pin 14 | U8, pin 15 | U8, pin 10 |
| NOTE: NA denotes n | t applicable. |  |  |  |

Table 4-11. Synthesizer Interface Modules (BFO Frequency)

| Data | Module |  |
| :--- | :--- | :--- |
|  |  <br> Description <br>  <br> Functional Circuit <br> Tri-State Buffer <br> BFO sign, RF/BFO frequency | BFO frequency register |
| Buffer Clock | register | U12 |
| Data Capture Latch | Internal (U12) | U13 |
| Data Capture Clock | NA | Internal (U13) |
| Address Capture Latch | NA | NA |
| Address Capture Clock | NA | NA |
| Address Decode | U9C, pin 8 | NA |
|  | U8, pin 11 | U5 |
| NOTE: NA denotes not applicable. | U9C, pin 8 |  |

Table 4-12. Synthesizer Interface Support Circuits

| Support Circuits | Supports | Function |
| :---: | :--- | :--- |
| U4 | U18* | Bus transceiver |
| U9A | U8 | WR buffer |
| U9B | U8 | Address decode |
| U9D | U4 | Transceiver direction control |
| U10A | U18* | RD line buffer |
| U10B | U18* | ALE line buffer |
| U10C | U18* | WR line buffer |
| U11A | U18* | Power up/down circuit |
| U11B | U18* | RC oscillator circuit |
| U11C | U18* | RC oscillator circuit |
| U11D | U18* | RC oscillator circuit |
| U11E | U18* | RC oscillator circuit |
|  | U18* | Power up/down circuit |

### 4.1.4.4 Front Panel Switch/Encode Assembly Modules (MFP-A1)

The Front Panel Switch/Encode Assembly consists of two subassemblies, interconnecting track, and amplifiers for the front panel phones jack. The two-channel amplifiers are discussed in paragraph 4.1.6, Audio Signal Troubleshooting. The following discussion pertains to the Front Panel Encode Board (MFP-A1A1) which contains the modular units associated with the microprocessor system. Troubleshooting the second A7 subassembly, the Front Panel Switchboard, is discussed in paragraph 4.1.5.

The Front Panel Encode Board interfaces the front panel switches, LED indicators, and LED displays with the microprocessor. The board contains two symmetrical modules which interface the two front panel switch matrixes, identified as Group 1 and Group 2 (refer to paragraph 3.1-2.2). Modules 3 and 4 interface the rotation of the front panel tuning wheel and the LINE AUDIO/SIGNAL STR meter functions with the microprocessor. Tables 4-13 and 4-14 provide all pertinent data concerning the Front Panel Encode Board modules. The tables should be used in conjunction with the information in paragraph 4.1.4.4.1.

Troubleshooting the Front Panel Encode Board is required when switches with common circuits on the encode board are malfunctioning.

### 4.1.4.4.1 Front Panel Encode Troubleshooting Procedure

1. Reference the following manual data to support Tables 4-13 and 4-14.
a. Circuit descriptions in paragraph 3.1.2.
b. Location of Components Diagram, Figure 5-7.
c. Schematic Diagram, Figure 6-3.
d. Modular troubleshooting concept, described in
paragraph 4.1.4.1.
2. The BITE program can be used to verify the operation of the Front Panel Encode Board or to isolate a problem to a section of the board. By referencing Tables 4-13 and 4-14, common circuits to Group 1 or Group 2 switches and LEDs can be noted, allowing a problem to be isolated to a defective module or to a defective IC within a module. Paragraph 4.1.5.1 contains an evaluation of the BITE program, relative to the front panel switches, LEDs, and logic circuits.
3. The symmetry of the Group 1 and Group 2 modules allows switching of the socket-mounted ICs, providing a simplified means of isolating defective circuits.
4. During operation, the SL0 through SL3 outputs of U1 and U2, and the outputs of U4, U5, U6, U8, U9, and U10 should always be toggling. If an output is stuck high or low, a problem with the IC, solder connections, Vec, or ground is evident.

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5. The general purpose digital test equipment described in paragraph 4.1.3.1 of this supplement, and the oscilloscope, digital voltmeter, and ohmmeter recommended in the WJ-8718 Series HF Receiver Instruction Manual, are the equipment required for MFP-A1A1 troubleshooting.

Table 4-13. Front Panel Encode Modules

| Data | Module |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  | 1 | 2 | 3 | 4 |
| Description | Group 1 keyboard/display | Group 2 keyboard/display | Tuning wheel | Meter switch |
| Functional Circuit | U1 | U2 | U17 | U14 |
| Tri-State Buffer | Internal (U1) | Internal (U2) | Internal (U17) | NA |
| Buffer Clock | U19, pin 14 | U19, pin 13 | U19, pin 15 | NA |
| Data Capture Latch | NA | NA | NA | U20 |
| Data Capture Clock | NA | NA | NA | U16, pin 11 |
| Address Capture Latch | U18 | U18 | U18 | U18 |
| Address Capture Clock | U13, pin 2 | U13, pin 2 | U13, pin 2 | U13, pin 2 |
| Address Decode | U19 | U19 | U19 | U19 |

Table 4-14. Front Panel Encode Support Circuits

| Support Circuit | Supports | Function |
| :---: | :---: | :---: |
| U3 | U1 | BCD-to-7 segment decoder (Group 1) |
| U4 | U1 | Cathode counter multiplexer (Group 1) |
| U5 | U1 | Cathode driver (Group 1) |
| U6 | U1 | Cathode driver (Group 1) |
| U7 | U2 | BCD-to-7 segment decoder (Group 2) |
| U8 | U2 | Cathode counter multiplexer (Group 2) |
| U9 | U2 | Cathode driver (Group 2) |
| U10 | U2 | Cathode driver (Group 2) |
| U11 | U1 | Column multiplexer (Group 1) |
| U12 | U2 | Column multiplexer (Group 2) |
| U13A | U18 | ALE inverter |
| U13B | NU |  |
| U13C | NU |  |
| U13D | NU |  |
| U13E | U1, U2 | Clock to Modules 1 and 2 Clock to Modules 1 and 2 |
| U13F | U1, U2 | Clock to Modules 1 and 2 |

Table 4-14. Front Panel Encode Support Circuits (Cont'd)

| Support Circuit | Supports | Function |
| :--- | :--- | :--- |
| U15A | U1, U2 | RESET buffer |
| U15B | U19 | Address decode for U19 |
| U15C | U1, U2, U17 | RD buffer |
| U15D | U1, U2, U16 | WR buffer |
| U16A | NU |  |
| U16B | NU |  |
| U16C | NU | CLK decode |
| U16D | U20 |  |
| U21A | NU | Tuning wheel CLK inverter |
| U21B | NU | CLR inverter |
| U21C | U7 | Tuning wheel DIR inverter |
| U21D | U17 |  |
| U21E | U17 | NU |

### 4.1.4.5 Optional Asynchronous I/O Board Modules (232M-A3)

The Asynchronous I/O Board (optional) is primarily responsible for the transfer of data between equipment via the RS-232/C interface. Troubleshooting is required when the receiver appears inoperative during remote operation only. The card contains two modules to support the microprocessor system. All pertinent information for the modules is given in Tables 4-15 and 4-16. The tables are to be used in conjunction with the information given in paragraph 4.1.4.5.1.

Table 4-15. Asynchronous I/O Board Modules (232M-A3)

|  | Module |  |
| :---: | :---: | :---: |
| Data | 1 | 2 |
| Description | Receiver address/ parity/master slave switch | USART |
| Functional Circuit | S2 | U1 |
| Tri-State Buffer | U3 | Internal (U1) |
| Buffer Clock | U4, pin 14 | U4, pin 15 |
| Data Capture Latch | NA | NA |
| Data Capture Clock | NA | NA |
| Address Capture Latch | U2 | U2 |
| Address Capture Clock | U6B, pin 6 | U6B, pin 6 |
| Address Decode |  |  |
| NOTE: NA indicates not | icable. |  |

### 4.1.4.5.1 Asynchronous I/O Board Troubleshooting Procedure

1. Reference the following manual data to support Tables 4-15 and 4-16.
a. Circuit descriptions in paragraph 3.4.1.
b. Location of Components Diagram in Figure 5-13.
c. Schematic Diagram, Figure 6-8.
d. Modular troubleshooting concept, described in paragraph 4.1.4.1.
2. Recommended test equipment for troubleshooting Asynchronous I/O Board components are an oscilloscope, digital voltmeter, ohmmeter (reference the equipment list in the WJ-8718 Series HF Receiver Instruction Manual), and the general digital troubleshooting equipment described in paragraph 4.1.3.1 of this supplement.
3. The following information concerns characteristics unique to the Asynchronous I/O Board and pertinent to troubleshooting efforts.

Table 4-16. Asynchronous I/O Board Support Circuits (232M-A3)

| Support Circuit | Supports | Function |
| :--- | :---: | :--- |
| U5 | U1 | Baud rate generator |
| U6A | U4 | Inverts output of U2 |
| U6B | U2 | ALE inverter |
| U7A | U1 | Converts RTS from TTL to 232/C level |
| U7B | U1 | DSR buffer |
| U7C | U1 | Converts TxD from TTL to 232/C level |
| U8A | U1 | Converts DTR from TTL to 232/C level |
| U8B | U1 | CTS buffer |
| U8C | U1 | RxD buffer |
| U9A | U1 | Converts RxD from 232/C to TTL level |
| U9B | U1 | Converts DSR from 232/C to TTL level |
| U9C | U1 | Converts CTS from 232/C to TTL level |
| U10A | U7A | RTS buffer |
| U10B | U7C | TxD buffer |
| U11 | U8A | DTR buffer |

### 4.1.4.6 Optional I/O Interface Modules (488M-A3)

The I/O Interface is primarily responsible for data transfer via the IEEE Standard 488-1975 bus when the WJ-8718A HF Receiver is equipped with the MFP and 488M Options

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TABLE 4-17
and is connected to a remote control device compatible with the 488 bus. Troubleshooting the I/O Interface may be required when the receiver appears inoperative only during remote operation. The interface contains two modules to support the microprocessor system. All pertinent information for the modules is given in Tables 4-17 and 4-18. The tables are to be used in conjunction with the information in paragraph 4.1.4.6.1.

Table 4-17. I/O Interface Modules (488M-A3)

|  | Module |  |
| :---: | :---: | :---: |
| Data | 1 | 2 |
| Description | Receiver address switch | GPIB interface |
| Functional Circuit | S1 | U1 |
| Tri-State Buffer | U4 | Internal (U1) |
| Buffer Clock | U7, pin 15 | U7, pin 14 |
| Data Capture Latch | NA | NA |
| Data Capture Clock | NA | NA |
| Address Capture Latch | U6 | U6 |
| Address Capture Clock | U8E, pin 10 | U8E, pin 10 |
| Address Decode | U7 |  |
| NOTE: NA denotes not | able. |  |

Table 4-18. I/O Interface Support Circuits (488M-A3)

| Support Circuit | Supports | Function |  |
| :---: | :---: | :--- | :---: |
| U2 | U1 | Transceiver |  |
| U3 | U1 | Transceiver |  |
| U5 | U4 | Pull-up resistive network |  |
| U8A | U7 | U6 output inverter |  |
| U8B | U1 | Clock generator |  |
| U8C | NU |  |  |
| U8D | U10 | RESET OUT inverter |  |
| U8E | U6 | ALE inverter |  |
| U8F | U1 | Clock generator |  |
| U9A | NU |  |  |
| U9B | U1 | ATN disable |  |
| U9C | NU |  |  |
| U9D | U10 | Clock input |  |
| U10 | U1 | Transceiver direction control |  |
|  |  |  |  |

### 4.1.4.6.1 I/O Interface Troubleshooting Procedure

1. Reference the following manual data to support Tables 4-17 and 4-18.
a. Circuit descriptions in paragraph 3.3.1.
b. Location of Components Diagram Figure 5-12.
c. Schematic Diagram, Figure 6-7.
d. Modular troubleshooting concept, described in paragraph 4.1.4.1.
2. Recommended test equipment for troubleshooting I/O Interface components are an oscilloscope, digital voltmeter, and ohmeter (reference the equipment list in the WJ- 8718 Series HF Receiver Instruction Manual), and the general purpose digital troubleshooting equipment, described in paragraph 4.1.3.1 of this supplement.

### 4.1.5 SWITCH AND LED TROUBLESHOOTING

The operation of front panel switches, LED indicators, and LED displays can be diagnosed by execution of the BITE program. Troubleshooting the front panel is indicated if a problem arises during the execution of the program which cannot be isolated to a module in the microprocessor structure (excluding MFP-A1 modules), discussed in paragraph 4.1.4.1.

The front panel section of the MFP Option consists of the Tuning Encoder Assembly and the Front Panel Switch/Encode Assembly, mounted directly behind the front panel. The latter assembly contains two subassemblies: the Front Panel Encode Board (MFP-A1A1) and the Front Panel Switchboard (MFP-A1A2). In addition, the assembly contains the track to interconnect the subassemblies and 2-channel audio amplifiers. The front panel boards are interconnected in a manner which facilitates troubleshooting. Plug-in connections allow extender cards to be mounted in the MFP-A1 board. The encode board can be plugged into the extender cards, allowing all points on the front panel to be probed with the unit in operation.

The Front Panel Encode Board contains the logic circuits which interface the front panel switches and LEDs with the microprocessor. The board design reflects the modular concept described in paragraph 4.1.4.1. Troubleshooting the board on a modular level is described in paragraph 4.1.4.4.1. The following troubleshooting information is based on an evaluation of BITE program results and involves both the Front Panel Encode and the Front Panel Switchboards. The switchboard contains two switch matrixes, associated LED indicators, and 7 -segment LED displays. The switches and LEDs are divided into two groups: Group 1 is associated with the right side of the front panel, and Group 2 is associated with the left side.

### 4.1.5.1 BITE Program Evaluation

1. Access and execute the BITE program, as described in paragraph 4.1.1.1. If a problem occurs, proceed to troubleshoot, while referencing the following manual data.
a. The front panel description in paragraph 2.4.4.
b. Schematic Diagrams for the Front Panel Switch/Encode Assembly (Figure 6-2), the Front Panel Switchboard (Figure 6-4), the Front Panel Encode Board (Figure 6-3), and the MFP Option, Main Assembly (Figure 6-9).
c. Location of Component Diagrams, Figures 5-6 through 5-8.
d. The circuit descriptions in paragraph 3.1.2.
e. The microprocessor system troubleshooting procedures in paragraphs 4.1.4 and 4.1.4.4.
2. If the problem is confined to a single switch, suspect the switch itself. Unplug the board behind the front panel, and use unpowered ohmmeter measurements of track and solder connections between the switch and the Front Panel Encode Board to isolate open or shorted circuits.
3. An LED display problem which appears heat related (for example, the loss of segments after the unit has been in operation for some time) may be corrected by adjusting the VR voltage. Refer to paragraph 4.1.2.1 for the adjustment procedure.
4. If all switches and LEDs are malfunctioning, verify the 1 MHz CLK input to U1 and U2 on the Front Panel Encode Board. The components associated with the CLK signal are U13E, U13F, R7, and C11.
5. If a problem is confined to either the Group 1 or Group 2 switches and LEDs, switch the socket-mounted U1 and U2 interface ICs and note any changes in operation. If the problem symptoms disappear, replace the defective IC. If symptoms persist, proceed to the next step.
6. If the problem is associated with all switches in a particular row or column, proceed to troubleshoot circuits common to the row or column. Tables 4-2 and 4-3 identify the Group 1 and Group 2 switches. A common $R$ code within a group identifies switches by row, and a common C code within a group identifies switches by column.
a. During operation, the inputs to column drivers U12 (Group 2) and U11 (Group 1) should count in binary from 0 through 7 and the outputs should toggle. If an oscilloscope probe is connected to the suspected row or column line during execution of the BITE program, activation of a switch in that row or column should cause a change in the oscilloscope display.
7. If a problem arises which affects all LEDs in either Group 1 or Group 2 (refer to paragraphs 3.1.2.2.1 or 3.1.2.2.2), suspect the common-cathode circuits.
a. During operation, the inputs to U4 (Group 1) and U8 (Group 2) should toggle. Likewise, toggle outputs should be evident at Group 1 (U4, U5, and U6) and Group 2 (U8, U9, and U10) common-cathode components. An input or output stuck in the low or high mode is evidence of a problem in the IC, or solder connections associated with the IC.

### 4.1.6 AUDIO SIGNAL TROUBLESHOOTING

Combined audio (AUDIO 1) from Audio Amplifier (A4A10), and ISB/LSB audio (AUDIO 2) from ISB Detector/Audio Board A4A8 are routed through MFP Option switches and amplifiers to the front panel phones jack. In addition, line audio voltage from Audio Amplifier A4A10 passes through MFP switches to the front panel meter. The IF Interface (MFP-A3) and the Front Panel Switch/Encode Assembly (MFP-A1) perform the functions provided by the standard Front Panel Interconnect (A6A2) and the Front Panel Control (A10). Figure 4-3 illustrates the signal flow through the MFP Option in all detection modes except ISB. Figure 4-4 illustrates signal flow in the ISB mode.

Section IV of the WJ-8718 Series HF Receiver Instruction Manual contains audio troubleshooting charts, tables, and checkout procedures designed to localize audio problems. These troubleshooting procedures can effectively localize problems to the MFP portion of the signal path, if references to the standard Front Panel Interconnect and the Front Panel Control are replaced by the equivalent MFP components, as shown in Figures 4-3 and 4-4.

### 4.1.7 POWER SUPPLY TROUBLESHOOTING

The MFP Option utilizes the +15 and $\mathbf{- 1 5}$ Vdc supplies from the receiver rear panel voltage regulators, and 10 V unregulated from the secondary windings of the power transformer T1. Troubleshooting these power supplies is described in the WJ-8718 Series HF Receiver Instruction Manual. Two additional supplies are associated with the MFP Option, described as follows.
a. Regulator MFP-A1A1U1 is located in the front panel section. A voltage of 10 V unregulated enters the regulator via MFP-A1A2J1. A voltage of approximately 7 Vdc should leave the regulator via MFP-A1A2J2. The output voltage is adjustable, as described in paragraph 4.1.2.1.
b. Voltage regulator MFP-U1 is mounted on the back side of the front panel. The regulator inputs 10 V unregulated and provides an output of +5 Vde to power the Front Panel Encode Board (MFP-A1A1). The regulator is mounted in a location that is difficult to reach with test equipment probes. However, MFP-A1A1U1 input +5 Vde from MFP-U1 and is easily accessed. The IC is located at the top side of the board and should have +5 Vdc on pin 40.

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NOTE:
LINE AUDIO

Figure 4-3. Line Audio Flow Chart


Figure 4-4. ISB/LSB Audio Flow Chart

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## SECTION V

## REPLACEMENT PARTS LIST

## 5.1 <br> UNIT NUMBERING METHOD

The unit numbering method of assigning reference designations (electrical symbol numbers) has been used to identify assemblies, subassemblies (and modules) and parts. An example of the unit numbering method follows:

Subassembly Designation A1
Identify from right to left as:

## R1 Class and No. of Item

First (1) resistor (R) of first (1) subassembly (A)

As shown on the main chassis schematic, components which are an integral part of the main chassis have no subassembly designation.

### 5.2 REFERENCE DESIGNATION PREFIX

Partial reference designations have been used on the equipment and on the illustrations in this manual. The partial reference designations consist of the class letter(s) and identifying item number. The complete reference designations may be obtained by placing the proper prefix before the partial reference designations. Reference Designation Prefixes are provided on drawings and illustrations in parentheses within the figure titles.

### 5.3 LIST OF MANUFACTURERS

$\left.\begin{array}{llll}\begin{array}{l}\text { Mfr. } \\ \text { Code }\end{array} & \begin{array}{l}\text { Name and Address }\end{array} & \begin{array}{l}\text { Mfr. } \\ \text { Code }\end{array} & \end{array} \begin{array}{l}\text { Name and Address }\end{array}\right\}$

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| Mfr. Code | Name and Address | Mfr. Code | Name and Address |
| :---: | :---: | :---: | :---: |
| 14632 | Watkins-Johnson Company 700 Quince Orchard Road Gaithersburg, MD 20878 | 71450 | CTS Corporation 905 N. West Boulevard Elkhart, IN 46514 |
| 15818 | Teledyne Semiconductor Teledyne Inc. Company 1300 Tera Bella Avenue Mountain View, CA 94043 | 72982 | Erie Technological Prod., Inc. 644 West 12 th Street Erie, PA 16512 |
| 18324 | Signetics Corporation 811 East Arques Avenue Sunnyvale, CA 94086 | 73138 | Beckman Instruments, Inc. Helipot Division 2500 Harbor Boulevard Fullerton, CA 92634 |
| 19209 | General Electric Company Battery Business Department P.O. Box 114 Gainsville, FL 32602 | 75037 | Minnesota Mining and Mfg. Co. Electro Products Division 3M Center <br> St. Paul, MN 55101 |
| 21604 | The Buckeye Stamping Co. 555 Marion Road Columbus, OH 43207 | 80103 | Lamda Electronics Corp. Div. of Veeco Instr., Inc. 51 Broad Hollow Road Melville, NY 11746 |
| 28480 | Hewlett-Packard Company Corporate Headquarters 1501 Page Mill Road Palo Alto, CA 94304 | 80131 | Electronic Industries Assoc. 2001 Eye Street, N. W. Washington, DC 20006 |
| 34649 | Intel Corporation 3585 S. W. 198th Avenue Aloha, CA 97005 | 81312 | Winchester Electronics Division of Litton Ind. Oakville, CT 06779 |
| 36665 | Mitel Corporation 350 Leggett Drive P.O. Box 13089 Kanata Ontario, CAN K2K 1 X3 | 81349 | Military Specifications |
| 50558 | Electronic Concepts, Inc. <br> 526 Industrial Way <br> West Eatontown, NJ 07724 | 81350 | Joint Army-Navy Specifications |
| 53848 | SMC Microsystems Corp. 35 Marcus Boulevard Hauppauge, NY 11787 | 82389 | Switcheraft, Incorporated 5555 North Elston Avenue Chicago, IL 60630 |
| 56289 | Sprague Electric Company Marshall Street <br> North Adams, MA 01247 | 91506 | Augat, Inc. <br> 33 Perry Avenue <br> P.O. Box 779 <br> Attleboro, MA 02703 |

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## 5.4 <br> PARTS LIST

The parts list which follows contains all electrical parts used in the equipment and certain mechanical parts which are subject to unusual wear or damage. When ordering replacement parts from Watkins-Johnson Company, specify the type and serial number of the equipment and the reference designation and description of each part ordered. The list of manufacturers provided in paragraph 5.3 and the manufacturer's part number for components are included as a guide to the user of the equipment in the field. These parts may not necessarily agree with the parts installed in the equipment; however, the parts specified in this list will provide satisfactory operation of the equipment. Replacement parts may be obtained from any manufacturer as long as the physical and electrical parameters of the part selected agree with the original indicated part. In the case of components defined by a military or industrial specification, a vendor which can provide the necessary component is suggested as a convenience to the user.

## NOTE

As improved semi-conductors become available, it is the policy of Watkins-Johnson to incorporate them in proprietary products. For this reason some transistors, diodes, and integrated circuits installed in the equipment may not agree with those specified in the parts list and schematic diagrams of this manual. However, the semi-conductors designated in the manual may be substituted in every case with satisfactory results.

### 5.5 ASSEMBLY REVISION LEVEL

The purpose of the Assembly Revision Level is to identify the "as built" configuration of an assembly or subassembly. The parts list and illustrations that follow, depict the revision levels of the assemblies and subassemblies at the time of preparation of the manual, which may or may not agree with the purchased equipment. However, they will serve as a guide for any necessary maintenance to be performed. Refer to Table 5-1 for the equipment assembly revision level record(s).

Table 5-1. Equipment Assembly Revision Level Record

| Type Number | Ref. Desig | Description | Assy. Rev. Level |
| :---: | :---: | :---: | :---: |
| WJ-8718A/MFP <br> 796175 <br> 794310-1 (Yel) <br> 794310-2 (Red) <br> 794310-3 (Green) <br> 796056-1 <br> 794309-1 (Yel) <br> 794309-2 (Red) <br> 794309-3 (Green) <br> 791202-5 <br> 794308-2 <br> 794275-X <br> WJ-8718A/488M <br> 796075 <br> WJ-8718A/232M <br> 796037 | A4A6 <br> MFP-A1 <br> MFP-A1 <br> MFP-A1 <br> MFP-A1A1 <br> MFP-A1A2 <br> MFP-A1A2 <br> MFP-A1A2 <br> MFP-A2 <br> MFP-A3 <br> MFP-A4 <br> 488M-A3 <br> 232M-A3 | Microprocessor Front Panel Option <br> AGC Amplifier <br> Front Panel Interconnect <br> Front Panel Interconnect <br> Front Panel Interconnect <br> Front Panel Encode <br> Front Panel Switchboard <br> Front Panel Switchboard <br> Front Panel Switchboard <br> Encoder Assembly <br> IF Interface <br> Synthesizer Interface/Memory <br> I/O Option <br> I/O Interface <br> I/O Option <br> I/O Interface | $\begin{aligned} & \mathrm{A} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{C} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~B} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~A} \\ & \mathrm{~A} \end{aligned}$ |

TYPE WJ-8718A/MFP

|  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| REF DESIG | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{array}{c\|} \text { RECM } \\ \text { VENDOR } \end{array}$ |
| A4A6 | AGC Amplifier | 1 | 796175-1 | 14632 |  |
| MFP-A1* | Front Panel Switch/Encode Assembly (Yellow LED) | X | 794310-1 | 14632 |  |
| MFP-A1* | Front Panel Switch/Encode Assembly (Red LED) | X | 794310-2 | 14632 |  |
| MFP-A1* | Front Panel Switch/Encode Assembly (Green LED) | X | 794310-3 | 14632 |  |
| MFP-A2 | Encoder Assembly | 1 | 791202-5 | 14632 |  |
| MFP-A3 | IF Interface | 1 | 794308-2 | 14632 |  |
| MFP-A4 | Synthesizer Interface/Memory ** | 1 | 794275-X | 14632 |  |
| MFP-C1 | Capacitor, Ceramic, Disc: $0.47 \mu \mathrm{~F}, 20 \%$, 50 V | 2 | 34452-1 | 14632 |  |
| MFP-C2 | Same as MFP-C1 |  |  |  |  |
| MFP-J1 | Connector, Phone Jack | 1 | L12B | 82389 |  |
| MFP-M1 | Meter | 1 | 34455-1 | 14632 |  |
| MFP-MP1 | Handle, Front | 2 | 32306-1 | 14632 |  |
| MFP-MP2 | Same as MFP-MP1 |  |  |  |  |
| MFP-MP3 | Knob | 1 | PS50D1/70C2/BLK | 21604 |  |
| MFP-MP4 | Knob | 1 | PS70D1/B | 21604 |  |
| MFP-MP5 | Knob | 1 | 280064-1 | 14632 |  |
| MFP-MP6 | Display Window (Small) | 1 | 170368-1 | 14632 |  |
| MFP-MP7 | Display Window (Large) | 1 | 271001 | 14632 |  |
| MFP-P1 | Connector, Plug (Part of MFP-W1) | 1 | 88475-2 | 00779 |  |
| MFP-P2 | Connector, Plug (Part of MFP-W1) | 1 | 88475-3 | 00779 |  |
| MFP-P3 | Connector, Plug (Part of MFP-W1) | 1 | 3332-0000 | 75037 |  |
| MFP-P4 | Not Used |  |  |  |  |
| MFP-P5 | Not Used |  |  |  |  |
| MFP-P6 | Connector, Plug (Part of MFP-W2) | 2 | 88476-7 | 14632 |  |
| MFP-P7 | Same as MFP-P6 (Part of MFP-W2) |  |  |  |  |
| MFP-P8 | Connector, Plug Faston Receptacle | 9 | 42236-1 | 00779 |  |
| $\begin{aligned} & \text { MFP-P9 } \\ & \text { Thru } \\ & \text { MFP-P13 } \end{aligned}$ | Same as MFP-P8 |  |  |  |  |
| MFP-P14 | Connector, Receptacle Faston Tab | 2 | 62073-1 | 00779 |  |
| MFP-P15 | Same as MFP-P14 |  |  |  |  |
| MFP-P16 | Connector, Plug | 1 | 1-87499-1 | 00779 |  |
| MFP-P17 | Same as MFP-P8 (Part of MFP-W3) |  |  |  |  |
| MFP-P18 | Same as MFP-P8 (Part of MFP-W3) |  | $\sim$ |  |  |
| MFP-P19 | Not Used |  |  |  |  |
| MFP-P20 | Not Used |  |  |  |  |
|  | * Customer selected option for front panel LED Display. |  |  |  |  |
|  | ** Customer selected due to specific software package. |  |  |  |  |

## Courtesy of http://BlackRadios.terryo.org


${ }^{\dagger}$ PART OF ASSEMBLY MFP-AIA2

Figure 5-1a. WJ-8718A/MFP Front Panel, Location of Components

Courtesy of http://BlackRadios.terryo.org

t PART OF ASSEMBLY MFP-AIA2

Figure 5-1b. WJ-8718A/MFP Front Panel, Location of Components

Courtesy of http://BlackRadios.terryo.org

REF DESIG PREFIX MICROPROCESSOR OPTION

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | $\begin{aligned} & \text { MFR. } \\ & \text { CODE } \end{aligned}$ | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| MFP-P21 | Same as MFP-P8 |  |  |  |  |
| MFP-P22 | Connector, Plug | 1 | 87499-5 | 00779 |  |
| MFP-U1 | Voltage Regulator | 1 | LAS1405 | 80103 |  |
| MFP-W1 | Cable Assembly | 1 | 371048-1 | 14632 |  |
| MFP-W2 | Cable Assembly | 1 | 380092-2 | 14632 |  |
| MFP-W3 | Cable Assembly | 1 | 380142-1 | 14632 |  |
| MFP-XU1 | Socket, Transistor | 1 | 8080-1G1 | 91506 |  |
| R1 | Potentiometer, Modified | 1 | 170422 | 14632 |  |
| R3 | Resistor, Fixed, Film: $200 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-200 OHMS/J | 09021 |  |
| R4 | Resistor, Fixed, Composition: 10ת, 5\%, 1/2 W | 1 | RCR20G100JS | 81349 |  |
| XU3 | Socket Assembly | X | 380132-1 | 14632 |  |



* EXAMPLE OPTION SHOWN

Figure 5-2. WJ-8718A/MFP Rear Panel, Location of Components


Figure 5-3. WJ-8718A/MFP Top View, Location of Components


Figure 5-4. WJ-8718A/MFP Bottom View, Location of Components

TYPE 796175-1 AGC AMPLIFIER
REF DESIG PREFIX A4A6

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION | QTY <br> PER <br> ASSY | MANUFACTURER'S PART NO. | MFR. CODE | RECM VENDOR |
| :---: | :---: | :---: | :---: | :---: | :---: |
| C1 | Capacitor, Electrolytic, Tantalum: $47 \mu \mathrm{~F}, 10 \%, 20 \mathrm{~V}$ | 2 | CS13BE476K | 81349 |  |
| C2 | Not Used |  |  |  |  |
| C3 | Capacitor, Ceramic, Disc: $0.47 \mu \mathrm{~F}, 20 \%, 50 \mathrm{~V}$ | 2 | 34452-1 | 14632 |  |
| C4 | Capacitor, Electrolytic, Tantalum: $33 \mu \mathrm{~F}, 10 \%, 10 \mathrm{~V}$ | 1 | CS13BC336K | 81349 |  |
| C5 | Same as C3 |  |  |  |  |
| C6 | Capacitor, Ceramic, Dise: $0.1 \mu \mathrm{~F}, 20 \%, 50 \mathrm{~V}$ | 2 | 34475-1 | 14632 |  |
| C7 | Same as C1 |  |  |  |  |
| C8 | Same as C6 |  |  |  |  |
| $\begin{aligned} & \text { CR1 } \\ & \text { Thru } \\ & \text { CR4 } \end{aligned}$ | Not Used |  |  |  |  |
| CR5 | Diode | 5 | 1N4449 | 80131 |  |
| $\begin{aligned} & \text { CR6 } \\ & \text { Thru } \\ & \text { CR9 } \end{aligned}$ | Same as CR5 |  |  |  |  |
| CR10 | Diode | 1 | 5082-2800 | 28480 |  |
| Q1 | Transistor | 5 | 2N2222A | 80131 |  |
| Q2 | Transistor | 1 | 2N2907/JAN | 81350 |  |
| Q3 | Transistor | 1 | U1899E | 15818 |  |
| Q4 Thru Q7 | Same as Q1 |  |  |  |  |
| R1 | Resistor, Fixed, Film: $100 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 6 | CF1/4-100K/J | 09021 |  |
| R2 | Resistor, Fixed, Film: $47 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-47K/J | 09021 |  |
| R3 | Resistor, Fixed, Film: $470 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-470K/J | 09021 |  |
| R4 | Resistor, Fixed, Film: $100 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 5 | CF1/4-100 OHMS/J | 09021 |  |
| R5 | Same as R1 |  |  |  |  |
| R6 | Resistor, Fixed, Film: $330 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CR1/4-330K/J | 09021 |  |
| R7 | Resistor, Fixed, Film: $6.8 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-6.8K/J | 09021 |  |
| R8 | Same as R4 |  |  |  |  |
| R9 | Resistor, Fixed, Film: $15 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 4 | CF1/4-15K/J | 09021 |  |
| R10 | Resistor, Fixed, Film: $150 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-150k/J | 09021 |  |
| R11 | Resistor, Fixed, Film: $10 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 9 | CF1/4-10K/J | 09021 |  |
| R12 | Resistor, Fixed, Film: $82 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-82K/J | 09021 |  |
| R13 | Resistor, Fixed, Film: $1.0 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 6 | CF1/4-1K/J | 09021 |  |
| R14 | Resistor, Fixed, Film: $1.2 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-1.2K/J | 09021 |  |
| R15 | Same as R13 |  |  |  |  |
| R16 | Resistor, Fixed, Film: $1.3 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-13K/J | 09021 |  |
| R17 | Resistor, Fixed, Film: $22 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-22K/J | 09021 |  |
| R18 | Resistor, Fixed, Film: $270 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-270K/J | 09021 |  |
| R19 | Resistor, Fixed, Film: $680 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-680K/J | 09021 |  |
| R20 | Same as R11 |  |  |  |  |
| R21 | Same as R9 |  |  |  |  |
| R22 | Same as R99 |  |  |  |  |

REF DESIG PREFIX A4A6

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R23 | Same as R3 |  |  |  |  |
| R24 | Same as R9 |  |  |  |  |
| R25 | Resistor, Fixed, Film: $1.5 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-1.5K/J | 09021 |  |
| R26 | Resistor, Fixed, Film: $2.2 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-2.2K/J | 09021 |  |
| R27 | Same as R11 |  |  |  |  |
| R28 | Same as R4 |  |  |  |  |
| R29 | Same as R4 |  |  |  |  |
| R30 | Same as R11 |  |  |  |  |
| R31 | Resistor, Fixed, Film: $2.7 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-2.7K/J | 09021 |  |
| R32 | Resistor, Fixed, Film: $390 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-390 OHMS/J | 09021 |  |
| R33 | Resistor, Fixed, Film: $360 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-360K/J | 09021 |  |
| R34 | Not Used |  |  |  |  |
| R35 | Resistor, Fixed, Film: $4.7 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-4.7K/J | 09021 |  |
| R36 | Resistor, Fixed, Film: $1.0 \mathrm{M} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-1M/J | 09021 |  |
| R37 | Same as R1 |  |  |  |  |
| R38 | Resistor, Fixed, Film: $68 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-68K/J | 09021 |  |
| R39 | Same as R36 |  |  |  |  |
| R40 | Same as R1 |  |  |  |  |
| R41 | Same as R7 |  |  |  |  |
| R42 | Same as R4 |  |  |  |  |
| R43 | Same as R2 |  |  |  |  |
| R44 | Not Used |  |  |  |  |
| R45 | Same as R11 |  |  |  |  |
| R46* | Resistor, Fixed, Film: $39 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-39K/J | 09021 |  |
| R47 | Resistor, Fixed, Film: $820 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-820 OHMS/J | 09021 |  |
| R48 | Resistor, Fixed, Film: $68 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-68 OHMS/J | 09021 |  |
| R49 | Same as R46 |  |  |  |  |
| R50* | Resistor, Fixed, Film: $3.9 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-3.9K/J | 09021 |  |
| R51* | Same as R32 |  |  |  |  |
| R52 | Same as R13 |  |  |  |  |
| R53 | Same as R1 |  |  |  |  |
| R54 | Same as R1 |  |  |  |  |
| R55 | Not Used |  |  |  |  |
| R56 | Not Used |  |  |  |  |
| R57 | Resistor, Fixed, Film: $1.5 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | RN55C1501F | 81349 |  |
| R58 | Same as R11 |  |  |  |  |
| R59 | Same as R13 |  |  |  |  |
| R60 | Same as R11 |  |  |  |  |
| R61 | Same as R13 |  |  |  |  |
| R62 | Same as R11 <br> * Nominal value, final value factory selected |  |  |  |  |

REF DESIG PREFIX A4A6

| REF <br> DESIG | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R63 | Same as R13 |  |  |  |  |
| R64 | Same as R11 |  |  |  |  |
| U1 | Integrated Circuit | 2 | MC3403P | 04713 |  |
| U2 | Same as U1 |  |  |  |  |
| U3 | Integrated Circuit | 1 | MC1458N | 18324 |  |
| U4 | Integrated Circuit | 1 | CA3018A | 02735 |  |
| VR1 | Diode, Zener: 5.6 V | 1 | 1N752A | 80131 |  |



Figure 5-5. Type 796175-1 AGC Amplifier (A4A6), Location of Components

REF DESIG PREFIX MFP-A1

| REF <br> DESIG | DESCRIPTION |  | MANUFACTURER'S PART NO. | $\begin{gathered} \text { MFR. } \\ \text { CODE } \end{gathered}$ | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A1 | Front Panel Encode | 1 | 796056-1 | 14632 |  |
| A2 | $\begin{array}{ll} \text { Front Panel Switch Board } & \text { (Yellow LED) } \\ & \text { (Used on 794310-1 only) } \end{array}$ | 1 | 794309-1 | 14632 |  |
| A2 | Front Panel Switch Board (Red LED) <br> (Used on 794310-2 only) | 1 | 794309-2 | 14632 |  |
| A2 | $\begin{array}{ll} \text { Front Panel Switch Board } & \begin{array}{l} \text { (Green LED) } \\ \\ \text { (Used on 794310-3 only) } \end{array} \end{array}$ | 1 | 794309-3 | 14632 |  |
| C1 | Capacitor, Ceramic, Disc: $0.1 \mu \mathrm{~F}, 20 \%, 100 \mathrm{~V}$ | 2 | 8131M100-651-104M | 72982 |  |
| C2 | Same as C1 |  |  |  |  |
| E1 | Connector, Board Joiner | 3 | WD22P6 | 81312 |  |
| $\begin{array}{\|l} \text { E2 } \\ \text { Thru } \\ \text { E66 } \end{array}$ | Same as E1 |  |  |  |  |
| E67 | Connector, Board Joiner | 1 | WD11P6 | 81312 |  |
| E68 <br> Thru <br> E77 | Same as E67 |  |  |  |  |
| J1 | Connector, Receptacle | 1 | 87227-7 | 00779 |  |
| J2 | Connector, Receptacle | 1 | 87227-8 | 00779 |  |
| J3 | Connector, Receptacle | 1 | 87227-3 | 00779 |  |
| J4 | Connector, Receptacle | 1 | 87224-6 | 00779 |  |
| J5 | Connector, Receptacle, Faston Tab | 4 | 62073-1 | 00779 |  |
| J6 <br> Thru <br> J8 | Same as J5 |  |  |  |  |
| R1 | Resistor, Fixed, Film: $100 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-100K/J | 09021 |  |
| R2 | Resistor, Fixed, Film: $1.0 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-1.0K/J | 09021 |  |
| R3 | Resistor, Fixed, Film: $220 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-220 OHMS/J | 09021 |  |
| R4 | Same as R3 |  |  |  |  |
| R5 | Same as R1 |  |  |  |  |
| R6 | Same as R2 |  |  |  |  |
| R7 | Resistor, Variable: $25 \mathrm{k} \Omega / 25 \mathrm{k} \Omega$ | 1 | 20M277-25K/25K | 01121 |  |
| R8 | Resistor, Variable: $25 \mathrm{k} \Omega, 10 \%, 1 \mathrm{~W}$ | 1 | 70 M 3 N 056 L 253 U | 01121 |  |
| U1 | Integrated Circuit | 1 | MC3403P | 04713 |  |
| XAI | Connector, Printed Circuit Board | 1 | 67987-6 | 00779 |  |



Figure 5-6. Type 794310-1, -1, -3 Front Panel Switch/Encode (MFP-A1), Location of Components
5.6.2.1 Type 796056-1 Front Panel Encode

REF DESIG PREFIX MFP-A1A1

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | $\begin{aligned} & \text { MFR. } \\ & \text { CODE } \end{aligned}$ | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| C1 | Capacitor, Ceramic, Disc: $0.1 \mu \mathrm{~F}, 20 \%, 100 \mathrm{~V}$ | 11 | $8131 \mathrm{M} 100-651-104 \mathrm{M}$ | 72982 |  |
| C2 <br> Thru C10 | Same as C1 |  |  |  |  |
| C11 | Capacitor, Ceramic, Disc: $0.01 \mu \mathrm{~F}, 20 \%$, 200 V | 1 | 8131 A 200 Z 5 U 103 M | 72982 |  |
| C12 | Capacitor, Electrolytic, Tantalum: $47 \mu \mathrm{~F}, 20 \%, 20 \mathrm{~V}$ | 2 | 196D476X0020PE4 | 56289 |  |
| C13 | Same as C12 |  |  |  |  |
| C14 | Capacitor, Electrolytic, Tantalum: $100 \mu \mathrm{~F}, 20 \%, 20 \mathrm{~V}$ | 3 | 196D107X0020TE4 | 56289 |  |
| C15 | Same as C14 |  |  |  |  |
| C16 | Same as C14 |  |  |  |  |
| C17 | Same as C1 |  |  |  |  |
| C18 | Capacitor, Ceramic, Disc: $0.47 \mu \mathrm{~F}, 20 \%, 100 \mathrm{~V}$ | 1 | $8131 \mathrm{M} 100-651-474 \mathrm{M}$ | 72982 |  |
| C19 | Capacitor, Electrolytic, Tantalum: $22 \mu \mathrm{~F}, 20 \%, 10 \mathrm{~V}$ | 1 | 196D226X0010JE3 | 56289 |  |
| CR1 | Diode | 3 | 1N4449 | 80131 |  |
| CR2 | Same as CR1 |  |  |  |  |
| CR3 | Same as CR1 |  |  |  |  |
| J1 | Connector, Receptacle | 1 | 1-87567-6 | 00779 |  |
| J2 | Connector, Receptacle Faston Tab | 1 | 62073-1 | 00779 |  |
| R1 | Resistor, Fixed, Film: $6.8 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-6.8K/J | 09021 |  |
| R2 | Resistor, Fixed, Film: $3.9 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-3.9K/J | 09021 |  |
| R3 | Resistor, Fixed, Film: $3.3 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-3.3K/J | 09021 |  |
| R4 | Resistor, Fixed, Film: $10 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 14 | CF1/4-10K/J | 09021 |  |
| R5 | Same as R4 |  |  |  |  |
| R6 | Same as R4 |  |  |  |  |
| R7 | Resistor, Fixed, Film: $43 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-43 OHMS/J | 09021 |  |
| R8* | Resistor, Fixed, Film: $1.8 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-1.8K/J | 09021 |  |
| R9 | Same as R4 |  |  |  |  |
| R10 | Not Used |  |  |  |  |
| R11 <br> Thru <br> R18 | Same as R4 |  |  |  |  |
| R19 | Resistor, Fixed, Film: $2.0 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-2.0K/J | 09021 |  |
| R20 | Same as R4 |  |  |  |  |
| R21 | Same as R4 |  |  |  |  |
| U1 | Integrated Circuit | 2 | P8279 | 34649 |  |
| U2 | Same as U1 |  |  |  |  |
| U3 | Integrated Circuit | 2 | DS8857N | 27014 |  |
| U4 | Integrated Circuit | 2 | CD4067BE | 02735 |  |
| U5 | Integrated Circuit | 4 | DS8863N | 27014 |  |
| U6 | Same as U5 |  |  |  |  |
| U7 | Same as U3 <br> * Nominal value, final value factory selected. |  |  |  |  |

Courtesy of http://BlackRadios.terryo.org


Figure 5-7. Type 796056-1 Front Panel Encode (MFP-A1A1), Location of Components

REF DESIG PREFIX MFP-A1A1

| REF DESIG | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{aligned} & \text { RECM } \\ & \text { VENDOR } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| U8 | Same as U4 |  |  |  |  |
| U9 | Same as U5 |  |  |  |  |
| U10 | Same as U5 |  |  |  |  |
| U11 | Integrated Circuit | 3 | SN74LS138N | 01295 |  |
| U12 | Same as U11 |  |  |  |  |
| U13 | Integrated Circuit | 1 | SN74LS14N | 01295 |  |
| U14 | Integrated Circuit | 1 | MC14053BCP | 04713 |  |
| U15 | Integrated Circuit | 1 | SN74LS08N | 01295 |  |
| U16 | Integrated Circuit | 1 | SN74LS32N | 01295 |  |
| U17 | Integrated Circuit | 1 | MM74C173N | 27014 |  |
| U18 | Integrated Circuit | 2 | SN74LS273N | 01295 |  |
| U19 | Same as U11 |  |  |  |  |
| U20 | Same as U18 |  |  |  |  |
| U21 | Integrated Circuit | 1 | MM74C14N | 27014 |  |
| VR1 | Diode, Zener: 5.1 V | 1 | 1 N 751 A | 80131 |  |
| XU1 | Socket, Integrated Circuit | 2 | ICN-406-S5-T | 06776 |  |
| XU2 | Same as XU1 |  |  |  |  |
| XU3 | Socket, Integrated Circuit | 6 | ICN-163-S3-T | 06776 |  |
| XU4 | Socket, Integrated Circuit | 2 | ICN-246-S5-T | 06776 |  |
| XU5 | Not Used |  |  |  |  |
| XU6 | Not Used |  |  |  |  |
| XU7 | Same as XU3 |  |  |  |  |
| XU8 | Same as XU4 |  |  |  |  |
| XU9 | Not Used |  |  |  |  |
| XU10 | Not Used |  |  |  |  |
| xU11 | Same as XU3 |  |  |  |  |
| XU12 | Same as XU3 |  |  |  |  |
| xU13 | Socket, Integrated Circuit | 3 | ICN-143-S3-T | 06776 |  |
| XU14 | Not Used |  |  |  |  |
| XU15 | Same as XU13 |  |  |  |  |
| XU16 | Same as XU13 |  |  |  |  |
| XU17 | Same as XU3 |  |  |  |  |
| XU18 | Socket, Integrated Circuit | 2 | ICN-203-S3-T | 06776 |  |
| XU19 | Same as XU3 |  |  |  |  |
| XU20 | Same as XU18 |  |  |  |  |

5.6.2.2 Type 794309-1, -2, -3 Front Panel Switchboard

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION | QTY <br> PER <br> ASSY | MANUFACTURER'S PART NO. | MFR. CODE | RECM VENDOR |
| :---: | :---: | :---: | :---: | :---: | :---: |
| C1 | Capacitor, Electrolytic, Tantalum: $1 \mu \mathrm{~F}, 20 \%, 35 \mathrm{~V}$ | 1 | 196D105X0035HE3 | 56289 |  |
| DS1* | Display, LED (Yellow) (Used on 794309-1 only) | 13 | 5082-7623 | 28480 |  |
| DS1* | Display, LED (Red) (Used on 79309-2 only) | 13 | 5082-7613 | 28480 |  |
| DS1* | Display, Led (Green) (Used on 794309-3 only) | 13 | 5082-7633 | 28480 |  |
| $\begin{aligned} & \text { DS2* } \\ & \text { Thru } \\ & \text { DS4 } \end{aligned}$ | Same as DS1 |  |  |  |  |
| DS5 | Not Used |  |  |  |  |
| DS6 | Not Used |  |  |  |  |
| DS7* <br> Thru <br> DS13* | Same as DS1 |  |  |  |  |
| DS14 | Not Used |  |  |  |  |
| DS15* | Same as DS1 |  |  |  |  |
| DS16* | Same as DS1 |  |  |  |  |
| DS17 | Not Used |  |  |  |  |
| DS18 | Not Used |  |  |  |  |
| DS19 | Part of S33 |  |  |  |  |
| DS20 | Part of S17 |  |  |  |  |
| DS21 | Part of S18 |  |  |  |  |
| DS22 | Part of S19 |  |  |  |  |
| DS23 | Part of S20 |  |  |  |  |
| DS24 | Part of S21 |  |  |  |  |
| DS25 | Part of S22 |  |  |  |  |
| DS26 | Part of S23 |  |  |  |  |
| DS27 | Part of S24 |  |  |  |  |
| DS28 | Part of S25 |  |  |  |  |
| DS29 | Part of S26 |  |  |  |  |
| DS30 | Part of S27 |  |  |  |  |
| DS31 | Not Used |  |  |  |  |
| DS32 | Part of S29 |  |  |  |  |
| DS33 | Part of S30 |  |  |  |  |
| DS34 | Part of S31 |  |  |  |  |
| DS35 | Not Used |  |  |  |  |
| DS36 | Part of S34 |  |  |  |  |
| DS37 | Part of S35 |  |  |  |  |
| DS38 | Part of S36 |  |  |  |  |
| DS39 | Part of S37 |  |  |  |  |
| DS40 | Part of S38 |  |  |  |  |
| DS41 | Not Used |  |  |  |  |
|  | * Luminous intensity code for display shall be matched at A2 level or brighter. |  |  |  |  |

REF DESIG PREFIX MFP-A1A2

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION | $\begin{array}{\|l\|} \hline \text { QTY } \\ \text { PER } \\ \text { ASSY } \end{array}$ | MANUFACTURER'S PART NO. | $\begin{aligned} & \text { MFR. } \\ & \text { CODE } \end{aligned}$ | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| DS42 | Not Used |  |  |  |  |
| DS43 | Not Used |  |  |  |  |
| DS44 | Part of S41 |  |  |  |  |
| DS45 | Part of S44 |  |  |  |  |
| DS46 | Part of S42 |  |  |  |  |
| DS47 | Part of S45 |  |  |  |  |
| DS48 | Part of S46 |  |  |  |  |
| DS49 | Part of S47 |  |  |  |  |
| DS50 | Not Used |  |  |  |  |
| DS51 | Part of S49 |  |  |  |  |
| DS52 | Part of S50 |  |  |  |  |
| DS53 | Part of S51 |  |  |  |  |
| DS54 | Part of S52 |  |  |  |  |
| DS55 | Part of S53 |  |  |  |  |
| DS56 | Part of S54 |  |  |  |  |
| DS57 | Not Used |  |  |  |  |
| DS58 | Not Used |  |  |  |  |
| DS59 | Not Used |  |  |  |  |
| DS60 | Diode, LED (Yellow) (Used on 794309-1 only) | 1 | 5082-4150 | 28480 |  |
| DS60 | Diode, LED (Red) (Used on 794309-2 only) | 1 | 5082-4160 | 28480 |  |
| DS60 | Diode, LED (Green) (Used on 794309-3 only) | 1 | 5082-4190 | 28480 |  |
| DS61 | Diode, LED (Red) | 3 | HLMP-1301 | 28480 |  |
| DS62 | Same as DS61 |  |  |  |  |
| DS63 | Same as DS61 |  |  |  |  |
| J1 | Connector, Receptacle, Faston | 2 | 62073-1 | 00779 |  |
| J2 | Same as J1 |  |  |  |  |
| R1 | Resistor, Fixed, Film: $510 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-510 OHMS/J | 09021 |  |
| R2 | Resistor, Variable: $10 \mathrm{k} \Omega, 10 \%, 3 / 4 \mathrm{~W}$ | 1 | 89PR10K | 73138 |  |
| R3 | Resistor, Fixed, Film: $240 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-240 OHMS/J | 09021 |  |
| S1 | Switch, Push Button Engraved (0) | 1 | 271050-21 | 14632 |  |
| S2 | Switch, Push Button Engraved (1) | 1 | 271050-22 | 14632 |  |
| S3 | Switch, Push Button Engraved (2) | 1 | 271050-23 | 14632 |  |
| S4 | Switch, Push Button Engraved (3) | 1 | 271050-24 | 14632 |  |
| S5 | Switch, Push Button Engraved (4) | 1 | 271050-25 | 14632 |  |
| S6 | Switch, Push Button Engraved (5) | 1 | 271050-26 | 14632 |  |
| S7 | Switch, Push Button Engraved (6) | 1 | 271050-27 | 14632 |  |
| S8 | Switch, Push Button Engraved (7) | 1 | 271050-28 | 14632 |  |
| S9 | Switch, Push Button Engraved (8) | 1 | 271050-29 | 14632 |  |
| S10 | Switch, Push Button Engraved (9) | 1 | 271050-30 | 14632 |  |
| S11 | Switeh, Push Button Engraved (.) | 1 | 271050-31 | 14632 |  |



Figure 5-8. Type 794309-1, -2, -3 Front Panel Switchboard (MFP-A1A2), Location of Components

REF DESIG PREFIX MFP-A1A2

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | RECM VENDOR |
| :---: | :---: | :---: | :---: | :---: | :---: |
| S12 | Switch, Push Button Engraved (CLEAR) | 1 | 271050-32 | 14632 |  |
| S13 | Switch, Push Button Engraved (BFO) (+/-) | 1 | 271050-18 | 14632 |  |
| S14 | Switch, Push Button Engraved (MHz) | 1 | 271050-19 | 14632 |  |
| S15 | Switch, Push Button Engraved ( kHz ) | 1 | 271050-20 | 14632 |  |
| S16 | Switch, Push Button Engraved (*) | 1 | 271050-46 | 14632 |  |
| S17 | Switch, Push Button Engraved (LOCK) | 1 | 271050-35 | 14632 |  |
| S18 | Switch, Push Button Engraved (SLOW) | 2 | 271050-13 | 14632 |  |
| S19 | Switch, Push Button Engraved (MED) | 1 | 271050-38 | 14632 |  |
| S20 | Switch, Push Button Engraved (FAST) | 2 | 271050-12 | 14632 |  |
| S21 | Switch, Push Button Engraved (AM) | 1 | 271050-1 | 14632 |  |
| S22 | Switch, Push Button Engraved (FM) | 1 | 271050-2 | 14632 |  |
| S23 | Switch, Push Button Engraved (USB) | 1 | 271050-3 | 14632 |  |
| S24 | Switch, Push Button Engraved (LSB) | 1 | 271050-4 | 14632 |  |
| S25 | Switch, Push Button Engraved (ISB) | 1 | 271050-5 | 14632 |  |
| S26 | Switch, Push Button Engraved (CWV) | 1 | 271050-42 | 14632 |  |
| S27 | Switch, Push Button Engraved (CWF) | 1 | 271050-43 | 14632 |  |
| S28 | Not Used |  |  |  |  |
| S29 | Switch, Push Button Engraved (MGC) | 1 | 271050-11 | 14632 |  |
| S30 | Same as S18 |  |  |  |  |
| S31 | Same as S20 |  |  |  |  |
| S32 | Not Used |  |  |  |  |
| S33 | Switch, Push Button Engraved (BFO) | 1 | 271050-36 | 14632 |  |
| S34 | Switch, Push Button Engraved (.3) | 1 | 271050-6 | 14632 |  |
| S35 | Switch, Push Button Engraved (1) | 1 | 271050-7 | 14632 |  |
| S36 | Switch, Push Button Engraved (3.2) | 1 | 271050-8 | 14632 |  |
| S37 | Switch, Push Button Engraved (6) | 1 | 271050-9 | 14632 |  |
| S38 | Switch, Push Button Engraved (16) | 1 | 271050-10 | 14632 |  |
| S39 | Not Used |  |  |  |  |
| S40 | Not Used |  |  |  |  |
| S41 | Switch, Push Button Engraved (LOCAL) | 1 | 271050-14 | 14632 |  |
| S42 | Switch, Push Button Engraved (RECALL) | 1 | 271050-16 | 14632 |  |
| S43 | Not Used |  | 50-15 |  |  |
| S44 | Switch, Push Button Engraved (REMOTE) | 1 | 271050-15 | 14632 |  |
| S45 | Switch, Push Button Engraved (STORE) | 1 | 271050-17 | 14632 |  |
| S46 | Switch, Push Button Engraved (LINE) (AUDIO) | 1 | 271050-40 | 14632 |  |
| S47 | Switch, Push Button Engraved (SIGNAL) (STR) | 1 | 271050-41 | 14632 |  |
| S48 | Not Used |  |  |  |  |
| S49 | Switch, Push Button Engraved (EXAM) | 1 | 271050-47 | 14632 |  |
| S50 | Switch, Push Button Engraved (HAND)(OFF) | 1 | 271050-48 | 14632 |  |
| S51 | Switch, Push Button Engraved (AUTO)(SCAN) | 1 | 271050-44 | 14632 |  |

Courtesy of http://BlackRadios.terryo.org

REF DESIG PREFIX MFP-A1A2

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{aligned} & \text { RECM } \\ & \text { VENDOR } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| S52 | Switch, Push Button Engraved (LOCK)(OUT) | 1 | 271050-45 | 14632 |  |
| S53 | Switch, Push Button Engraved (THRS) | 1 | 271050-39 | 14632 |  |
| S54 | Switch, Push Button Engraved (DWELL) | 1 | 271050-37 | 14632 |  |
| $\begin{aligned} & \text { S55 } \\ & \text { Thru } \\ & \text { S57 } \end{aligned}$ | Not Used |  |  |  |  |
| U1 | Voltage Regulator | 1 | LM317T | 27014 |  |

TYPE 791202-5 ENCODER ASSEMBLY
REF DESIG PREFIX MFP-A2

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. <br> CODE | $\begin{array}{\|c\|} \text { RECM } \\ \text { VENDOR } \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| P1 | Housing | 1 | 87456-2 | 00779 |  |
| U1 | Encoder Assembly | 1 | 34836-1 | 14632 |  |



Figure 5-9. Type 791202-5 Encoder Assembly
(MFP-A2), Location of Components

| 5.6.4 | TYPE 794308-2 IF INTERFACE | REF | DESIG PREFX MFP-A3 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION | $\begin{array}{\|l\|} \hline \text { QTY } \\ \text { PER } \\ \text { ASSY } \\ \hline \end{array}$ | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| C1 | Capacitor, Ceramic, Disc: $0.1 \mu \mathrm{~F}, 20 \%, 50 \mathrm{~V}$ | 9 | 34475-1 | 14632 |  |
| $\begin{aligned} & \text { C2 } \\ & \text { Thru } \\ & \text { C5 } \end{aligned}$ | Same as C1 |  |  |  |  |
| C6 | Capacitor, Ceramic, Chip: $220 \mathrm{pF}, 5 \%, 100 \mathrm{~V}$ |  |  |  |  |
| C7 | Same as C1 |  |  |  |  |
| C8 | Capacitor, Electrolytic, Tantalum: $18 \mu \mathrm{~F}, 10 \%, 20 \mathrm{~V}$ | 2 | 196D186X9020KE3 | 56289 |  |
| C9 | Same as C8 |  |  |  |  |
| C10 | Same as C1 |  |  |  |  |
| C11 | Same as C1 |  |  |  |  |
| C12 | Capacitor, Electrolytic, Tantalum: $22 \mu \mathrm{~F}, 20 \%, 15 \mathrm{~V}$ | 1 | 196D226X0015KE3 | 56289 |  |
| C13 | Not Used |  |  |  |  |
| C14 | Capacitor, Electrolytic, Tantalum: $4.7 \mu \mathrm{~F}, 20 \%, 35 \mathrm{~V}$ | 1 | 196D475X0035JE3 | 56289 |  |
| C15 | Capacitor, Electrolytic, Tantalum: $47 \mu \mathrm{~F}, 20 \%, 20 \mathrm{~V}$ | 1 | 196D476X0020PE4 | 56289 |  |
| C16 | Same as C1 |  |  |  |  |
| CR1 | Diode | 3 | 1N4449 | 80131 |  |
| CR2 | Same as CR1 |  |  |  |  |
| CR3 | Same as CR1 |  |  |  |  |
| J1 | Connector, Receptacle | 1 | 1-87567-6 | 00779 |  |
| J2 | Connector, Receptacle | 3 | 87567-4 | 00779 |  |
| J3 | Same as J2 |  |  |  |  |
| J4 | Same as J2 |  |  |  |  |
| R1 | Resistor, Fixed, Film: $12 \mathrm{k} \Omega, 5 \%, 1 / 8 \mathrm{~W}$ | 2 | CF1/8-12K/J | 09021 |  |
| R2 | Resistor, Fixed, Composition: $22 \Omega, 5 \%, 1 / 2 \mathrm{~W}$ | 1 | RCR20G220JS | 81349 | 01121 |
| R3 | Resistor, Fixed, Composition: $2.7 \mathrm{k} \Omega, 5 \%, 1 / 2 \mathrm{~W}$ | 1 | RCR20G272JS | 81349 | 01121 |
| R4 | Same as R1 |  |  |  |  |
| R5 | Resistor, Fixed, Film: $1.0 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-1.0K/J | 09021 |  |
| R6 | Resistor, Fixed, Film: $100 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-100 OHMS/J | 09021 |  |
| R7 | Resistor, Variable: $20 \mathrm{k} \Omega, 10 \%, 1 / 2 \mathrm{~W}$ | 1 | 62PAR20K | 73138 |  |
| R8 | Resistor, Fixed, Film: $12 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-12K/J | 09021 |  |
| R9 | Resistor, Fixed, Film: $10 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 6 | CF1/4-10K/J | 09021 |  |
| R10 <br> Thru <br> R13 | Not Used |  |  |  |  |
| R14 | Resistor, Fixed, Film: $39 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-39K/J | 09021 |  |
| R15 | Same as R9 |  |  |  |  |
| R16 | Resistor, Fixed, Film: $82 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-82K/J | 09021 |  |
| R17 | Resistor, Fixed, Film: $8.2 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 2 | CF1/4-8.2K/J | 09021 |  |
| R18 | Same as R9 |  |  |  |  |
| R19 | Same as R5 |  |  |  |  |
| R20 | Same as R17 |  |  |  |  |
| R21 | Same as R9 |  |  |  |  |

REF DESIG PREFIX MFP-A3

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION | $\left.\begin{array}{\|c\|} \hline \text { QTY } \\ \text { PER } \\ \text { ASSY } \end{array} \right\rvert\,$ | MANUFACTURER'S PART NO. | MFR. CODE | RECM VENDOR |
| :---: | :---: | :---: | :---: | :---: | :---: |
| R22 | Resistor, Fixed, Film: $750 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-750 OHMS/J | 09021 |  |
| R23 | Same as R9 |  |  |  |  |
| R24 | Same as R9 |  |  |  |  |
| RN1 | Resistor, Network: $10 \mathrm{k} \Omega, 2 \%, 3 / 16 \mathrm{~W}$ | 3 | $765-1-\mathrm{R} 10 \mathrm{~K}$ | 73138 |  |
| RN2 | Same as RN1 |  |  |  |  |
| RN3 | Resistor, Network: $10 \mathrm{k} \Omega, 2 \%, 3 / 16 \mathrm{~W}$ | 1 | 764-1-R10K | 73138 |  |
| RN4 | Same as RN1 |  |  |  |  |
| U1 |  |  |  |  |  |
| Thru U4 | Not Used |  |  |  |  |
| U5 | Integrated Circuit | 2 | MM74C374N | 27014 |  |
| U6 | Not Used |  |  |  |  |
| U7 | Not Used |  |  |  |  |
| U8 | Integrated Circuit | 1 | MD74SC374AC | 36665 |  |
| U9 | Same as U5 |  |  |  |  |
| U10 | Integrated Circuit | 1 | MM74PC00N | 27014 |  |
| U11 | Integrated Circuit | 2 | MM74PC138N | 27014 |  |
| U12 | Same as U11 |  |  |  |  |
| U13 | Integrated Circuit | 2 | MM74PC32N | 27014 |  |
| U14 | Integrated Circuit | 1 | MM74PC04N | 27014 |  |
| U15 | Integrated Circuit | 1 | MM74PC08N | 27014 |  |
| U16 | Integrated Circuit | 1 | CD4066BE | 02735 |  |
| U17 | Not Used |  | N |  |  |
| U18 | Integrated Circuit | 1 | MM74C14N | 27014 |  |
| U19 | Same as U13 |  |  |  |  |
| U20 | Integrated Circuit | 1 | MD74SC245AC | 36665 |  |
| U21 | Integrated Circuit | 1 | CD4013BF | 02735 |  |
| U22 | Integrated Circuit | 1 | MC14053BCP | 04713 |  |
| U23 | Integrated Circuit | 1 | DAC0830LCN | 27014 |  |
| U24 | Integrated Circuit | 1 | TL064CN | 01295 |  |
| U25 | Integrated Circuit | 1 | ADC0808CCN | 27014 |  |
| VR1 | Diode, Zener: 5.1 V | 3 | 1N751A | 80131 |  |
| VR2 | Same as VR1 |  |  |  |  |
| VR3 | Same as VR1 |  |  |  |  |
| XU1 | Socket, Integrated Circuit | 7 | ICN-163-S3-T | 06776 |  |
| XU2 | Same as XU1 |  |  |  |  |
| XU3 | Socket, Integrated Circuit | 8 | ICN-203-S3-T | 06776 |  |
| XU4 | Same as XU3 |  |  |  |  |
| XU5 | Same as XU3 |  |  |  |  |
| XU6 | Same as XU1 |  |  |  |  |



Figure 5-10. Type 794308-2 IF Interface (MFP-A3), Location of Components

Courtesy of http://BlackRadios.terryo.org

REF DESIG PREFIX MFP-A3

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{array}{c\|} \text { RECM } \\ \text { VENDOR } \end{array}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { XU7 } \\ & \text { Thru } \\ & \text { XU9 } \end{aligned}$ | Same as XU3 |  |  |  |  |
| XU10 | Socket, Integrated Circuit | 9 | ICN-143-S3-T | 06776 |  |
| XU11 | Same as XU1 |  |  |  |  |
| XU12 | Same as XU1 |  |  |  |  |
| XU13 <br> Thru <br> XU16 | Same as XU10 |  |  |  |  |
| XU17 | Same as XU1 |  |  |  |  |
| XU18 | Same as XU10 |  |  |  |  |
| XU19 | Same as XU10 |  |  |  |  |
| XU20 | Same as XU3 |  |  |  |  |
| XU21 | Same as XU10 |  |  |  |  |
| XU22 | Same as XU1 |  |  |  |  |
| XU23 | Same as XU3 |  |  |  |  |
| XU24 | Same as XU10 |  |  |  |  |
| XU25 | Socket, Integrated Circuit | 1 | ICN-286-S5-T | 06776 |  |

5.6.5 TYPE 794275-X SYNTHESIZER INTERFACE/MEMORY REF DESIG PREFIX MFP-A4

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION | $\begin{array}{\|l\|} \hline \text { QTY } \\ \text { PER } \\ \text { ASSY } \end{array}$ | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{aligned} & \text { RECM } \\ & \text { VENDOR } \end{aligned}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| BT1 C1 C2 Thru C5 C6 C7 C8 C9 C9 C10 C11 C12 CR1 CR2 Q1 R1 R2 R3 R4 R5 R6 R7 R8 R9 R10 Thru R13 R14 R15 Thru R19 U1* U2 U2 U3 U4 U5 U5 U3 | Battery: 2.4 V <br> Capacitor, Ceramic, Dise: $0.1 \mu \mathrm{~F}, 20 \%, 100 \mathrm{~V}$ <br> Same as C1 <br> Capacitor, Electrolytic, Tantalum: $22 \mu \mathrm{~F}, 20 \%, 15 \mathrm{~V}$ <br> Same as C6 <br> Capacitor, Ceramic, Dise: $680 \mathrm{pF}, 5 \%, 100 \mathrm{~V}$ <br> Capacitor, Ceramic, Disc: $0.47 \mu \mathrm{~F}, 20 \%, 50 \mathrm{~V}$ <br> Capacitor, Polycarbonate, Tubular: $1 \mu \mathrm{~F}, 10 \%, 30 \mathrm{~V}$ <br> Same as C1 <br> Same as C1 <br> Diode <br> Same as CR1 <br> Transistor <br> Not Used <br> Resistor, Fixed, Film: $820 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ <br> Resistor, Fixed, Film: $82 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ <br> Resistor, Fixed, Film: $150 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ <br> Resistor, Fixed, Film: $390 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ <br> Same as R3 <br> Resistor, Fixed, Film: $10 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ <br> Same as R5 <br> Resistor, Fixed, Film: $1.0 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ <br> Same as R9 <br> Resistor, Fixed, Film: $3.3 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ <br> Same as R7 <br> Integrated Circuit (EPROM) <br> Not Used <br> Integrated Circuit <br> Integrated Circuit <br> Integrated Circuit <br> Same as U5 <br> * Nominal value, final value factory selected. <br> ** Customer selected due to specific software package. | 1 7 1 2 1 1 1 1 1 2 1 1 2 1 2 | ```41B901BD16G1 8131M100-651-104M 196D226X0015KE3 8121-100-C0G0-681J 34452-1 ECR105AK 1N4449 2N3251 CF1/4-820 OHMS/J CF1/4-82K/J CF1/4-150 OHMS/J CF1/4-390 OHMS/J CF1/4-10K/J CF1/4-1.0K/J CF1/4-3.3K/J B2732 841093 MD74SC245AC MD74SC374AC``` | 19209 <br> 72982 <br> 56289 <br> 72982 <br> 14632 <br> 50558 <br> 80131 <br> 80131 <br> 09021 <br> 09021 <br> 09021 <br> 09021 <br> 09021 <br> 09021 <br> 14632 <br> 36665 <br> 36665 <br> 09621 |  |

Courtesy of http://BlackRadios.terryo.org


Figure 5-11. Type 794275-X Synthesizer Interface/Memory (MFP-A4), Location of Components

## Courtesy of http://BlackRadios.terryo.org

REF DESIG PREFIX MFP-A4

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| U7 | Integrated Circuit | 2 | MD74SC138AC | 36665 |  |
| U8 | Same as U7 |  |  |  |  |
| U9 | Integrated Circuit | 1 | MM74PC00N | 27014 |  |
| U10 | Integrated Circuit | 1 | SN74LS09N | 01295 |  |
| U11 | Integrated Circuit | 1 | SN74LS14N | 01295 |  |
| U12 | Integrated Circuit | 5 | MM74C374N | 27014 |  |
| U13 |  |  |  |  |  |
| $\begin{aligned} & \text { Thru } \\ & \text { U16 } \end{aligned}$ | Same as U12 |  |  |  |  |
| U17 | Not Used |  |  |  |  |
| U18 | Integrated Circuit | 1 | P8085A | 34649 |  |
| VR1 | Diode, Zener: 8.2 V | 3 | 1N756A | 80131 |  |
| XU1 | Socket, Integrated Circuit | 3 | ICN-246-S5-T | 06776 |  |
| XU2 | Same as XU1 |  |  |  |  |
| XU3 | Same as XU1 |  |  |  |  |
| XU4 | Socket, Integrated Circuit | 9 | ICN-203-S3-T | 06776 |  |
| XU5 | Same as XU4 |  |  |  |  |
| XU6 | Same as XU4 |  |  |  |  |
| XU7 | Socket, Integrated Circuit | 2 | ICN-163-S3-T | 06776 |  |
| XU8 | Same as XU7 |  |  |  |  |
| XU9 | Socket, Integrated Circuit | 3 | ICN-143-S3-T | 06776 |  |
| XU10 | Same as XU9 |  |  |  |  |
| XU11 | Same as XU9 |  |  |  |  |
| $\begin{aligned} & \text { XU12 } \\ & \text { Thru } \\ & \text { XU17 } \end{aligned}$ | Same as XU4 |  |  |  |  |
| XU18 | Socket, Integrated Circuit | 1 | ICN-406-S5-T | 06776 |  |

Courtesy of http://BlackRadios.terryo.org

| REF DESIG | DESCRIPTION | $\begin{array}{\|l\|} \hline \text { QTY } \\ \text { PER } \\ \text { ASSY } \end{array}$ | MANUFACTURER'S PART NO. | $\begin{aligned} & \text { MFR. } \\ & \text { CODE } \end{aligned}$ | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 488-A1 | Not Used |  |  |  |  |
| 488-A2 | Not Used |  |  |  |  |
| 488-A3 | I/O Interface | 1 | 796075 | 14632 |  |
| 488-J1 | Connector, Receptacle (Part of W2) | 1 | 57-20240-2 | 02660 |  |
| 488-P1 | Not Used |  |  |  |  |
| 488-P2 | Not Used |  |  |  |  |
| 488-P3 | Not Used |  |  |  |  |
| 488-P4 | Connector, Plug (Part of W2) | 1 | 2-87456-0 | 00779 |  |
| 488-W1 | Not Used |  |  |  |  |
| 488-W2 | Cable Assembly | 1 | 380095-1 | 14632 |  |



Figure 5-12. Type 796075 I/O Interface (488M-A3), Location of Components

| 5.7.1 | TYPE 796075 I/O INTERFACE | REF | DESIG PREFIX 488M-A3 |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION | $\begin{array}{\|l\|} \hline \text { QTY } \\ \text { PER } \\ \text { ASSY } \end{array}$ | MANUFACTURER'S PART NO. | $\begin{aligned} & \text { MFR. } \\ & \text { CODE } \end{aligned}$ | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| C1 | Capacitor, Ceramic, Disc: $0.01 \mu \mathrm{~F}, 20 \%, 50 \mathrm{~V}$ | 1 | 34453-1 | 14632 |  |
| C2 | Capacitor, Electrolytic, Tantalum: $22 \mu \mathrm{~F}, 20 \%, 15 \mathrm{~V}$ | 1 | 196D226X0015KE3 | 56289 |  |
| C3 | Capacitor, Ceramic, Disc: $0.1 \mu \mathrm{~F}, 20 \%, 50 \mathrm{~V}$ | 3 | 34475-1 | 14632 |  |
| C4 | Same as C3 |  |  |  |  |
| C5 | Same as C3 |  |  |  |  |
| J1 | Connector, Receptacle | 1 | 87571-9 | 00779 |  |
| R1 | Resistor, Fixed, Film: $43 \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-43 OHMS/J | 09021 |  |
| R2 | Resistor, Fixed, Film: $10 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 1 | CF1/4-10 K/J | 09021 |  |
| R3 | Not Used |  |  |  |  |
| R4 | Not Used |  |  |  |  |
| R5 | Resistor, Fixed, Film: $2.2 \mathrm{k} \Omega, 5 \%, 1 / 4 \mathrm{~W}$ | 3 | CF1/4-2.2K/J | 09021 |  |
| R6 | Same as R5 |  |  |  |  |
| R7 | Same as R5 |  |  |  |  |
| S1 | Switch, Dipped | 1 | P1P8 | 01686 |  |
| U1 | Integrated Circuit | 1 | P8291A | 34649 |  |
| U2 | Integrated Circuit | 1 | SN75160N | 01295 |  |
| U3 | Integrated Circuit | 1 | SN75161N | 01295 |  |
| U4 | Integrated Circuit | 1 | DM811S96N | 27014 |  |
| U5 | Resistor/Network | 1 | 765-1-R10K | 73138 |  |
| U6 | Integrated Circuit | 1 | SN74LS273N | 01295 |  |
| U7 | Integrated Circuit | 1 | SN74LS138N | 01295 |  |
| U8 | Integrated Circuit | 1 | SN74LS14N | 01295 |  |
| U9 | Integrated Circuit | 1 | SN74LS32N | 01295 |  |
| U10 | Integrated Circuit | 1 | SN74LS74N | 01295 |  |
| XU1 | Socket, Integrated Circuit | 1 | ICN-406-S5-T | 06776 |  |
| XU2 | Socket, Integrated Circuit | 1 | ICN-203-S3-T | 06776 |  |
| XU3 | Same as XU2 |  |  |  |  |
| XU4 | same as XU2 |  |  |  |  |
| XU5 | Not Used |  |  |  |  |
| XU6 | Same as XU2 |  |  |  |  |
| XU7 | Socket, Integrated Circuit | 1 | ICN-163-S3-T |  |  |
| XU8 | Socket, Integrated Circuit | 1 | ICN-143-S3-T | 06776 |  |

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5.8 TYPE WJ-8718A/232M I/O OPTION (Accessory item for use with 8718A/MFP)

| REF DESIG | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 232-A1 | Not Used |  |  |  |  |
| 232-A2 | Not Used |  |  |  |  |
| 232-A3 | I/O Interface | 1 | 796037 | 14632 |  |
| 232-J1 | Connector, Receptacle (Part of W2) | 1 |  |  |  |
| 232-P1 | Not Used |  |  |  |  |
| 232-P2 | Not Used |  |  |  |  |
| 232-P3 | Not Used |  |  |  |  |
| 232-P4 | Connector, Plug (Part of W2) | 1 | 88475-4 | 00779 |  |
| 232-W1 | Not Used |  |  |  |  |
| 232-W2 | Cable Assembly | 1 | 380059-1 | 14632 |  |



Figure 5-13. Type 796037 I/O Interface (232M-A3),
Location of Components

TYPE 796037 I/O INTERFACE
REF DESIG PREFIX 232M-A3

| $\begin{aligned} & \text { REF } \\ & \text { DESIG } \end{aligned}$ | DESCRIPTION |  | MANUFACTURER'S PART NO. | MFR. CODE | $\begin{gathered} \text { RECM } \\ \text { VENDOR } \end{gathered}$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| C1 | Capacitor, Electrolytic, Tantalum: $22 \mu \mathrm{~F}, 20 \%, 15 \mathrm{~V}$ | 3 | 196D226X0015KE3 | 56289 |  |
| C2 | Same as C1 |  |  |  |  |
| C3 | Same as C1 |  |  |  |  |
| C4 | Capacitor, Ceramic, Disc: $0.1 \mu \mathrm{~F}, 20 \%, 100 \mathrm{~V}$ | 5 | 8131M100-651-104M | 72982 |  |
| $\begin{aligned} & \text { C5 } \\ & \text { Thru } \\ & \text { C8 } \end{aligned}$ | Same as C4 |  |  |  |  |
| C9 | Capacitor, Ceramic, Mono: $470 \mathrm{pF}, 5 \%, 100 \mathrm{~V}$ | 6 | 8121-100-C0G0-471J | 72982 |  |
| $\begin{array}{\|l} \hline \text { C10 } \\ \text { Thru } \\ \text { C14 } \end{array}$ | Same as C9 |  |  |  |  |
| J1 | Connector, Receptacle | 2 | 87567-6 | 00779 |  |
| J2 | Same as J1 |  |  |  |  |
| R1 | Resistor, Fixed, Film: 100 , 5\%, 1/4 W | 2 | CF1/4-100 OHMS/J | 09021 |  |
| R2 | Same as R1 |  |  |  |  |
| R3 | Not Used |  |  |  |  |
| R4 | Resistor, Fixed, Film: $3.3 \mathrm{k} \Omega, 5 \%, 1 / 8 \mathrm{~W}$ | 1 | CF1/8-3.3K/J | 09021 |  |
| R5 | Resistor, Fixed, Film: $10 \mathrm{k} \Omega, 5 \%, 1 / 8 \mathrm{~W}$ | 4 | CF1/8-10K/J | 09021 |  |
| $\begin{array}{\|l} \text { R6 } \\ \text { Thru } \\ \text { R8 } \end{array}$ | Same as R5 |  |  |  |  |
| S1 | Switch, Slide | 1 | 206-4 | 71450 |  |
| S2 | Switch, Slide | 1 | 206-8 | 71450 |  |
| U1 | Integrated Circuit | 1 | P8251A | 34649 |  |
| U2 | Integrated Circuit | 1 | MM74C374N | 27014 |  |
| U3 | Integrated Circuit | 1 | DM81LS96N | 27014 |  |
| U4 | Integrated Circuit | 1 | MM74HC138N | 27014 |  |
| U5 | Integrated Circuit | 1 | COM5046P | 53848 |  |
| U6 | Integrated Circuit | 1 | MM74HC04N | 27014 |  |
| U7 | Integrated Circuit | 2 | MC1488P | 04713 |  |
| U8 | Same as U7 |  |  |  |  |
| U9 | Integrated Circuit | 2 | MC1489P | 04713 |  |
| U10 | Same as U9 |  |  |  |  |
| U11 | Resistor Network | 1 | 764-1-R10K | 73138 |  |
| VR1 | Diode, Zener: 12 V | 2 | 1N759A | 80131 |  |
| VR2 | Same as VR1 |  |  |  |  |
| XU1 | Socket, Integrated Circuit | 1 | ICN-286-S5-T | 06776 |  |
| XU2 | Socket, Integrated Circuit | 2 | 1CN-203-S3-T | 06776 |  |
| XU3 | Same as XU2 |  |  |  |  |
| XU4 | Socket, Integrated Circuit | 1 | ICN-163-S3-T | 06776 |  |
| XU5 | Not Used |  |  |  |  |
| XU6 | Socket, Integrated Circuit | 1 | ICN-143-S3-T | 06776 |  |
| Y1 | Crystal | 1 | 91805-4 | 14632 |  |

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Courtesy of http://BlackRadios.terryo.org

## SECTION VI

SCHEMATIC DIAGRAMS

Courtesy of http://BlackRadios.terryo.org

## Courtesy of http://BlackRadios.terryo.org




Figure 6-2. Type 794310-1-2-3 Front Panel Interconnect (MFP-A1), Schematic Diagram 570291

## Courtesy of http://BlackRadios.terryo.org

```
Nores:
```



```
\begin{tabular}{|c|c|c|c|c|}
\hline \multicolumn{5}{|c|}{tablea} \\
\hline & 6NO & vec, & -5V & \\
\hline U1,U2 & 20 & 40 & & \\
\hline U12.urtulu & - & 16 & & \\
\hline U4, U8 & 12 & 24 & & \\
\hline US, U6, Us, uio & 9 & 18 & & \\
\hline U13015.016 & 7 & 14 & & \\
\hline 0.14 & \({ }^{8}\) & 16 & 7 & \\
\hline U18, U2\% & 10 & 20 & & \\
\hline U3,07 & 8 & & & 16 \\
\hline
\end{tabular}
```





``` \begin{tabular}{c|c}
\(796055-1\) & STANDARD \\
\hline \(79655-2\) & Cowformal coatima \\
\hline
\end{tabular}
```




Figure 6-4. Type 794309-1-2-3 Front Panel Switch Board (MFP-A1A2), Schematic Diagram 570287

## Courtesy of http://BlackRadios.terryo.org




```
0日暗
M,
*)
```



## Courtesy of http://BlackRadios.terryo.org

```
M,
*)
```



## Courtesy of http://BlackRadios.terryo.org



Figure 6-7. Type 796075 I/O Interface ( $488 \mathrm{M}-\mathrm{A} 3$ ), Schematic Diagram 480096

## Courtesy of http://BlackRadios.terryo.org




